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THESIS

DESIGN, CONSTRUCTION AND TESTING OF A
PROTOTYPE FIN-LINE MAGIC-TEE AND FIN-LINE
MONOPULSE SYSTEM SUITABLE FOR
MILLIMETER-WAVE APPLICATIONS

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December 1985

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86 2 25 034

AD-A164521

REPORT DOCUMENTATION PAGE

1a. REPORT SECURITY CLASSIFICATION		1b. RESTRICTIVE MARKINGS			
2a. SECURITY CLASSIFICATION AUTHORITY		3. DISTRIBUTION/AVAILABILITY OF REPORT			
2b. DECLASSIFICATION/DOWNGRADING SCHEDULE		Approved for public release; distribution is unlimited.			
4 PERFORMING ORGANIZATION REPORT NUMBER(S)		5. MONITORING ORGANIZATION REPORT NUMBER(S)			
6a. NAME OF PERFORMING ORGANIZATION Naval Postgraduate School	6b. OFFICE SYMBOL (If applicable) Code 62	7a. NAME OF MONITORING ORGANIZATION Naval Postgraduate School			
6c. ADDRESS (City, State, and ZIP Code) Monterey, California 93943-5004		7b. ADDRESS (City, State, and ZIP Code) Monterey, California 93943-5004			
8a. NAME OF FUNDING/SPONSORING ORGANIZATION	8b. OFFICE SYMBOL (If applicable)	9. PROCUREMENT INSTRUMENT IDENTIFICATION NUMBER			
8c. ADDRESS (City, State, and ZIP Code)		10. SOURCE OF FUNDING NUMBERS			
		PROGRAM ELEMENT NO.	PROJECT NO	TASK NO.	WORK UNIT ACCESSION NO.

11. TITLE (Include Security Classification) DESIGN, CONSTRUCTION AND TESTING OF A PROTOTYPE FIN-LINE MAGIC-TEE AND FIN-LINE MONOPULSE SYSTEM SUITALBE FOR MILLIMETER-WAVE APPLICATIONS

12. PERSONAL AUTHOR(S)
Rowley, James William

13a TYPE OF REPORT Master's thesis	13b TIME COVERED FROM _____ TO _____	14 DATE OF REPORT (Year, Month, Day) December 1985	15 PAGE COUNT 114

16 SUPPLEMENTARY NOTATION

17	COSATI CODES		18. SUBJECT TERMS (Continue on reverse if necessary and identify by block number)
FIELD	GROUP	SUB-GROUP	millimeter wave comparator; fin-line comparator; fin-line magic-tee; monopulse comparator; monopulse antenna; monopulse radar

19 ABSTRACT (Continue on reverse if necessary and identify by block number)
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The three-dimensional junction of a waveguide magic-tee is replaced with fin-line slots, coupled fin-line slots and microstrip lines mounted in a waveguide fixture. The planar geometry on the substrate provides significant reduction in size and eliminates the waveguide surface that is associated with conventional hybrids. Ports one and two are flared into fin-line horns to produce a fin-line monopulse system. Suggestions for further development of the fin-line magic-tee and monopulse system are presented.

20 DISTRIBUTION/AVAILABILITY OF ABSTRACT ☒ UNCLASSIFIED/UNLIMITED ☐ SAME AS RPT ☐ DTIC USERS	21. ABSTRACT SECURITY CLASSIFICATION Unclassified	
22a NAME OF RESPONSIBLE INDIVIDUAL Jefferey B. Knorr	22b TELEPHONE (Include Area Code) (408) 646-2815	22c OFFICE SYMBOL Code 62Ko

DD FORM 1473, 84 MAR   83 APR edition may be used until exhausted   SECURITY CLASSIFICATION OF THIS PAGE
All other editions are obsolete

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Design, Construction and Testing of a Prototype Fin-Line
Magic-Tee and Fin-Line Monopulse System Suitable for
Millimeter-Wave Applications

by

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Submitted in partial fullfillment of the
requirements for the degree of

MASTER OF SCIENCE IN ELECTRICAL ENGINEERING

from the

NAVAL POSTGRADUATE SCHOOL
December 1985

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ABSTRACT

This thesis describes a fin-line 180 degree hybrid (magic-tee) that is suitable for use in monopulse radar antennas at microwave and millimeter-wave frequencies.

The three-dimensional junction of a waveguide magic-tee is replaced with fin-line slots, coupled fin-line slots and microstrip lines mounted in a waveguide fixture. The planar geometry on the substrate provides significant reduction in size and eliminates the waveguide ratrace that is associated with conventional hybrids. Ports one and two are flared into fin-line horns to produce a fin-line monopulse system. Suggestions for further development of the fin-line magic-tee and monopulse system are presented.

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ACKNOWLEDGMENT

I. INTRODUCTION

A. BACKGROUND

This thesis describes the successful design, construction and testing of both a fin-line magic-tee and monopulse system. Each of these units is probably the first devices of their type constructed.

There are two main categories of magic-tees: the conventional waveguide hybrids, and the new planar devices [Ref. 1: pp. 259-261]. The fin-line magic-tee, developed in this thesis, is in the latter category.

The 180 degree hybrid, or magic-tee, is a passive four port device. It is an essential part of many microwave components, including balanced mixers, single-sideband modulators, frequency multiplexers, linear phase shifters, constant impedance filters, IFM (instantaneous frequency measurement) receivers, interferometers, duplexers and monopulse radar comparators [Ref. 1: pp. 259-261 and Ref. 2: pp. 16-18]. While many of these applications use only one hybrid, monopulse antennas contain a complex network of magic-tee comparators.

Monopulse comparator networks require exact symmetry between numerous sections of transmission lines for proper operation. Unbalanced phase shifts of only a few degrees,

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due to uneven lengths of transmission line, destroy the target tracking capability of the system [Ref. 3].

Wavequide comparator networks, with bolted together flanges or soldered connections, have difficulty achieving these exacting lengths at millimeter-wave frequencies. This effect limits most of the current monopulse radars to frequencies well below the millimeter-wave band [Ref. 3].

Another limitation of the wavequide hybrid occurs because three of the four ports are on mutually orthogonal axes. It is this three dimensional shape that produces the complex wavequide ratrace and intricate network of soldered flanges in conventional monopulse antennas [Ref. 2: pp. 10-17 and Ref. 3].

Unlike the wavequide magic-tee, all four ports of the new hybrids are in the same plane. This two dimensional geometry greatly simplifies the interconnections between the numerous comparators in a monopulse antenna system. With the planar magic-tees, the confusing and costly maze of wavequide discussed above can be eliminated.

In addition to the fin-line magic-tee designed in this thesis, two other types of planar magic-tees have recently been developed [Ref. 3 and Ref. 4: pp. 523-528]. However, only one of these new hybrids is designed as a monopulse comparator [Ref. 3].

The fin-line magic-tee developed in this thesis and the new fin-line horn antenna developed in a parallel thesis [Ref. 5] are designed to be computer manufactured as a

single unit. This combination of fin-line horns and fin-line magic-tees constructed in an integrated unit, is hereafter called a fin-line monopulse system, or a fin-line monopulse comparator. Fin-line monopulse systems, operating at millimeter-wave frequencies, can potentially be mass produced for a few hundred dollars each.

B. RELATED WORK

1. Slotline Magic-Tee

Mr. M. Aikawa and Mr. H. Ogawa [Ref. 4: pp. 523-528] introduced a slotline magic-tee. The fin-line magic-tee developed in this thesis (Fig. 1) is similar to two of these slotline hybrids placed back to back in a waveguide fixture.

There are five major differences between the fin-line magic-tee developed in this thesis and the slotline magic-tee described in Reference 4.

First, the fin-line hybrid is essentially two slotline hybrids in parallel with each other. This configuration significantly changes the impedance characteristics of the circuit.

Second, the waveguide fixture that surrounds the fin-line device, changes all of the impedances as a function of frequency.

Third, the unique microstrip to coaxial transition used in the fin-line device does not have a counterpart in the slotline magic-tee. The purpose of this transition is to

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maintain a fairly wide bandwidth at the point where the microstrip passes through the waveguide wall.

Fourth, the slotline hybrid uses a tapered section to transition from the main loop to the coupled slot area. In the corresponding portion of the fin-line magic-tee, each section of microstrip maintains a constant width up to the point where the two sections join.

Finally, all of the bends in the slots of the fin-line magic-tee are computer designed to keep incidental reflections as low as possible.

2. Planar Waveguide Comparators

The system described by Syrigos, Crossland and Van Wyck [Ref. 3] is not related to this thesis at all. However, it does meet part of the objectives of this work by utilizing a totally different design and manufacturing concept.

3. Fin-Line Horn Antenna

The fin-line horn antenna was perfected by LCDR Mumtaz-ul-Haq [Ref. 5]. Haq discovered that a substantial amount of the electric field can be launched between the parallel fins. This undesirable effect can be successfully controlled by making the horn fins half of a wavelength wide and shorting the exposed edges together with copper tape. This short reflects another short across the edges of the fin-line slots. The reflected short seals the gap between the fins and keeps the electric field in the slots.

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The parallel slots leading to ports one and two (Fig. 1) are separated by a quarter wavelength. Each slot appears as an open to the other slot, reflecting back a short a quarter of a wavelength away. The reflected short seals off the two slots, thereby preventing inadvertent coupling. This concept is a direct result of the applied research performed by Haq [Ref. 5].

C. PURPOSE

The first objective of this work is to design, build and successfully test a fin-line magic-tee at 10-GHZ. This includes the design and construction of a suitable fixture to hold the magic-tee. The final magic-tee is to be integrated with two fin-line horn antennas [Ref. 5], to form a two dimensional fin-line monopulse system (only one difference channel).

The second objective is to produce reasonable sum and difference antenna patterns from the combined one piece unit. This integrated unit is intended to be the prototype of a millimeter-wave fin-line monopulse system that has both azimuth and elevation difference channels.

The third and most important objective is to computerize the design procedure. This will insure the ability to accurately and quickly reproduce these devices at a future date.

The final objective of this research is to recommend possible improvements for the fin-line magic-tee and

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fin-line monopulse system. Specific recommendations for
expanding the fin-line monopulse system into a three channel
(sum channel, azimuth channel and elevation channel)
monopulse system are required.

II. THEORETICAL PRINCIPLES

The design of the fin-line magic-tee and monopulse
system is based on established microwave principles. These
theoretical concepts are discussed in this chapter.

A. WAVEGUIDE MAGIC-TEE

The theoretical properties of a waveguide 180 degree
hybrid are illustrated in Figures 2 and 3. An input at any
one of the four ports is equally divided with half of the
power coupling into two orthogonal ports. In phase signals
at ports one and two combine in the H-plane arm and cancel
in the E-plane arm. Out of phase inputs at ports one and two
produce the opposite results. There is complete isolation
between ports one and two and between ports three and four.
The reverse of these conditions is also true. The 180 degree
phase shift depicted in Figure 3 can occur in either $S23$ and
$S32$ as shown, or in $S13$ and $S31$ [Ref. 6: pp. 190-191].

B. MONOPULSE ANTENNAS

The typical monopulse radar antenna contains four
identical antenna elements, which are interconnected with
three or more magic-tees. The signals to and from the four
elements are added and subtracted in various combinations to
produce three system ports.

All four of the antennas are summed together in phase to
produce the sum channel. This signal is connected to the
radar via a T/R (Transmit/Receive) Device.

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The two difference channels are the radar's source of target tracking information. The elevation port produces the difference between the upper and lower antennas while the azimuth port produces the difference between the left and right antennas. If a monopulse antenna system is pointing exactly at a target, there will be a strong signal in the sum channel and absolutely no signal in either the elevation or azimuth channels. The null in the difference channels is the result of shifting two identical signals 180 degrees from each other and then adding them together. The phase shift and addition occur within the magic-tees.

The radar return from a target that is slightly left of the monopulse antenna's extended center line reaches the left elements of the antenna before it reaches the right elements. This produces a slight phase shift due to the difference in arrival times at the left and right elements. Complete cancellation does not occur in the difference channel with this configuration. The resulting difference signal increases in amplitude and shifts in phase as the target gets farther away from the antennas extended center line. There is also a 180 degree phase shift in the difference channel as a target crosses the antenna's extended center line [Ref. 7: pp. 10-29].

Since the null in the difference port can only occur with exact target/antenna alignment, it signifies the exact

center of both the sum and difference antenna patterns. Therefore, a target that produces a strong return in the sum channel and no return in one of the difference channels is on a plane that bisects the two antennas which develop the difference signal. When this information is combined with range data, the target's location is limited to an arc on the plane that bisects the two halves of the antenna. The point where the elevation and azimuth arcs cross is directly in front of the antenna. The line between the monopulse antenna and this point is commonly referred to as the boresight of the antenna.

If the outputs of the difference channels are monitored while a target is within the main beam of the sum pattern, the target can be classified as exactly centered, a little left or right, a little high or low, or any combination of these directions.

Skolnik [Ref. 7: pp.10-28] defines an amplitude comparison monopulse system as one that compares the amplitude of the difference channel with the amplitude of the sum channel. This information is used to determine how far a target is away from the antenna's boresight. The phase (+/- only) of the difference signal is used to determine which side of boresight the target is on.

This technique was initially called simultaneous lobing since all of the radiation lobes are sampled during each and every pulse. The ability to obtain a complete tracking

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solution in only one pulse led to the current designation of monopulse [Ref. 7: p. 10].

Earlier tracking radars such as conical scanning and lobe switching systems required numerous radar returns to obtain the same information. The accuracy of these systems is often degraded by pulse to pulse amplitude variations. Monopulse radars, which are free of this distortion, have achieved tracking accuracies of 0.003 degrees [Ref. 7: P. 10].

Two popular forms of amplitude comparison monopulse antennas are illustrated in Figures 4 and 5.

C. THEORETICAL ANTENNA PATTERNS

The far field effect an antenna has on the environment is a direct result of the current distribution across the face of the antenna and can be represented mathematically with a specialized Fourier Transform [Ref. 8: pp. 345-369 and Ref. 9: pp. 170-195]. This technique is used to develop computer simulations of actual sum and difference patterns. The simulated patterns are used to illustrate the theoretical properties of a monopulse system.

The first step in predicting a theoretical monopulse antenna pattern is to obtain the element pattern by taking the Fourier Transform of the current distribution across the face of one of the elements. The second step is to determine the array or group pattern by taking the Fourier Transform of the entire array, assuming that each antenna element is a

delta function. The final step is to multiply the element pattern with the array pattern.

An actual E-plane antenna pattern from one of the fin-line horns of Reference 5 is closely simulated by adding 20% of a sin(x)/x pattern, to 80% of sin(x)/x squared pattern (Fig. 6). This simulation is used as the predicted element pattern.

The group pattern is obtained by taking the trivial transformation of two delta functions. This results in a cosine function for the sum pattern and a sine function for the difference pattern. These theoretical group patterns with an element spacing of one wavelength are illustrated in Figures 7 and 8.

The multiplication of the simulated element and group pattern is done in a short Basic program on an HP-9845B Computer. The theoretical sum and difference patterns computed in this program are shown in Figures 9 and 10, respectively.

D. SLOTLINE CHARACTERISTICS

Slotline consists of a narrow slot in a thin plating of metal foil adhered to one side of a thin layer of dielectric. The other side of the supporting substrate is void of metal. The electric field is guided between the edges of the slot. The impedance of the slot is directly proportional to the width of the slot. A detailed analysis

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of slotline properties is presented by Gupta, Garg and Bahl
[Ref. 10: pp. 195-228].

E. FIN-LINE CHARACTERISTICS

The bilateral fin-line illustrated in Figure 11 has one
straight slot on each side of the dielectric. The fin-line
magic-tee uses a similar, but more complex configuration
which has two slots on each side of the dielectric. The
slots in the magic-tee have a mixture of straight, angled
and curved sections which join together in a 180 degree bend
(Fig. 1). Both of these designs are formed by suspending
dual sided slotline in the E-plane of a section of
waveguide. As such, they are essentially shielded slotline.
As is the case with slotline, the impedance of fin-line is
directly proportional to the width of the slot. The detailed
properties of fin-line are discussed by Sharma and Hoefer
[Ref. 11: pp. 350-355] and Meier [Ref. 12: pp. 1209-1215].

F. EVEN AND ODD MODES IN COUPLED SLOTS

Two identical and parallel slots in either slotline or
fin-line are considered coupled if the impedance and
electrical length of one slot is effected by the proximity
of the other slot. Coupled slots are the key to the
operation of both the slotline magic-tee discussed in
Reference 4 and the fin-line magic-tee developed in this
thesis.

Knorr and Kuchler [Ref. 13: PP. 541-547] define two
dominate modes in coupled slots: even or odd. The even mode
exists when the electric fields within the two slots are in

phase with each other; whereas, the odd mode exists when the two electric fields are out of phase with each other.

In the even mode, the magnetic fields which surround each slot couple together smoothly to form one continuous magnetic field. In this configuration, the fields aid each other and produce a coupled impedance which is lower than the uncoupled impedance [Ref. 10: p. 352].

As the slots are brought closer together, the coupling increases. This effect causes the even mode impedance to decrease. In the limit when the separation between the slots vanishes, each slot has half of the impedance of the new slot which is twice as wide as each of the original slots. At the other limit when the distance between the slots is infinite, each slot retains its uncoupled characteristics [Ref. 10: p. 355].

In the odd mode, the magnetic fields oppose each other and do not join together smoothly. Because of this opposition, the odd mode impedance is higher than the uncoupled impedance. As the two slots are brought closer together, the opposition between the two magnetic fields and the odd mode impedance increase [Ref. 10: p. 352].

In the limit, when the separation between the slots vanishes, there is complete opposition between two equal and opposite magnetic fields. In this condition, the odd mode impedance of each slot is exactly twice the uncoupled impedance of one of the original slots [Ref. 10: p. 355].

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The electrical length of coupled slots is longer than the uncoupled length in the even mode and shorter than the uncoupled length in the odd mode. These differences become larger as the two slots are brought closer together [Ref. 13: pp. 543-545].

G. MICROSTRIP CHARACTERISTICS

The microstrip that is used in this thesis is composed of a narrow center conductor with dielectric material and a ground plane on both sides. If the two ground planes are bent around the center conductor until they touch, it will resemble coaxial cable. This dual sided microstrip, defined as "triplate line" by Reference 6, and coaxial cable both operate in the TEM mode [Ref. 6: pp. 38-61].

H. MICROSTRIP TO SLOTLINE TRANSITIONS

A simple and effective transition from slotline to microstrip is presented by Knorr [Ref. 14: pp. 548-553]. With this technique an open length of microstrip on one side of a section of dielectric overlaps a shorted slot on the other side of the same substrate at a 90 degree angle. The portion of the slot that extends past the microstrip, and the portion of the microstrip that extends past the slot are exactly one quarter of a wavelength long. At the junction there is a reflected short in the microstrip and a reflected open in the slot. This procedure is used in both the fin-line magic-tee and the fin-line monopulse system.

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With thin substrates and high dielectric constants, the power transfer through this type of transition is close to 1:1 [Ref. 14: pp. 548-553]. The fin-line magic-tee and the fin-line monopulse system developed in this thesis are made from Epsilam-10 which is 0.025 inches thick and has a dielectric constant of 10.2. The return loss of these transitions is minimal.

I. IMPEDANCE MATCHING

A quarter wave length impedance matching technique is used in the fin-line magic-tee. This procedure joins two slots of unequal impedances together with a quarter wave length matching section. The impedance of the center matching slot is equal to the square root of the product of the two original impedances.

If the original impedances are called Za and Zb, and the matching section is called Zo, the relationship becomes Zo equals the square root of Za times Zb.

This relationship is easily visualized on a normalized Smith Chart. Zo is the center of the chart and Za is somewhere to the left of center on the real line. Traveling a quarter of a wave length through the matching section of impedance Zo is represented on the Smith Chart by a half circle of radius Za, which is centered at Zo. The end of this half circle will be on the real line to the right of Zo. If the algebraic relationship discussed above is maintained, this point will be Zb.

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The proof of this relationship is based on the fact that all of the points on the left half of the real line of a Smith Chart are the reciprocals of the equal distance points on the right side of the real line. Therefore, as long as Za and Zb are on different sides of the real line and equal distance from the center, their product is unity. Since the center of a normalized Smith Chart is Zo which has a value of 1.0, the algebraic relationship discussed above produces a perfect impedance match.

III. FIN-LINE MAGIC-TEE

The fin-line magic-tee is designed to mate with two
fin-line horns from Reference 5 to form a fin-line monopulse
antenna. To provide a smooth transition between the
magic-tee and the horns, a bilateral fin-line magic-tee
enclosed in a waveguide fixture is developed. The dielectric
is Epsilam-10 which has a thickness of 0.025 inches and a
dielectric constant of 10.2.

A. ENGINEERING APPROXIMATIONS

Four major engineering approximations are used in the
design of the magic-tee. These approximations greatly
facilitate the design process without introducing
significant inaccuracies.

1. Parallel Circuit Approximation

An extensive search of the current literature failed
to turn up design data for a complex fin-line structure with
coupled bilateral fin-line slots at a dielectric constant of
10.2. Therefore, the bilateral fin-line is modeled as two
identical unilateral (one sided) circuits in parallel.

Sharma and Hoefer [Ref. 11: pp. 350-355] compared
the properties of bilateral fin-line and unilateral fin-line
with dielectric constants of 2.22 and 3.0. An 80 ohm, 0.050
inch thick section of bilateral fin-line has 57% of the
impedance of a similar section of 0.025 inch thick

unilateral fin-line. Changing the dielectric constant to 3.0 drops the ratio to 55%. The same comparison with 0.025 inch thick, 200 ohm unilateral fin-line produces an 84% ratio at a dielectric constant of 2.22 and a 73% ratio with the 3.0 dielectric constant.

At a dielectric constant of 10.2, the low impedance slots should closely fit the parallel circuit model. There may be a slight error for the largest unilateral impedance in the fin-line magic-tee, which is 200 ohms. However, this error should be minimal.

2. Fixtures Effect on Impedance

Kuchler [Ref. 15: p. 103] compared the impedances of shielded and unshielded slotline with a dielectric constant of 20. Between 10- and 12-GHZ for a dielectric thickness of 0.050 inches, the shielded and unshielded impedances are almost indentical. The impedance of the shielded slot is constant to 6-GHZ. The impedance of the unshielded slot is approximately 10% lower at 6-GHz than it is at 12-GHZ.

The quarter wavelength impedance matching technique discussed in Chapter Two is the only type of impedance matching used in the fin-line magic-tee. This technique's simple algebric relationship will factor out any uniform change in impedance. Therefore, the fixture should not effect the impedance matching between the slots. There may be a slight mismatch between the slots and the microstrip

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leads, but this mismatch should be minimal [Ref. 15: p. 103].

Based on the first two engineering approximations, all of the impedances in the fin-line magic-tee are calculated as slotline impedances with a dielectric thickness of 0.025 inches. All of the slotline impedances are exactly twice the desired bilateral fin-line impedances.

3. Length of the Coupled Slots

The electrical length of coupled slots is different in the even and odd modes [Ref. 13: pp. 541-547]. The actual length of the coupled slots in the slotline magic-tee [Ref. 4: p. 527] is the average of the even and odd mode quarter wavelengths.

Knorr and Kuchler [Ref. 13: pp. 544, 545] graphed the even and odd mode wavelengths for dielectric constants of 11.0 and 16.0. On both of these graphs, the average of the even and odd mode wavelengths is 97% of the uncoupled electrical length. This percentage holds for a very wide range of slot separation distances, including the separation that is used in the fin-line magic-tee.

The length of the coupled slots in the fin-line magic-tee is designed to be 97% of the value of a similar section of uncoupled slotline.

4. Dielectric Termination within the Fixture

Due to the complex geometry of the fin-line magic-tee (Fig. 1) the conventional configuration for the

bilateral fin-line shown in Figure 11 is not feasible. The fin-line magic-tee uses a 0.02 inch groove in the fixture wall to support the dielectric. This groove shorts and electrically seals the edges of the fins.

B. EQUIVALENT CIRCUIT

The fin-line magic-tee is similar to two slotline magic-tees placed back to back and mounted in a waveguide fixture [Ref. 4: pp. 523-527]. The equivalent circuit for the fin-line magic-tee is similar to the circuit shown for the slotline magic-tee on page 525 of Reference 4. The fin-line device is represented by two of these circuits connected in parallel. The fin-line version is surrounded by a shield.

1. Theoretical Operation

An actual 1:1 scale drawing of the fin-line magic-tee is illustrated in Figure 1. With the exception of port three, the entire magic-tee is symmetric about an axis that extends through the center of the port four microstrip line. The theoretical operation of the fin-line magic-tee is identical to the operation of the slotline magic-tee [Ref. 4: pp. 523-527].

The microstrip leads from all four ports use the microstrip to slot transition technique discussed in Chapter Two. The short in the loop caused by the port four microstrip effectively isolates ports one and two. Ports three and four are isolated from each other by the 3/4 of a

wavelength distance between the short in the loop and the short caused by the port three microstrip.

Two signals that enter the coupled slots in the even mode will couple into port three. There will be a slight phase shift between the two signals in the port three microstrip line. This error is inherent in the design of the device and can only be minimized by keeping the slot separation as small as possible. The even mode signals cancel at port four due to the loop geometry.

Two signals that enter the coupled slots in the odd mode will be out of phase at port three but will couple in phase at port four. The theoretical operation of the fin-line magic-tee is identical to the scattering matrix for the waveguide magic-tee (Fig. 3).

2. Theoretical Impedance Matching

The unilateral impedances in the fin-line magic-tee are 200 ohms in the loop, 100 ohms in the slots that connect to ports one and two, and 70.7 ohms and 141.4 ohms for the even and odd mode in the coupled slots, respectively. These reduce to bilateral impedances of 100 ohms, 50 ohms, 35.4 ohms and 70.7 ohms, respectively. These bilateral values are identical to the impedances listed for the "case three" slotline magic-tee [Ref. 4: p. 527]. A detailed discussion of the theoretical impedance matching in the slotline magic-tee is presented by Aikawa and Ogawa [Ref. 4: pp. 524-527].

There are four basic relationships that require impedance matching within the magic-tee. The paths from port three to ports one and two and from port four to ports one and two must be matched in the even and odd modes. In all of these cases the coupled slots act as a quarter wavelength matching section. The impedance matching problem either reduces to three series impedances of 100, 70.7 and 50 ohms. or 50, 35.4 and 25 ohms. In all cases, the impedance of the coupled slots is the square root of the other two impedances.

C. SLOT IMPEDANCE AND ELECTRICAL LENGTH

1. Uncoupled Slots

Cohn [Ref. 16: p. 1092] graphed slotline impedance and effective wavelength for dielectric constants of 9.6 and 11.0. The data points on Figure 12 and 13 are extrapolated from Cohn's graphs for a dielectric constant of 10, a dielectric thickness of 0.025 inches and a frequency of 10-GHZ. All of the slot widths and electrical lengths in the fin-line magic-tee are derived from this information.

2. Coupled Slots

As discussed in Chapter Two, the odd mode impedance of coupled slots with no separation is equal to twice their uncoupled impedance. Their even mode impedance with no separation is one half of the uncoupled impedance of the new larger slot [Ref. 10: p. 355].

Knorr and Kuchler [Ref. 13: pp. 544, 545] graphed the even and odd mode impedances of coupled slots for dielectric constants of 11 and 16. A normalized version of this information is plotted in Figure 14 for a dielectric thickness of 0.025 inches and a frequency of 10-GHZ.

The normalization in Figure 14 is non-standard. The normalizing impedance is the slot impedance for S/D (slot separation/dielectric thickness) approaching infinity. The minimum value of zero is equivalent to the even mode impedance with no slot separation. This correlates to half of the uncoupled impedance of a slot that is twice as wide as the original slots. The center value of 1.0 represents an infinite slot separation. This is equivalent to the uncoupled impedance of each slot. The maximum value of 2.0 is the same as the odd mode impedances with zero slot separation. This corresponds to twice the impedance of the uncoupled slots.

The width of the coupled slots and the separation between the coupled slots are determined by trial and error. During this process impedances from Figure 12 are assigned to the normalized values of zero, 1.0 and 2.0 shown in Figure 14. The curve for a dielectric constant of 11.0 is used in this procedure. Based on the comparison between the curves for dielectric constants of 11.0 and 16.0 (Fig. 14), little error will be induced by this approximation.

First, an arbitrary value for the uncoupled impedance is chosen. To illustrate this procedure, an

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initial value of 100 ohms is used for the uncoupled
impedance. The normalized impedance of 1.0 (Fig. 14) is
assigned this value (100 ohms). The normalized impedance
value of 2.0 (Fig. 14) is set equal to twice this value (200
ohms).

Second, the slot width (0.4340 millimeters) for this
arbitrary impedance (100 ohms) is obtained from the 0.635
millimeter curve in Figure 12. The impedance (132.2 ohms) of
a slot twice this wide (0.8680 millimeters) is also
calculated from Figure 12. The normalized impedance of zero
(Fig. 14) is set equal to half of this value (66.15 ohms).

Third, the impedances in Figure 14 are now scaled
for a dielectric constant of 10. The lower portion of the
graph is linear from 66.15 ohms (normalized value of zero)
to 100.0 ohms (normalized value of 1.0). The upper portion
is linear from 100.0 ohms (normalized value of 1.0) to 200.0
ohms (normalized value of 2.0). On this scaled version of
Figure 14, an S/D is picked that corresponds to an odd mode
impedance of 141.4 ohms. If the correct value is selected
for the uncoupled impedance (step one), the even mode
impedance will be the required 70.7 ohms. In this example,
the normalized odd mode impedance of 1.414 corresponds to an
S/D of 0.5613 and a normalized even mode impedance of 0.4174
(Fig. 14). This equates to an actual even mode impedance of
80.28 ohms.

Fourth, this process is repeated until the scaled version of Figure 14 produces an odd mode impedance of 141.4 ohms and an even mode impedance of 70.7 ohms for the same S/D. There is only one uncoupled impedance (Fig. 12) and one S/D (Fig. 14) that produce this relationship for each specific dielectric constant.

For Epsilam-10, the uncoupled impedance that matches this criteria is 92.3 ohms. The actual slot width is 0.0134 inches (0.3401 millimeters). The correct S/D is 0.4187. The normalized impedances are 0.3383 for the even mode and 1.5320 for the odd mode. The actual distance between the slots is 0.0105 inches (0.2659 millimeters).

The last step is to determine the slot wavelength from the 0.635 millimeter curve in Figure 13. The proper value for an S/D of 0.4187 is 0.5132 times the free space wavelength (Fig. 13). As discussed in the engineering approximations section, 97% of a quarter wavelength is used for the length of the coupled slots. This value is 0.1470 inches (3.7335 millimeters).

All of these calculations are done on two foot by three foot computer generated replicas of Figures 12, 13 and 14. The values are picked off of these large charts with calipers.

D. FIXTURE DESIGN AND ASSEMBLY

The fixture for the magic-tee is used to hold the fin-line monopulse system. A 1:1 scale drawing of the original version of the fixture is illustrated in Figure 15.

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The fixture is constructed from WR-90 copper waveguide. The two mirror image halves have 0.020 inch deep and 0.025 inch wide grooves in all three of the edges of the joining seam.

The two sections of dielectric are placed into these grooves. Then the microstrip launchers for ports three and four are set into place and soldered to the microstrip leads. Only one of the mirror image sections of dielectric is etched with microstrip. The other piece is void of metal on the inside.

After the two launchers are soldered in place, the fixture is bolted together. At this point, the launchers for ports one and two are attached.

The original design used OSM 2070-5029-02 launchers for all four ports. With this design, the launchers are held in place by the fixture when it is bolted together. The launcher is not screwed into the fixture. This arrangement does not provide a satisfactory electrical contact between the fixture and the launchers.

The fixture was modified to accept OSM 2052-1658-02 two hole flange mount jacks. The flange mounted jacks are attached to the fixture with small screws. This configuration provides good electrical continuity between the fixture and the connector.

E. FIN-LINE MAGIC-TEE DESIGN

   1. Slotted Side

      The length of the loop is 3/4 of a wavelength from the port three microstrip to the port four microstrip. The

width of the loop slot and the separation between the two halves of the loop are operator adjustable variables. The computer program (Appendix A) calculates the length of the portions of the loop, and adjusts the length of the straight section in the loop to keep the overall length equal to 3/4 of a wavelength.

The top and bottom lengths of the coupled slots are vastly different. The calculated quarter wave length is equal to the average of these two lengths.

All of the bends in the slots are constructed in the same manner. A line extending through a bend from corner to corner will bisect the angle of the bend. In this manner, the corners in a 90 degree bend are offset from each other by 45 degrees. In a 45 degree bend, the offset angle is 22.5 degrees. This arrangement insures that the slotwidth in the bend is at least as wide as the slots that lead into the bend.

All of these relationships are calculated by the computer program which draws the magic-tee.

The slots that lead to ports one and two are spaced wide apart for the first two magic-tees and closely together for the third magic-tee and for the fin-line monopulse system. In magic-tees one and two, the slots are 0.01 inches from the edge of the fixture.

In the third magic-tee and the fin-line monopulse system, the distance from the slots to the fixture groove is

twice as large as the distance between the slots. At 11.4-GHZ these distances are one half and one quarter of a wavelength, respectively. The half wavelength short and the quarter wave length open effectively seal the edges of the fins within the slot.

2. Microstrip Side

The widths of the microstrip lines are taken from Saad's microstrip impedance graph [Ref. 17: p. 117].

The transitions from microstrip to coaxial cable are unique. The metal foil on the slotline side of the dielectric acts as the ground plane for the microstrip. Near the connectors, this ground plane is parted in a "V" shape (Fig. 1). At the same time the width of the microstrip is flared at a lesser angle. The flare in the microstrip is adjusted to maintain a 50 ohm impedance at every point in the line.

The notch in the ground plane is flared at a 45 degree angle in each direction. The total angle of the notch is 90 degrees. The maximum width of the notch occurs at the inside edge of the fixture. At this point, the width of the notch exactly matches the diameter of the hole in the fixture wall, which exactly matches the outside diameter of the dielectric within the flanged launcher.

The magic-tee dielectric extends through the hole in the fixture and touches the launcher's dielectric. With this alignment, the inside wall of the shield on the coaxial

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cable transitions smoothly through the launcher to the edge of the ground plane notch.

This microstrip to coaxial cable transition minimizes physical discontinuities that could cause inductive or capacitive reactances [Ref. 18: p. 95]. This lack of reactance maximizes the bandwidth through the fixture wall.

F. MANUFACTURING PROCESS

A simple form of computer aided design is adapted for this project. A BASIC Computer Program, written on an HP 9845B computer (with an HP Graphics ROM), controls an HP 9872C Plotter. The plotter draws the outline of the magic-tee four times the actual size (4:1 scale). The outline is filled in by hand, using black marking pens. The completed drawing is photographically reduced.

The negatives, which are exactly 1:1 scale, are used as etching masks. The negative for the slot side and the negative for the microstrip side are taped to opposite sides of a scrap section of dielectric. The alignment between these two negatives is done on a light table. The phase difference at port four, between signals that originate at ports one and two can be altered by poor alignment.

The section of dielectric that is to be etched is placed between the two negatives during the etching process. The completed etching is hand cut and sanded to fit the fixture.

G. COMPUTER AIDED DESIGN

A simplified version of CAD (Computer Aided Design) is used to draw the outline of the magic-tee and fin-line monopulse system (Appendix A). This process starts by establishing an imaginary "X, Y" Cartesian Grid. Every significant point that defines the drawing is assigned two grid lines. One in the X-direction and one in the Y-direction. The grid line labels start at Xa in the horizontal direction and at Ya in the vertical direction. The origin of the grid is the point where the port three and port four microstrip lines would intersect if extended. The distances between the grid lines are defined by variables. There are numerous points on most of the grid lines, but no two points share the same horizontal and vertical grid line.

Each point which defines the drawing is numbered, starting at one, and assigned its own unique coordinates. If point one lies on the intersection of Xc and Yv, then X1 is set equal to Xc and Y1 is equated with Yv. This procedure is repeated for every point that defines the drawing.

The awkwardness of this approach is more than offset by two key advantages. First, the program that draws the magic-tee is very easy to write. The command MOVE X1, Y1 followed by DRAW X2, Y2 draws a line from point one to point two.

The second and perhaps most dramatic advantage occurs when the value of one of the variables is changed. Only one

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line in the program has to be changed to adjust the width of any slot. All of the grid lines that are affected by this variable are automatically adjusted as the program is executed.

Once the program is written, this approach is faster than roughly sketching new ideas by hand. Accurate scaled drawings showing numerous variations can be completed in a few minutes without the aid of a draftsman or expensive drafting computer.

The computer program adjusts the locations of the numbered points to compensate for half of the thickness of the line that the plotter makes. There is also a few variable fudge factors that the operator can assign to any point in the program to account for anomalies introduced by the plotter. Each finished drawing is measured under a microscope. The variable that adjusts for the width of the plotter's pen and the fudge factors are adjusted at this time.

The same computer program draws the magic-tee, the fin-line horn, and the fin-line monopulse system (Appendix A). The interactive program asks the operator a string of questions. These questions include: Which drawing is to be made?; Which side (slot or microstrip) is to be drawn?; and Is it a rough draft or smooth copy? The lines on the smooth copy take considerably longer to draw but are extremely accurate (Appendix A).

The program can add a scaled three inch ruler to the bottom of the drawing. This ruler is used to check the accuracy of the photo reduction. The ruler was only used on the first photography work order. The precision work accomplished by the Naval Postgraduate School Photo Lab is exemplary.

H. RESULTS

1. Fixture

As discussed above, the original fixture design does not provide sufficient electrical continuity between the launchers and the fixture. The flanged jacks, which attach to the modified fixture with screws, solve this problem.

The soldered joints in the fixture came apart on two occasions. Repair is extremely difficult due to the heat conducting properties of copper. The heat required to fix a seam is sufficient to loosen an adjacent joint.

Due to the flexibility of Epsilam-10, the fixture does not apply enough pressure to make good electrical contact with the port three and four microstrip. This problem is solved by soldering the launcher probes to the microstrip.

2. Magic-Tee Number One

The loop of this magic-tee tapers into the coupled slot region. The second magic-tee does not have this taper. Other than that, the first two magic-tees are identical.

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The first magic-tee was destroyed in testing. The soldered connection on port four ripped the microstrip off of the dielectric when the screws that attach the launcher to the fixture were tightened.

Preliminary reflection checks on a scalar analyzer were performed prior to this damage. The reflections for this magic-tee are slightly worse than the similar reflections in the second magic-tee.

3. Magic-Tee Number Two

The slots that lead to ports one and two in this magic-tee are 0.01 inches inside of the fixture wall. The actual scattering matrix for this device is illustrated in Figure 16. The reflections in this magic-tee are quite high.

The phase shifts are almost perfect, except for a slight phase error associated with port four. This is probably caused by improper alignment of the two negatives prior to the etching process.

4. Magic-Tee Number Three

The closeness of the port one and two slots to the fixture wall in the second magic-tee causes an apparent discontinuity where the fixture terminates. These slots were set well inside the fixture on the third magic-tee in hopes of eliminating this discontinuity.

The actual scattering matrix for the third magic-tee is shown in Figure 17. This magic-tee also has near perfect

phase shifts. However, the reflections are a little higher than they were for magic-tee number two.

The actual phase and magnitude from 8- to 12-GHZ for each of the 16 points in magic-tee number three's scattering matrix are illustrated in graphical form in Figures 18 to 33.

IV. FIN-LINE MONOPULSE SYSTEM

A. DESIGN

The fin-line monopulse system is a combination of fin-line magic-tee number three and two of the fin-line horns developed by Haq in Reference 5. The horns are driven from ports one and two of the magic-tee. The width of the slot in the horns is the same width as the matching slots in the magic-tee. The half wave length geometry of the fin-line horn and fin-line magic-tee are maintained in the fin-line monopulse system (Fig. 34). The monopulse system is drawn at a 2:1 scale. The HP-9872C Plotter will not accommodate a larger drawing.

All of the exposed edges of the dielectric, except the actual horn openings, are sealed with copper tape. This prevents energy from leaking out of the slots and destroying the antenna patterns.

The distance between the center of the two horns is 30.44 millimeters. At 10.3-GHZ, this distance is equal to 1.05 wavelengths in air and 3.34 wavelengths in Epsilam-10. The antennas use both air and dielectric to propagate the antenna pattern. Therefore, the effective element spacing of the fin-line monopulse system is between the limits of 1.05 and 3.34 wavelengths at 10.3-GHZ.

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The E-plane pattern for a single fin-line horn with a dielectric constant of 10.2 is illustrated in Figure 6. The associated H-plane pattern is shown in Figure 35. The extreme width of the H-plane pattern is due to the high dielectric constant of 10.2. Similar horns constructed with a dielectric constant of 2.54, have nearly symmetrical E- and H-plane patterns [Ref. 5]. The monopulse effect of this system is entirely in the E-plane. Therefore, the wide H-plane element pattern will produce a wide H-plane monopulse system pattern.

The gain of each antenna is obtained by comparing the amplitude of the antenna pattern with the pattern from a standard gain horn.

B. RESULTS

The actual E- and H-plane patterns for the monopulse system are shown in Figures 36 and 37 respectively. The gain of the sum pattern is eight dB. This is four dB above the element pattern. A three dB improvement is expected (Fig. 9).

The first nulls in the sum pattern (Fig. 36) are approximately 29 degrees left and right of boresight. The element spacing in the computer simulated sum pattern (Fig. 38) is adjusted until the first nulls occur at plus and minus 29 degrees. The corresponding simulated difference pattern is shown in Figure 39. The effective element spacing that produces this match is 1.60 wavelengths.

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The difference port microstrip was torn off of the dielectric during testing. This occurred after the patterns in Figures 36 and 37 were taken. The damage was repaired by soldering a small section of copper foil across the break.

Following this repair, the fin-line monopulse system was tested, in an effort to find the bottom of the difference null. Figure 40 shows the results of this test. The recorder gain is maximum, and the eight dB sum pattern is saturated against the top of the recorder. Yet, the bottom of the difference null is still not visible.

V. CONCLUSIONS AND RECOMMENDATIONS

A. FIXTURE

1. Conclusions

The original fixture design is unsatisfactory. It is too fragile and it does not provide good electrical continuity between the magic-tee and the launchers at ports three and four. The modified fixture, which has screwed on coaxial to microstrip launchers at ports three and four, is marginally effective.

The electrical transistions through the fixture wall at ports three and four do not appreciably restrict the bandwidth of the fin-line magic-tee (Figs. 10-33). Therefore, the holes in the fixture at ports three and four and the notches in the magic-tee's ground plane at ports three and four function reasonably well.

2. Recommendations

Testing should not be resumed until an improved fixture is constructed. Each half of the new fixture should be machined out of a solid piece of metal. All of the launchers which connect to the fixture should be attached with screws. A stiff or semi-rigid dielectric should be used to avoid the necessity of soldering the launchers to the microstrip lines. The microstrip to coaxial transition

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concept which is used in this thesis should be incorporated
in the new fixture design.

B. FIN-LINE MAGIC-TEE

  1. Conclusions

      The fin-line magic-tee worked better than expected
after only three iterations of the developmental process.
The phase shifts are very close to the theoretical
parameters. However, the dielectric material is too flexible
and the reflections at all four ports need to be reduced.

      There are four factors that could be causing the
high relections. First, as noted in the engineering
approximations (Chap. 3), the bilateral fin-line impedance
of the loop might be slightly above the desired 100 ohms.

      Second, the soldered connections on the microstrip
lines at ports three and four probably introduce significant
reflections.

      Third, the magic-tee is not completely symeteric.
The port three microstrip line couples with two slots while
the microstrip at port four couples with a single slot.

      Fourth, due to the different dielectric constants of
air and Epsilam-10, the distribution of the electric field
within the dielectric is not the same as the distribution of
the electric field in the air. Therefore, the reactive
interference in the air is not the same as the reactive
interference within the dielectric material. These unmatched
reactances probably contribute to the harmonic pattern of
reflections observed at all four ports (Figs. 18-33).

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2. Recommendations

Future magic-tees should be made out of a stiff or semi-rigid dielectric material. The combination of a stiffer material and a new fixture design should eliminate the need for soldered connections.

Two new magic-tees should be made with unilateral loop impedances of 190 ohms and 195 ohms. Tests of these tees will help identify the proper unilateral loop impedance and check the accuracy of the first engineering approximation.

Two sections of dielectric material which are totally void of metal foil can be added to the outsides of the existing two sections of dielectric. In this configuation, the slots will be totally surrounded by the same dielectric constant. This should improve the symmetry of the electric field distribution near the slots. thereby reducing the amount of reflected energy within the magic-tee.

C. FIN-LINE MONOPULSE SYSTEM

1. Conclusions

The performance of the fin-line monopulse system is very satisfactory. The null in the difference pattern is more than 40 dB below the peak of the sum pattern at 10.3-GHZ (Fig. 40). This deep unmeasurable null which exactly splits the main peak of the sum pattern (Fig. 40) is the essence of a good monopulse antenna. The minor problems

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caused by high reflections should be resolved when the fin-line magic-tee is improved.

The peak in the fin-line monopulse system's difference pattern is approximately nine dB lower than the peak in it's sum pattern (Fig. 36). This difference is approximately three dB more than the corresponding relationship in Skolnik's patterns [Ref. 7: pp. 17-23]. The fin-line comparator's weak difference pattern is probably the result of reflections caused by the large quanities of solder on the port four microstrip line. The shape of the fin-line monopulse system's sum and difference patterns and the exact centering of the difference null are almost identical to Skolnik's illustrations.

Skolnik [Ref. 7: pp. 10-28] shows that the second null in a theoretical sum pattern is not present on an actual monopulse pattern. This effect is roughly reproduced in this thesis (Figs. 36 and 38). The second and third peaks in the simulated sum pattern (Fig. 38) are at 40 and 72 degrees respectively. Corresponding peaks are evident at 50 and 77 degrees on the actual pattern (Fig. 36). The actual pattern is filled in between these two peaks just as Skolnik predicted.

As discussed in Chapter Four, the effective element spacing at 10.3-GHZ is 1.60 wavelengths. This is 152% of the element spacing in air and 48% of the element spacing in the dielectric. These percentages are probably a function of the

dielectric constant. Therefore, the effective element
spacing of the fin-line monopulse system can probably be
controlled by proper selection of the dielectric constant.

2. Recommendations

The fin-line monopulse system should be enlarged to
include two orthoganal difference ports. This will make it
useful as a target tracking antenna [Ref. 7: pp. 10-23].
The diamond configuration shown in Figure 4 and the square
design illustrated in Figure 5 can both be adapted to the
fin-line monopulse technique.

a. Diamond Fin-Line Monopulse System

A simple arrangement of four horns connected to
three fin-line magic-tees can be constructed using a
semi-rigid dielectric material. This configuration will
consist of two of the fin-line monopulse comparators
developed in this thesis placed side by side. The difference
ports of these two comparators will be the elevation and the
azmuith ports for the new three port system. The sum ports
will connect to the inputs of the third magic-tee. The third
magic-tee's sum port will be the system's sum port. The
difference port of the third magic-tee will be loaded.

The individual azimuth and elevation comparators
will be electrically orthogonal to each other if they are
each twisted 45 degrees in opposing directions. The twisting
should be confined to the area between the fixture and the
horns. In this configuration the two groups of antennas will

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be orthogonal to each other, yet all three magic-tees will be in the same plane.

b. Square Fin-Line Monopulse System

A dual plane fin-line monopulse system constructed out of semi-rigid dielectric material is illustrated in Figure 41. The dielectric material will have to be flexible enough to bend, yet rigid enough to provide good electrical contact between the microstrip lines and the launchers. This device will require careful engineering to properly align the sheets of dielectric material. Once these problems are solved, mass production should be easy.

D. COMPUTER AIDED DESIGN

1. Conclusions

The computer aided design portion of this thesis is useable, however it should be enlarged to include more of the design process.

2. Recommendations

Figures 12, 13, and 14 should be modified to include a wider range of dielectric constants and dielectric thicknesses. The data in the new versions of Figures 12 and 13 can probably be represented by simple algebraic equations [Ref. 10: pp. 226-228]. The new version of Figure 14 could be approximated by exponential equations. The four step trial and error process presented in Chapter Three and the design parameters for the fin-line horns [Ref. 5] can probably be written in the form of a short computer program.

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information should be computerized and added to the existing program (Appendix A).

The improved version of the computer aided design program could ask the operator; What gain/beam width combination is desired for the fin-line monopulse system? The program could then recommend the approriate dielectric constant and prompt the operator to place the paper on the plotting table. When the first drawing is finished, the program could prompt the operator to change the paper. Using this concept, made-to-order monopulse target tracking antennas could be inexpensively mass produced in a matter of days.

E. MEETING THESIS OBJECTIVES

Magic-tees number two and three and the monopulse system satisfy the first objective listed in the introduction. The sum and difference patterns (Figs. 36, 37, and 40) are better than expected for limited iterations of the design process. They meet the second objective. The interactive computer aided design program (Appendix A) satisfies the third objective. The recommendations presented in this chapter fulfill the final objective of this thesis.

APPENDIX A

COMPUTER PROGRAM LISTING
MSEE THESIS

```
18   !
20   !
30   !        FIN-LINE MAGIC TEE, FIN-LINE HORN AND
40   !            FIN-LINE MONOPULSE SYSTEM
50   !
60   !            BY LCDR JAMES W. ROWLEY
70   !
88   !                AUGUST 1985
90   !
108  !        NAVAL POSTGRADUATE SCHOOL, MONTEREY, CA.
110  !
128  !        (WRITTEN ON AN HP-9845B; WITH AN HP-9872C PLOTTER)
130  !
140  !
150  ! MISC. PROGRAMING INFO:
168  !
170  !
180  ! THIS PROGRAM IS WRITTEN IN "BASIC". THE FOLLOWING RULES APPLY:
198  !
208  !    BUILT IN COMMANDS: THE ENTIRE WORD IS IN UPPER CASE LETTERS.
210  !        DEFINITION OF NON-STANDARD BUILT IN COMMANDS:
220  !            ATN(): RETURNS THE ARC TANGENT OF THE VARIABLE IN THE BRACKETS.
230  !            MSCALE A,B:  PLOTTER IS SCALED IN MILLIMETERS. THE ORIGIN IS
248  !                "A" MILLIMETERS LEFT AND "B" MILLIMETERS UP FROM THE LOWER
258  !                LEFT HAND CORNER OF THE PLOTTING AREA. WHEN "MSCALE" IS
268  !                USED, ALL UNLABLED UNITS IN THE PROGRAM ARE IN MILLIMETERS.
278  !            PLOTTER IS "": COORDINATES THE COMPUTER AND THE PLOTTER.
208  !            DEG: TELLS THE COMPUTER THAT ALL ANGLES ARE IN DEGREES.
298  !
308  !    VARIABLES:  ONE UPPERCASE LETTER FOLLOWED BY NOTHING, OR FOLLOWED
310  !        BY A STRING OF LOWERCASE LETTERS AND/OR NUMBERS. A VARIABLE
320  !        CAN BE UP TO 15 CHARACTERS LONG. IF A VARIABLE IS ENDED IN A
330  !        DOLLAR SIGN (S), IT IS A STRING VARIABLE. DUE TO THE EXTREMELY
348  !        LARGE NUMBER OF VARIABLES IN THIS PROGRAM, THEY ARE DEFINED AND
358  !        ASSIGNED INITIAL VALUES AT THE SAME TIME. THIS PROGRAM DEFINES
368  !        KEY POINTS IN TERMS OF "X" AND "Y" REFERENCE PLANES. THESE PLANES
378  !        ARE DEFINED BY A COMBINATION OF A FEW KEY DIMENSIONS. THIS CHANGE
308  !        OF VARIABLES APPROACH WAS USED TO FACILITATE FUTURE DESIGN
398  !        MODIFICATIONS. A FOLLOW ON PROGRAMER CAN CHANGE ANY ONE OR MORE
408  !        OF THE KEY DIMENSIONS IN THIS PROGRAM AND THE ENTIRE DRAWING WILL
418  !        BE AUTOMATICALLY ADJUSTED. THIS IS A LIMITED APPLICATION OF CAD
428  !        (COMPUTER AIDED DESIGN).
438  !
440  ! STEP ONE: DETERMINE WHICH DRAWING THE OPERATOR DESIRES.
450  INPUT "MAGIC-TEE (TEE); HORN (HORN); OR MONOPULSE SYSTEM (MONO)?",Drawings
468  IF Drawings="TEE" THEN 500
478  IF Drawings="HORN" THEN 500
408  IF Drawings="MONO" THEN 500
498  GOTO 450 !ASK QUESTION UNTIL PROPER REPLY IS GIVEN.
500  INPUT "DRAW FIN-LINE OR MICROSTRIP SIDE? (F/M),CONT",Sides
518  IF Sides="F" THEN 550
528  IF Sides="M" THEN 550
538  GOTO 550!WRONG REPLY.
548  !
550  INPUT "DRAW RULER AND COMMENTS? (Y/N),CONT",Rulers
568  IF Rulers="Y" THEN 600!CHECK FOR PROPER REPLY.
578  IF Rulers="N" THEN 600
500  GOTO 550 !ASK QUESTION UNTIL PROPER REPLY IS GIVEN.
590  !
600  INPUT "QUICK LINE (1), OR PRECISION LINE (4)",Linenumber
618  IF Linenumber=1 THEN 668
628  IF Linenumber=4 THEN 648
638  GOTO 600!ASK UNTIL VALID REPLY GIVEN.
648  Segmentsize=.04     !DISTANCE BETWEEN THE DOTS, IF A DOTTED LINE IS USED.
658  !
668  ! STEP TWO: ADJUST THE GENERAL LAYOUT OF THE DRAWING.
```

```
670.  !
680   ! THIS SECTION CONTAINS REQUIRED COMMANDS FOR COM'..ER/PLOTTER
690   ! COORDINATION AND GENERAL COMMANDS AND VARIABLES THAT ARE .
700   ! USED TO ORGANIZE THE DRAWING.
710   PLOTTER IS "9872A"   !COORDINATES COMPUTER AND PLOTTER.
720   DEG                  !ALL ANGLES IN THE PROGRAM ARE IN DEGREES.
730   Title$="3"           !THIS WILL BE PRINTED ON THE DRAWING.
740   Flip=1               !-1 PUTS THE SUM NOTCH ON THE OPPOSITE SIDE.
750   Xcenter=82           !CENTER OF DRAWING ON PAGE IN mm FROM LOWER LEFT.
760   Ycenter=95           !CENTER OF DRAWING ON PAGE IN mm FROM LOWER LEFT.
770   MSCALE Xcenter,Ycenter   !POSITIONS THE DRAWING ON THE PAPER,
780   Scale=1              !THE ENTIRE DRAWING IS SCALED TO THIS FACTOR.
790   Mm=Scale             !DEFINES Mm=MILLIMETERS (DRAWN TO SCALE).
800   In=25.4*Scale        !DEFINES In=25.4 MILLIMETERS = INCH (DRAWN TO SCALE).
810   LINE TYPE Linenumber,Segmentsize      .  .
820                        ! IF Linenumber EQUALS "1", THE PLOTTER WILL DRAW
830                        ! A REGULAR LINE. IF Linenumber EQUALS "4", THE
840                        ! PLOTTER WILL DRAW A DOTTED LINE WITH OVERLAPPING
850                        ! DOTS. THE DOTTED LINE ELIMINATES PEN OSCILLATION
860                        ! ERRORS, AND PRODUCES A VERY HIGH QUALITY SOLID LINE.
870   !
880   !
890   ! STEP THREE: DEFINE THE DIMENSIONS OF THE DRAWING.
900   ! ;
910   ! "THE FOLLOWING VARIABLES DESCRIBE THE ENTIRE DRAWING. THE REST OF THE
920   ! PROGRAM USES THESE VALUES TO CONSTRUCT THE DRAWINGS.
930   ! ANY CHANGE IN THESE VARIABLES WILL RESULT IN ALL ASSOCIATED PARAMETERS
940   ! IN THE DRAWING BEING ADJUSTED AUTOMATICALLY.
950   !     VARIABLES ASSOCIATED WITH THE MAGIC TEE END IN A "t" SUBSCRIPT.
960   !     VARIABLES ASSOCIATED WITH THE SINGLE HORN END WITH AN "h" SUBSCRIPT.
970   !     VARIABLES ASSOCIATED WITH THE MONOPULSE SYSTEM END IN AN "mh"
980   !     SUBSCRIPT.
990   !
1000  !
1010  ! USER DEFINED VARIABLES:
1020  !
1030  Er=10.2                          !DIELECTRIC CONSTANT. Er IS ONLY USED
1040                                   ! TO CALCULATE THE TEM WAVELENGTH IN
1050                                   ! THE MICROSTRIP LINES.
1060  Lambdazero=26.306*Mm             !FREE SPACE WAVELENGTH AT 11.4 GHZ.
1070  Lambdad=Lambdazero/SQR(Er)       !TEM WAVELENGTH.
1080  Qwcoupledslots=.4998*Lambdazero/4 !QUARTER WAVELENGTH IN COUPLED SLOTS.
1090                                   ! .4998 IS 97% OF THE VALUE FOR A
1100                                   ! SINGLE SLOT THAT IS NOT COUPLED.
1110  Qwloop=.6360*Lambdazero/4        !QUARTER WAVELENGTH IN LOOP SLOTS.
1120  Qwmicrostrip=Lambdad/4           !QUARTER WAVELENGTH IN MICROSTRIP (TEM).
1130  Qwhornslot=.5216*Lambdazero/4    !QUARTER WAVELENGTH IN 108 OHM SLOTS.
1140  !
1150  ! THE FOLLOWING DIMENSIONS ARE EXACTLY MATCHED TO THE SIZE OF THE FIXTURE;
1160  At=.165*In           !WIDTH OF THE NOTCH IN THE COPPER FOIL THAT MATCHES THE
1170                       ! HOLE IN THE FIXTURE THAT CONNECTS THE DEVICE TO THE
1180                       ! COAXIAL LAUNCHER.
1190  Bt=.0230*In          !WIDTH OF THE MICROSTRIP TRANSITION AND OF THE WIDTH
1200                       ! OF THE CENTER CONDUCTOR IN THE COAXIAL ADAPTER.
1210  Ct=.83*In            !THICKNESS OF THE FIXTURE WALL THAT IS BEYOND THE
1220                       ! SUBSTRATE GROOVE.
1230  Dt=.22*In            !DISTANCE FROM CENTER OF DIFFERENCE PORT TO THE EDGE
1240                       ! OF THE DIELECTRIC THAT IS INSIDE OF THE FIXTURE.
1250  Et=.770*In           !LEFT EDGE OF THE SUBSTRATE TO THE CENTER OF SUM PORT.
1260                       ! THIS VALUE MAKES THE SUM AND DIFFERENCE PORTS
1270                       ! APPROXIMATELY THE SAME LENGTH.
1280  Ft=1.5*In            !CENTER OF SUM THE PORT TO THE RIGHT EDGE OF THE FIXTURE.
1290                       ! THIS ALLOWS SUFFICIENT ROOM TO MOUNT THE FIXTURE
1300                       ! WHILE TAKING ANTENNA PATTERNS.
1310  Grt=.5*In            !ALLOWS ROOM FOR LAUNCHERS ON RIGHT SIDE OF PORTS 1&2.
1320  Glt=.25*In           !ALLOWS ROOM FOR LAUNCHERS ON LEFT SIDE OF PORTS 1&2.
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1330  I
1340  I THE FOLLOWING DIMENSIONS DEFINE THE LAYOUT OF THE MAGIC TEE;
1350  I THREE DIFFERENT VALUES FOR Ht ARE GIVEN. PICK THE APPROPRIATE ONE.
1360  I Ht=1.3117*Mm  IWIDTH OF 160 OHM LOOP SLOT.
1370  I Ht=1.674*Mm  IWIDTH OF 180 OHM LOOP SLOT.
1380  Ht=2.056*Mm        IWIDTH OF 200 OHM LOOP SLOT.
1390  It=.340*Mm         IWIDTH OF THE COUPLED SLOTS.
1400  Jt=.43995*Mm       IWIDTH OF THE 100 OHM "HORN" SLOTS.
1410  Kt=.2086*Mm        IWIDTH OF MICROSTIP.
1420  Lt=.0781*In        ILENGTH OF THE TRANSITION IN THE SUM AND DIFFERENCE
1430     .              I MICROSTIP LINES.
1440  Mt=2.54*Mm         IDISTANCE BETWEEN THE SLOTS IN THE LOOP.
1450  Nt=Nt+Mt/2-(It+Ot/2) IMAKES ANGLE BETWEEN PTS 36,37 AND 38 135 DEGS.
1460  Ot=.266*Mm         IDISTANCE BETWEEN COUPLED SLOTS.
1470  Pt=1.5*Mm          IWIDTH OF THE ISOLATION SLOT.
1480  Qt=1.0*(Dt-Jt)     IMAKES THE PARALLEL SLOTS I/4 WAVELENGTH APART AND
1490                     I AND HALF A WAVELENGTH FROM FIXTURE WALLS AT 11.4 GHZ.
1500  Rt=.25*In          IDISTANCE BETWEEN FIXTURE AND FIRST CORNER IN SLOTS.
1510  St=.2*In           IDISTANCE FROM FIXTURE TO VERTICAL SLOTS.
1520  IF Rt+.1*In>St THEN St=Rt+.1*InIALLOWS ROOM FOR LOWER BEND IN THE
1530                                 I IN THE INPUT SLOTS (OUTSIDE OF FIXTURE).
1540  Tt=.3*In           ISIDE OF 45 DEGREE CUTOUT.
1550  Ut=0*In            ILENGTH OF VERTICAL SLOTS.
1560  Vt=15*Mm           INALF OF THE SEPARATION BETWEEN THE "HORN" SLOTS.
1570  I THE NEXT LINE INSURES ADEQUATE SEPARATION BETWEEN THE TIPS OF THE
1580  I MICROSTIP THAT EXTENDS FROM PORTS ONE AND TWO.
1590  IF Vt<.35*In THEN Vt=.35*In
1600  Wt=.5*In           IDISTANCE FROM "HORN" SLOTS TO EDGE OF DIELECTRIC.
1610                     I THIS ALLOWS SUFFICIENT ROOM TO MOUNT THE LAUNCHER.
1620  I
1630  I THE FOLLOWING LINES DEFINE THE NORN.
1640  Angleh=9.1 INALF OF THE NORN ANGLE.
1650  Hh=8               INh*Lambdad=LENGTH OF NORN FLARE (Bh).
1660  I
1670  I THE FOLLOWING DIMENSIONS ARE USED TO ADD THE MISC. DATA TO THE DRAWING.
1680  Sixteenth=1/16*In  IUSED TO BUILD RULER AT THE BOTTOM OF THE DRAWING.
1690  Opace=.07*In       ISPACES THE COMMENTS AND RULER AWAY FROM THE DRAWING.
1700  I
1710  I
1720  I THE FOLLOWING VARIABLES ARE COMBINATIONS OF THE USER DEFINED VARIABLES,
1730  I AND/OR SHORT VARIABLES THAT REPLACE LONGER MORE DESCRIPTIVE VARIABLES.
1740  I
1750  Bat=Qumicrostrip   ILENGTH OF MICROSTIP OVERLAPS.
1760  Bbt=Quhornslot     ILENGTH OF HORN SLOT OVERLAP.
1770  I
1780  I THERE ARE TWO VALUES FOR Bct AND Bdt GIVEN. ONE SET MAKES A TAPERED
1790  I LOOP, AND THE OTHER SET MAKES A NON-TAPERED LOOP AS INDICATED. PICK
1800  I THE DESIRED TYPE OF LOOP GEOMETRY.
1810  I
1820  I TAPERED LOOP:
1830  I Angle0 IS THE ANGLE USED TO DETERMINE Bct AND Bdt. IT IS ONE HALF OF THE
1840  I AVERAGE OF THE TWO ANGLES FORMED BY POINTS 20,21,22 AND 36,37,38. Bct
1850  I AND Bdt ARE USED TO MAKE A SMOOTH TRANSITION AROUND THE CORNERS IN
1860  I THE LOOP.
1870  I Angle0=((180-ATN((Ht+Mt/2-It-Ot/2)/Ht))/2+(180-ATN((Mt/2-Ot/2)/Ht))/2)/2
1880  I Bct=Nt/TAN(Angle0) ITHE OFFSET AT THE RIGHT SIDE OF THE TAPERED LOOP.
1890  I Bdt=It/TAN(Angle0) IOFFSET AT EDGE OF TAPERED LOOP.
1900  I
1910  I NON-TAPERED LOOP:
1920  Angle00=ATN(Ht/(Nt+Mt/2-(It+Ot/2)))  IFOR NON-TAPERED LOOP.
1930  Bct=Nt*TAN(45-Angle00/2) IFOR NON-TAPERED LOOP.
1940  Bdt=Ht/COS(Angle00)-It*TAN(Angle00) IFOR NON-TAPERED LOOP.
1950  I
1960  I
1970  Bet=Jt/COS(45)-It   IOFFSET AT LEFT EDGE OF COUPLED SLOTS.
1980  Bft=Qucoupledslots-Bet/2-Bdt/2  ILENGTH OF THE TOP OF THE COUPLED SLOTS.
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1990  Bgt=TAH(22.5)*Jt      !OFFSET FOR 45 DEGREE BENDS IN 100 OHM SLOTS.
2000  !
2010  !
2020  ! THE FOLLOWING CALCULATIONS MAKE THE CENTERLINE OF THE LOOP 3/4 OF A
2030  ! WAVELENGTH LONG.
2040  Arc=(Mt+Ht)*PI/4-Kt/2 !ARC LENGTH, EXCLUDING PORTION THAT OVERLAPS THE
2050                        ! MICROSTRIP.
2060  Slantpartofloop=SQR((Nt+Bct/2-Bdt/2)^2+((Mt+Ht-Ot-It)/2)^2) !LENGTH OF THE
2070                                                 ! SLANT PART OF
2080                                                 ! THE LOOP,
2090                                                 ! MEASURED ON
2100                                                 ! CENTERLINE.
2110  Flatpartofloop=3*Qwloop-Arc-Slantpartofloop-Bdt/2!ADJUSTABLE PART OF LOOP.
2120  Bht=Flatpartofloop+Bct/2  !THIS MAKES THE LOOP 3/4 OF A WAVELENGTH LONG.
2130  !
2140  Bit=Dt-Qt-It-Ot/2  !LENGTH OF 45 DEG SLOTS (IN THE "X" AND "Y" DIRECTIONS)
2150                     ! THAT ARE INSIDE OF THE FIXTURE.
2160  Bjt=Dt-Qt-Jt       ! DISTANCE FROM CENTERLINE TO INPUT SLOTS (INSIDE
2170                     ! OF THE FIXTURE).
2180  Bkt=St-Rt          !LENGTH OF LOWER 45 DEGREE SECTION (IN "X" AND "Y"
2190                     ! DIRECTION) THAT IS OUTSIDE OF THE FIXTURE.
2200  Blt=Vt-Bjt-Bkt-Ut  !LENGTH OF UPPER 45 DEGREE SECTION (IN THE "X" AND "Y"
2210                     ! DIRECTION) THAT ARE OUTSIDE OF THE FIXTURE.
2220  Bmt=2^.5*(Dt-Qt-Jt-Pt/2)-Ot/2  !DISTANCE FROM COUPLED SLOTS TO THE
2230                                  ! TIP OF THE ISOLATION SLOT (IN THE "X"
2240                                  ! DIRECTION).
2250                                  ! (IN THE "X" AND "Y" DIRECTIONS).
2260  Bnt=Pt/2           !LENGTH OF THE VEE AT THE END OF THE ISOLATION SLOT.
2270                     ! Bnt KEEPS THE SLANT PORTION OF  THE ISOLATION SLOT
2280                     ! AND THE HORIZONTAL PART OF THE ISOLATION SLOT
2290                     ! EQUADISTANCE FROM THE 100 OHM SLOTS (INSIDE OF THE
2300                     ! FIXTURE).
2310  !
2320  ! THE FOLLOWING ARE COMBINATIONAL HORN VARIABLES.
2330  IF Drawing$="HORN" THEN 2360 ! USED TO MATCH HORN TO FIXTURE.
2340  Ah=Qt            !WIDTH OF HORNS MATCHES FIXTURE FOR MAGIC-TEE.
2350  GOTO 2370
2360  Ah=Dt-Jt/2  !WIDTH OF HORN STRIP FOR SINGLE HORN.
2370  Bh=Nh*Lambdad  !LENGTH OF HORN FLARE FROM Xoh.
2380  Ch=Ah*SIN(Angleh)  !LATERAL OFFSET DISTANCE AT END OF HORN.
2390  Dh=Ah*COS(Angleh)  !VERTICAL OFFSET DISTANCE AT END OF HORN.
2400  Eh=Bh*SIN(Angleh)  !HALF OF HORN APERTURE.
2410  Fh=Bh*COS(Angleh)  !HORIZONTAL LENGTH OF HORN FROM Xoh.
2420  !
2430  !
2440  ! THE FOLLOWING LINE PLACES A CONDITION ON THE LENGTH OF Glt. THIS IS
2450  ! DONE TO ALLOW SUFFICENT ROOM BETWEEN THE MICROSTRIP CROSSING POINT,
2460  ! AND THE 45 DEGREE BEND IN THE SLOT TO THE LEFT OF THE MICROSTRIP.
2470  IF Glt<St+Jt+Blt+5*Mm THEN Glt=St+Jt+Blt+5*Mm
2480  !
2490  !
2500  ! STEP FOUR: DEFINE "X" AND "Y" PLANES IN TERMS OF VARIABLES LISTED ABOVE.
2510  !           ALL OF THE POINTS THAT WILL BE USED TO DEFINE THE DRAWING
2520  !           SHOULD BE ON THE INTERSECTION OF TWO OF THESE PLANES. OTHER
2530  !           POINTS CAN BE DRAWN TOO, BUT ONLY IF THEY ARE DEFINED WITHIN
2540  !           THE "DRAW" COMMAND (THIS SHOULD BE AVOIDED, AS IT WILL DESTROY
2550  !           THE "CAD" ASPECT OF THIS PROGRAM).
2560  !
2570  Xot=0            !SETS LOCATION OF DRAWING WITH RESPECT TO THE ORIGIN.
2580  Xat=Xot-Et-Ct    !LEFT EDGE OF DIFFERENCE PORT.
2590  Xbt=Xot-Et       !LEFT EDGE OF THE DIELECTRIC WITHIN THE FIXTURE.
2600  Xct=Xbt+Lt       !TIP OF THE DIFFERENCE NOTCH.
2610  Xdt=Xot-Kt/2-Nt-Bht !CENTER OF THE LOOP HALF CIRCLE.
2620  Xet=Xdt-Mt/2+Bat  !EDGE OF DIFFERENCE MICROSTRIP.
2630  Xft=Xdt+Bht-Bct  !LOWER LEFT LOOP CORNER.
2640  Xgt=Xdt+Bht      !UPPER LEFT LOOP CORNER.
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2650  Xht=Xot-At/2            !LEFT EDGE OF THE TAB OF DIELECTRIC ON THE SUM PORT.
2660  Xit=Xot-Bt/2            !LEFT EDGE OF SUM MICROSTRIP TRANSITION.
2670  Xjt=Yot-Kt/2-Bdt        !LOWER RIGHT LOOP CORNER.
2680  Xkt=Xot-Kt/2            !LEFT EDGE OF SUM MICROSTRIP.
2690  Xit=Xot+Kt/2            !RIGHT EDGE OF SUM MICROSTRIP.
2700  Xmt=Xot+Bt/2            !RIGHT EDGE OF SUM MICROSTRIP TRANSITION.
2710  Xnt=Xot+At/2            !RIGHT EDGE OF THE TAB OF DIELECTRIC ON THE SUM PORT.
2720  Xpt=Xot-Kt/2+Bft        !UPPER RIGHT EDGE OF COUPLED SLOTS.
2730  Xqt=Xpt+Bet            !LOWER RIGHT EDGE OF COUPLED SLOTS.
2740  Xrt=Xpt+Bit            !UPPER LEFT EDGE OF "INPUT" SLOTS.
2750  Xst=Xrt+Bgt            !LOWER LEFT EDGE OF "INPUT" SLOTS.
2760  Xtt=Xqt+Bmt            !LEFT TIP OF THE ISOLATION SLOT.
2770  Xut=Xtt+Bnt            !START OF ISOLATION SLOT TIP.
2780  Xvt=Xot+Ft            !RIGHT EDGE OF THE FIXTURE.
2790  Xwt=Xvt+Rt            !UPPER RIGHT EDGE OF "INPUT" SLOT.
2800  Xxt=Xwt+Bgt            !LOWER RIGHT EDGE OF "INPUT" SLOT.
2810  Xyt=Xvt+Tt            !TOP OF 45 DEGREE CUTOUT.
2820  Xzt=Xvt+St            !LEFT SIDE OF VERTICAL SLOT.
2830  Xaat=Xzt+Jt            !RIGHT SIDE OF VERTICAL SLOT.
2840  Xabt=Xaat+Bit-Bgt      !LEFT EDGE OF UPPER SLOT CORNER.
2850  Xact=Xaat+Bit          !RIGHT EDGE OF UPPER SLOT CORNER.
2860  Xadt=Xvt+Git          !LEFT EDGE OF INPUT NOTCH.
2870  Xaet=Xadt+At/2-Bt/2    !LEFT EDGE OF INPUT TRANSITION.
2880  Xaft=Xadt+At/2-Kt/2    !LEFT EDGE OF INPUT MICROSTRIP.
2890  Xagt=Xadt+At/2          !CENTER OF INPUT NOTCH.
2900  Xaht=Xagt+Kt/2          !RIGHT EDGE OF INPUT MICROSTRIP.
2910  Xait=Xagt+Bt/2          !RIGHT EDGE OF INPUT TRANSITION.
2920  Xajt=Xadt+At            !RIGHT EDGE OF INPUT NOTCH.
2930  Xakt=Xaht+Bbt          !RIGHT END OF SLOTS.
2940  Xalt=Xajt+Grt          !RIGHT EDGE OF DIELECTRIC.
2950  !
2960  !
2970  Xoh=Xvt+Ch-Jt/(2*TAN(Angleh))  !APEX OF HORN OPENING ANGLE
2980                                 ! REFERENCED TO EDGE OF FIXTURE.
2990  Xah=Xoh+Fh            !OUTER EDGE OF HORN OPENING.
3000  Xbh=Xah-Ch            !UPPER OUTER EDGE OF HORN.
3010  Xch=Xoh+Jt/(2*TAN(Angleh))    !HORN MOUTH.
3020  Xdh=Xch-Ch            !HORN/FIXTURE JOINT.
3030  Xeh=Xkt-Bbt          !END OF HORN SLOT.
3040  !
3050  !
3060  Xwmh=Xvt-TAN(22.5)*Lambdad/2  !USED TO MOVE X34t LEFT TO X34mh.
3070  Xxmh=Xxt+Lambdad*(SQR(.5)-.125)  !MOVES X25t RIGHT TO X25mh.
3080  Xzmh=Xzt-Lambdad/2            !USED TO MOVE X32t/X33t LEFT TO X32mh/X33mh.
3090  Xaamh=Xaat+Lambdad/2          !USED TO MOVE X26t/X27t RIGHT TO X26mh/X27mh.
3100  Xabmh=Xabt-TAN(22.5)*(Lambdad/2) !USED TO MOVE X31t LEFT TO X31mh.
3110  Xacmh=Xact+TAN(22.5)*(Lambdad/2) !USED TO MOVE X20t RIGHT TO X20mh.
3120  !
3130  !
3140  Yot=0
3150  Yat=Yot+Vt+Jt+Wt      !TOP OF THE DIELECTRIC.
3160  Ybt=Yat-(Ct+Lt)        !TIP OF INPUT NOTCH.
3170  Yct=Yat-Tt            !BOTTOM OF 45 DEGREE CUTOUT.
3180  Ydt=Yot+Vt+Jt        !UPPER EDGE OF THE HORN SLOT.
3190  Yet=Yot+Vt            !LOWER EDGE OF THE HORN SLOT.
3200  Yft=Yet-Bit+Bgt      !UPPER EDGE OF THE UPPER SLOT CORNER.
3210  Ygt=Yet-Bit          !LOWER EDGE OF THE UPPER SLOT CORNER.
3220  Yht=Yot+Dt+Ct        !TOP EDGE OF SUM PORT DIELECTRIC.
3230  Yit=Yet-Bat          !END OF PORT ONE MICROSTRIP.
3240  Yjt=Yot+Dt            !TOP OF DIELECTRIC IN THE FIXTURE.
3250  Ykt=Yot+Bjt+Bkt+Bgt  !UPPER EDGE OF LOWER SLOT CORNER.
3260  Ylt=Ykt-Bgt          !LOWER EDGE OF LOWER SLOT CORNER.
3270  Ymt=Yjt-Lt          !TIP OF SUM NOTCH.
3280  Ynt=Yot+At/2          !TOP EDGE OF THE DIFFERENCE PORT DIELECTRIC TAB.
3290  Ypt=Yot+Mt/2+Ht      !TOP EDGE OF THE SLOT IN THE LOOP.
3300  Yqt=Yjt-Ot          !TOP EDGE OT THE INPUT SLOT.
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3310  Yrt=Yjt-Qt-Jt           !LOWER EDGE OF THE INPUT SLOT.
3320  Yst=Yot+Mt/2            !LOWER EDGE OF THE LOOP SLOT.
3330  Ytt=Yot+Pt/2            !TOP EDGE OF THE ISOLATION SLOT.
3340  Yut=Yot+Ot/2+It         !UPPER EDGE OF THE COUPLED SLOT.
3350  Yvt=Yot+Bt/2            !UPPER EDGE OF THE DIFFERENCE PORT CENTER CONDUCTOR.
3360  Ywt=Yot+Ot/2            !LOWER EDGE OF THE COUPLED SLOT.
3370  Yxt=Yot+Kt/2            !UPPER EDGE OF THE DIFFERENCE MICROSTRIP.
3380  Yyt=Yot-Kt/2            !LOWER EDGE OF THE DIFFERENCE MICROSTRIP.
3390  Yzt=Yot-Bt/2            !LOWER EDGE OF DIFFERENCE PORT CENTER CONDUCTOR.
3400  Yaat=Yot-Ot/2-It-Bat    !END OF SUM PORT MICROSTRIP.
3410  Yabt=-Yit               !END OF PORT 2 MICROSTRIP.
3420  Yact=-Ybt               !TIP OF LOWER INPUT NOTCH (USED TO DRAW MICROSTRIP).
3430  Yadt=-Yat               !BOTTOM OF LOWER INPUT PORT.
3440  Yaet=Yadt-Space         !TOP OF THE RULER.
3450  Yaft=Yaet-2*Sixteenth   !LOWER TIP OF A RULER MARK.
3460  Yagt=Yaft-Sixteenth     !LOWER TIP OF A RULER MARK.
3470  Yaht=Yagt-Sixteenth     !LOWER TIP OF A RULER MARK.
3480  Yalt=Yaht-Sixteenth     !LOWER TIP OF A RULER MARK.
3490  Yajt=Yalt-Sixteenth     !LOWER TIP OF A RULER MARK.
3500  Yakt=Yaet-.57*In        !FIRST LINE OF WORDS UNDER THE RULER.
3510  Yalt=Yaet-.72*In        !SECOND LINE OF WORDS UNDER THE RULER.
3520  !
3530  !
3540  Yoh=Yot                 !CENTER OF HORN.
3550  Yah=Yot+Jt/2            !TOP OF HORN SLOT.
3560  Ybh=Yot+Eh              !END OF HORN OPENING.
3570  IF Drawing$="MONO" THEN 3600
3580  Ych=Yjt                 !PLACE WHERE HORN JOINS THE FIXTURE.
3590  GOTO 3610
3600  Ych=Yot+Qt+Jt/2         !THIS KEEPS THE COPPER Lambdad/2 ABOVE THE SLOT.
3610  Ydh=Ybh+Dh              !TOP OF HORN.
3620  !
3630  !
3640  Ydmh=Ydt+Lambdad/2                      !USED TO MOVE X31t UP TO X31mh.
3650  Yemh=Yet-Lambdad/2                      !USED TO MOVE X28t DOWN TO X28mh.
3660  Yfmh=Yft+TAN(22.5)*(Lambdad/2)  !USED TO MOVE X32t UP TO X32mh.
3670  Ygmh=Ygt-TAN(22.5)*(Lambdad/2)  !USED TO MOVE X27t DOWN TO X26mh:
3680  Ykmh=Ykt+TAN(22.5)*(Lambdad/2)  !USED TO MOVE X33T UP TO X33mh.
3690  Ylmh=Ylt-TAN(22.5)*(Lambdad/2)  !USED TO MOVE X26t DOWN TO X26mh.
3700  Yqmh=Yqt+Lambdad/2                      !USED TO MOVE X34t UP TO X34mh.
3710  Yrmh=Yrt-Lambdad/2                      !USED TO MOVE X25t DOWN TO X25mh.
3720  ! STEP FIVE: CALCULATION OF CORRECTIONS THAT WILL BE USED TO REMOVE
3730  !               ERRORS FROM THE DRAWING THAT ARE CAUSED BY THE PLOTTER.
3740  !
3750  :     THERE IS A SLIGHT BIT OF LOSENESS IN THE PEN CABLE ON THE HP 9072C
3760  !     PLOTTER. THE AMOUNT OF ERROR THAT THIS INDUCES IS WITHIN THE PLOTTERS
3770  !     SPECIFICATIONS AND DOES NOT CAUSE ANY PROBLEM IN MOST PLOTTER
3780  !     APPLICATIONS. HOWEVER, DUE TO THE EXTREMELY SMALL DIMENSIONS IN THE
3790  !     AREA NEAR THE COUPLED SLOTS, THESE ERRORS MUST BE ACCOUNTED FOR IN
3800  !     THIS PROGRAM. THESE CORRECTIONS CAME FROM EMPIRICAL TESTING AND
3810  !     WILL HAVE TO BE ADJUSTED TO MATCH THE PARTICULAR PLOTTER THAT IS
3820  !     BEING USED. THE In/Mm SCALING IS INTENTIONALLY OMITTED TO KEEP THE
3830  !     THE CORRECTIONS TO THE SCALE OF THE ACTUAL DRAWING AND PEN.
3840  Plottercorrect1=.07
3850  Plottercorrect2=8
3860  Plottercorrect3=8
3870  Plottercorrect4=0
3880  Pen=.52                 !THIS IS THE PEN THICKNESS, AND IT IS CORRECTED
3890                          ! FOR AUTOMATICALLY IN THE DRAWING.
3900  !
3910  !
3920  ! STEP SIX: ENTER TWO MAJOR LOOPS THAT MAKE THE DESIRED DRAWING:
3930  !
3940  !     THE FIRST LOOP (THE "FILL" LOOP) IS USED TO FILL IN THE BLACK
3950  !     PORTIONS OF THE DRAWING. THIS IS DONE BY MOVING EACH POINT OF THE
3960  !     FIGURE IN 68% OF THE PEN THICKNESS, AND THEN DRAWING IT OVER AGAIN.
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3970  |
3980  |    THE SECOND LOOP (THE "Im" LOOP) IS USED TO DRAW THE FIN-LINE PATTERN.
3990  |    THE BOTTOM HALF IS ESSENTIALLY A MIRROR IMAGE OF THE TOP. THEREFORE,
4000  |    ONLY THE TOP HALF IS DEFINED. THE BOTTOM HALF IS DRAWN BY MULTIPLYING
4010  |    THE "Y" COORDINATES BY MINUS ONE. ANY NON-SYMETRIC AREAS (SUCH AS
4020  |    THE SUM PORT) ARE TAKEN CARE OF WITH "IF" STATEMENTS, THAT CHECK TO
4030  |    SEE IF THE TOP OR BOTTOM HALF IS BEING DRAWN.
4040  |
4050  |
4060  FOR Fill=1 TO 1 STEP 2              !START OF THE "Fill" LOOP.
4070  FOR Image=0 TO 2 STEP 2             !START OF THE "Im" LOOP.
4080  Im=1-Image                         !Im=1 DRAWS THE TOP, AND Im=-1 DRAWS THE
4090                                      ! BOTTOM OF THE SLOTLINE SIDE.
4100  IF Fill=3 THEN Segmentsize=Segmentsize*3
4110  IF Fill>1 THEN GOTO 4180           !SKIPS TO THE "Fill" ROUTINE.
4120  IF Im=1 THEN Penn=Pen*Fill           !PLOTTER CORRECTION FOR THE
4130                                      ! TOP HALF OF SLOTLINE SIDE.
4140  IF Im=-1 THEN Penn=Pen*Fill+Plottercorrect1 !PLOTTER CORRECTION FOR THE
4150                                      ! BOTTOM HALF OF THE SLOTLINE
4160                                      ! SIDE.
4170  GOTO 4230                          !SKIPS "Fill" ROUTINE ON FIRST PASS.
4180  IF Im=1 THEN Penn=Pen*Fill*.60     !"Fill" ROUTINE. MOVES PEN IN 60% OF
4190                                      ! IT'S WIDTH AND THEN DRAWS THE
4200                                      ! FIGURE AGAIN.
4210  |
4220  |:
4230  | STEP SEVEN: CALCULATION OF CORRECTIONS THAT WILL BE USED TO REMOVE
4240  |             ERRORS FROM THE DRAWING THAT ARE CAUSED BY THE PEN.
4250  |
4260  |    THE FOLLOWING CALCULATIONS ARE USED TO ELIMINATE ANY INACCURACY IN
4270  |    THE DRAWING DUE TO THE WIDTH OF THE PEN THAT IS BEING USED.
4280  |    THESE ADJUSTMENTS WILL AUTOMATICALLY MOVE THE PEN THE APPROPRIATE
4290  |    AMOUNT AND DIRECTION TO ACCOUNT FOR HALF OF THE THICKNESS OF THE PEN.
4300  |
4310  Angle1=(180-ATN((Bt-Kt)/(2*Lt)))/2      !HALF OF THE ANGLE FORMED BY
4320                                           ! POINTS 102,103 AND 104.
4330  Penoffset1=Penn/(2*TAN(Angle1))          !CORRECTS POINTS 102, 103, 106 AND
4340                                           ! 107 IN THE "X" DIRECTION AND
4350                                           ! POINTS 111, 112, 115 AND 116 IN
4360                                           ! "Y" DIRECTION.
4370  |
4380  Angle2=ATN(At/(2*Lt))                    !HALF OF THE ANGLE FORMED BY
4390                                           ! POINTS 2,1 AND THE IMAGE OF 1.
4400  Penoffset2=Penn/(2*SIN(Angle2))          !CORRECTS POINT 1 IN THE "X"
4410                                           ! DIRECTION AND POINT 5 IN THE
4420                                           ! "Y" DIRECTION.
4430  |
4440  Angle3=(180-ATN((Bt-Kt)/(2*(Lt+Ct))))/2 !HALF OF THE ANGLE FORMED BY
4450                                           ! POINTS 118,119 AND 120.
4460  Penoffset3=Penn/(2*TAN(Angle3))          !CORRECTS POINTS 119,122,125 AND
4470                                           ! 120 IN THE "Y" DIRECTION.
4480  |
4490  Angle3a=(90-(180-2*Angle3))/2            ! HALF OF THE ANGLE FORMED BY
4500                                           ! POINTS 117,118 AND 119.
4510  Penoffset3a=Penn/(2*TAN(Angle3a))        !CORRECTS POINTS 117,118, 123 AND
4520                                           ! 124 IN THE "X" DIRECTION.
4530  |
4540  Angle4=ATN(At/(2*(Lt+Ct)))               ! HALF OF THE ANGLE FORMED BY
4550                                           ! POINTS 11, 12 AND 13.
4560  Penoffset4=Penn/(2*SIN(Angle4))          !CORRECTS POINT 12 IN THE "Y"
4570                                           ! DIRECTION.
4580  |
4590  Angle5=(180-ATN(2*Lt/At))/2              !HALF OF THE ANGLE FORMED BY
4600                                           ! POINTS 1,2 AND 3.
4610  Penoffset5=Penn/(2*TAN(Angle5))          !CORRECTS POINT 2 IN THE "Y"
4620                                           ! DIRECTION AND POINTS 4 AND 5
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4630                                          I IN THE "X" DIRECTION.
4640    I
4650    Angle6=(180-ATN(2*(Lt+Ct)/At))/2     IHALF OF THE ANGLE FORMED BY POINTS
4660                                          I 10,11 AND 12.
4670    Penoffset6=Penn/(2*TAN(Angle6))      ICORRECTS POINTS 11 AND 13 IN THE
4680                                          I "X" DIRECTION.
4690    I
4700    Penoffset7=Pen/(2*TAN(67.5))         ICORRECTS POINTS 23, 24, 25, 28
4710                                          I 31, 34, 35 AND 36 IN THE "X"
4720                                          I DIRECTION, AND POINTS 26, 27, 32
4730                                          I AND 33 IN THE "Y" DIRECTION.
4740    Penoffset7a=Penn/(2*COS(45))         ICORRECTS POINT 18 IN THE "X"
4750                                          I DIRECTION.
4760    I
4770    Angle8=(180-ATN((Mt/2+Ht-Ot/2-It)/Ht))/2 IHALF OF THE ANGLE FORMED BY
4780                                          I POINTS 37, 38 AND 39.
4790    Penoffset8=Penn/(2*TAN(Angle8))      ICORRECTS POINTS 37 AND 38 IN
4800                                          I THE "X" DIRECTION.
4810    I
4820    Angle9=(180-ATN((Mt/2-Ot/2)/(Ht+Bet-Bdt)))/2 IHALF OF THE ANGLE FORMED
4830                                          I BY POINTS 21, 22 AND 23.
4840    Penoffset9=Penn/(2*TAN(Angle9))      ICORRECTS POINTS 21 AND 22 IN
4850                                          I THE "X" DIRECTION.
4860    I
4870    Angleh1=45-Angleh  IUSED TO DETERMINE OFFSETS AT END OF HORN.
4880    Angleh2=(180-Angleh)/2  IUSED TO DETERMINE OFFSETS AT MOUTH OF HORN.
4890    I
4900    Penoffseth1=SQR(2)*Penn/2*SIN(Angleh1) ICORRECTS 2h IN "X" DIRECTION AND
4910                                          I 3h IN "Y" DIRECTION.
4920    Penoffseth2=SQR(2)*Penn/2*COS(Angleh1) ICORRECTS 3h IN "X" DIRECTION AND
4930                                          I 2h IN THE "Y" DIRECTION.
4940    Penoffseth3=Penn/2/TAN(Angleh2)      ICORRECTS 1h AND 4h IN THE "X"
4950                                          I DIRECTION.
4960    I
4970    I
4980    I STEP EIGHT: DEFINE ALL NUMBERED POINTS ON THE DRAWING IN TERMS OF THE
4990    I            "X" AND "Y" REFERENCE PLANES LISTED ABOVE.
5000    I
5010    I    THE PEN OFFSET CALCULATIONS ARE ENTERED HERE. THIS KEEPS THE
5020    I    REFERENCE PLANES FREE OF PEN CORRECTIONS.
5030    I
5040    I THE FOLLOWING POINTS DEFINE THE FIN-LINE SIDE.
5050    I
5060    X1t=Xct+Penoffset2
5070    X2t=Xbt+Penn/2          ITHIS IS FOR THE FIN-LINE SIDE OF POINT 2.
5080    X2at=Xbt-Penn/2         ITHIS IS FOR THE DIELECTRIC TAB SIDE OF POINT 2.
5090    X3t=Xbt+Penn/2
5100    X4t=Xht-Penoffset5      ITHIS IS FOR THE FIN-LINE SIDE OF POINT 4.
5110    X4at=Xht-Penn/2         ITHIS IS FOR THE DIELECTRIC TAB SIDE OF POINT 4.
5120    X5t=Xet
5130    X6t=Xnt+Penoffset5      ITHIS IS FOR THE FIN-LINE SIDE OF POINT 6.
5140    X6at=Xnt+Penn/2         ITHIS IS FOR THE DIELECTRIC TAB SIDE OF POINT 6.
5150    X7t=Xvt+Penn/2
5160    X8t=Xvt+Penn/2
5170    X9t=Xvt+Penn/2
5180    X10t=Xyt+Penoffset7
5190    X11t=Xadt-Penoffset6
5200    X12t=Xagt
5210    X13t=Xajt+Penoffset6
5220    X14t=Xalt-Penn/2
5230    X15t=Xalt-Penn/2
5240    X16t=Xalt-Penn/2
5250    X17t=Xut-Penoffset7
5260    X18t=Xtt-Penoffset7a
5270    X19t=Xdt   ITHE PEN OFFSET FOR THE ARC IS DONE WHILE MAKING THE DRAWING.
5280    X20t=Xdt
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5290    X21ι=Xfι-Penoffset9.
5300    X22ι=Xjι-Penoffset5
5310    X23ι=Xqι+Penoffset7
5320    X24ι=Xsι+Penoffset7
5330    X25ι=Xxι+Penoffset7
5340    X26ι=Xaaι+Penn/2
5350    X27ι=Xaaι+Penn/2
5360    X28ι=Xacι+Penoffset7
5370    X29ι=Xakι+Penn/2
5380    X30ι=Xakι+Penn/2
5390    X31ι=Xabι-Penoffset7
5400    X32ι=Xzι-Penn/2
5410    X33ι=Xzι-Penn/2
5420    X34ι=Xuι-Penoffset7
5430    X35ι=Xrι-Penoffset7
5440    X36ι=Xpι-Penoffset7
5450    X37ι=Xkι+Penoffset8
5460    X38ι=Xgι+Penoffset8
5470    X39ι=Xdι
5480    X40ι=Xaι-Penn/2
5490    X41ι=Xaι-Penn/2
5500    X42ι=Xhι-Penn/2
5510    X43ι=Xnι+Penn/2
5520    !
5530    !
5540    !. THE FOLLOWING POINTS DEFINE THE MICROSTRIP SIDE.
5550    !
5560    !
5570    X101ι=Xaι+Penn/2
5580    X102ι=Xbι-Penoffset1
5590    X103ι=Xcι-Penoffset1
5600    X104ι=Xeι-Penn/2
5610    X105ι=Xeι-Penn/2
5620    X106ι=Xcι-Penoffset1
5630    X107ι=Xbι-Penoffset1
5640    X100ι=Xaι+Penn/2
5650    X109ι=Xlι+Penn/2
5660    X110ι=Xaι-Penn/2
5670    X111ι=Xaι-Penn/2
5680    X112ι=Xlι-Penn/2
5690    X113ι=Xlι-Penn/2
5700    X114ι=Xkι+Penn/2
5710    X115ι=Xkι+Penn/2
5720    X116ι=Xlι+Penn/2
5730    X117ι=Xaeι+Penoffset3a
5740    X118ι=Xalι-Penoffset3a
5750    X119ι=Xahι-Penn/2
5760    X120ι=Xahι-Penn/2
5770    X121ι=Xafι+Penn/2
5780    X122ι=Xafι+Penn/2
5790    X123ι=Xalι-Penoffset3a
5800    X124ι=Xaeι+Penoffset3a
5810    X125ι=Xafι+Penn/2
5820    X126ι=Xafι+Penn/2
5830    X127ι=Xahι-Penn/2
5840    X128ι=Xahι-Penn/2
5850    !
5860    ! THE FOLLOWING POINTS DEFINE THE HORN.
5870    X1h=Xdh+Penoffseth3.
5880    X2h=Xbh-Penoffseth1
5890    X3h=Xah-Penoffseth2
5900    X4h=Xch-Penoffseth3
5910    X5h=Xeh-Penn/2
5920    X6h=Xeh-Penn/2
5930    !
5940    ! THE FOLLOWING POINTS ARE USED TO MERGE THE MAGIC-TEE AND HORN INTO THE
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5950  ! MONOPULSE SYSTEM.
5960  X25mh=Xxmh-Penoffset7
5970  X26mh=Xaamh-Penn/2
5980  X27mh=Xaamh-Penn/2
5990  X28mh=Xacmh-Penoffset7
6000  X31mh=Xabmh+Penoffset7
6010  X32mh=Xzmh+Penn/2
6020  X33mh=Xzmh+Penn/2
6030  X34mh=Xumh+Penoffset7
6040  !
6050  ! THE FOLLOWING POINTS ARE USED TO DRAW THE RULER, BULL'S EYES AND
6060  ! COMMENTS.
6070  X150t=Xat              !LEFT SIDE OF THE RULER.
6080  Xbullleft=Xat+.25*In   !CENTER OF LEFT BULL'S EYE.
6090  Xbullright=Xalt-.25*In  !CENTER OF RIGHT BULL'S EYE.
6100  !
6110  !
6120  ! THE FOLLOWING LINES DEFINE THE "Y" COORDINATES OF THE FIN-LINE SIDE.
6130  !
6140  Y1t=Yot
6150  Y2t=Ynt+Penoffset5
6160  Y2at=Ynt+Penn/2
6170  Y3t=Yjt-Penn/2
6180  Y4t=Yjt-Penn/2
6190  Y4at=Yjt+Penn/2
6200  Y5t=Ynt-Penoffset2
6210  Y6t=Yjt-Penn/2
6220  Y6at=Yjt+Penn/2
6230  Y7t=Yjt-Penn/2
6240  Y8t=Yct-Penn/2
6250  Y9t=Yat-Penn/2
6260  Y10t=Yat-Penn/2
6270  Y11t=Yat-Penn/2
6280  Y12t=Ybt-Penoffset4
6290  Y13t=Yat-Penn/2
6300  Y14t=Yat-Penn/2
6310  Y15t=Yt1+Penn/2
6320  Y16t=Yot
6330  Y17t=Yt1+Penn/2
6340  Y18t=Yot
6350  Y19t=Yot
6360  Y20t=Yst-Penn/2
6370  Y21t=Yst-Penn/2
6380  Y22t=Yut-Penn/2
6390  Y23t=Yut-Penn/2
6400  Y24t=Yrt-Penn/2
6410  Y25t=Yrt-Penn/2
6420  Y26t=Ylt-Penoffset7
6430  Y27t=Ygt-Penoffset7
6440  Y28t=Yet-Penn/2
6450  Y29t=Yet-Penn/2
6460  Y30t=Ydt+Penn/2
6470  Y31t=Ydt+Penn/2
6480  Y32t=Yft+Penoffset7
6490  Y33t=Ykt+Penoffset7
6500  Y34t=Yqt+Penn/2
6510  Y35t=Yqt+Penn/2
6520  Y36t=Yut+Penn/2
6530  Y37t=Yut+Penn/2
6540  Y38t=Ypt+Penn/2
6550  Y39t=Ypt+Penn/2
6560  Y40t=Yot
6570  Y41t=Ynt+Penn/2
6580  Y42t=Yht+Penn/2
6590  Y43t=Yht+Penn/2
6600  !
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6610   !
6620   ! THE FOLLOWING POINTS DEFINE THE MICROSTRIP SIDE.
6630   Y101t=Yvt-Penn/2
6640   Y102t=Yvt-Penn/2
6650   Y103t=Yxt-Penn/2
6660   Y104t=Yxt-Penn/2
6670   Y105t=Yyt+Penn/2
6680   Y106t=Yyt+Penn/2
6690   Y107t=Yzt+Penn/2
6700   Y108t=Yzt+Penn/2
6710   Y109t=Yht-Penn/2
6720   Y110t=Yht-Penn/2
6730   Y111t=Yjt+Penoffset1
6740   Y112t=Ymt+Penoffset1
6750   Y113t=Yaat+Penn/2
6760   Y114t=Yaat+Penn/2
6770   Y115t=Ymt+Penoffset1
6780   Y116t=Yjt+Penoffset1
6790   Y117t=Yat-Penn/2
6800   Y118t=Yat-Penn/2
6810   Y119t=Ybt+Penoffset3
6820   Y120t=Ylt+Penn/2
6830   Y121t=Ylt+Penn/2
6840   Y122t=Ybt+Penoffset3
6850   Y123t=Yadt+Penn/2
6860   Y124t=Yadt+Penn/2
6870   Y125t=Yact-Penoffset3
6880   Y126t=Yabt+Penn/2
6890   Y127t=Yabt-Penn/2
6900   Y128t=Yact-Penoffset3
6910   !
6920   Y1h=Ych-Penn/2
6930   Y2h=Ydh-Penoffseth2
6940   Y3h=Ybh+Penoffseth1
6950   Y4h=Yah+Penn/2
6960   Y5h=Yah+Penn/2
6970   Y6h=Yot
6980   !
6990   ! THE FOLLOWING POINTS ARE USED TO MERGE THE MAGIC-TEE AND THE HORN
7000   ! INTO THE MONOPULSE SYSTEM.
7010   Y25mh=Yrmh+Penn/2
7020   Y26mh=Ylmh+Penoffset7
7030   Y27mh=Ygmh+Penoffset7
7040   Y28mh=Yemh+Penn/2
7050   Y31mh=Ydmh-Penn/2
7060   Y32mh=Yfmh-Penoffset7
7070   Y33mh=Ykmh-Penoffset7
7080   Y34mh=Yqmh-Penn/2
7090   !
7100   ! THE FOLLOWING POINTS ARE USED TO DRAW THE BULL'S EYES.
7110   Ybulltop=Yat+2*Space
7120   Ybullbottom=-Ybulltop
7130   !
7140   ! STEP NINE: MAKE THE DESIRED DRAWING.
7150   !
7160   ! THE CENTER OF THE DRAWING IS Xo,Yo. THIS IS THE POINT WHERE THE
7170   ! CENTERS OF THE SUM AND DIFFERENCE MICROSTRIP LINES WOULD CROSS IF
7180   ! EXTENDED. ALL OF THE DRAWINGS ARE REFERENCED FROM Xo,Yo.
7190   !
7200   IF Drawings="TEE" THEN 7600 !SKIP TO MAGIC-TEE SECTION.
7210   IF Drawings="MONO" THEN 8700 !SKIP TO THE MONOPULSE SYSTEM SECTION.
7220   !
7230   ! THIS SECTION DRAWS THE FIN-LINE SIDE OF A SINGLE FIN-LINE HORN.
7240   !
7250   IF Sides="M" THEN 7450  !SKIP TO MICROSTRIP SIDE.
7260   MOVE X1t,Y1t
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7270  DRAW X2i,Y2i+Im
7280  DRAW X3i,Y3i+Im
7290  IF Im=-1+Flip THEN 7330
7300  DRAW X4i,Y4i+Im
7310  DRAW X5i,Y5i+Im
7320  DRAW X6i,Y6i+Im
7330  GOSUB 10250  !DRAWS HORN.
7340  ! THE NEXT FEW LINES OUTLINE THE DIELECTRIC TABS TO MATCH THE FIXTURE.
7350  MOVE X40i,Y40i
7360  DRAW X41i,Y41i+Im
7370  DRAW X2ai,Y2ai+Im
7380  IF Im=-1+Flip THEN 7430
7390  MOVE X4ai,Y4ai+Im
7400  DRAW X42i,Y42i+Im
7410  DRAW X43i,Y43i+Im
7420  DRAW X6ai,Y6ai+Im
7430  NEXT Image
7440  GOTO 7570!SKIP MICROSTRIP SIDE.
7450  ! THIS SECTION DRAWS THE MICROSTRIP SIDE OF THE THE FIN-LINE HORN.
7460  MOVE X109i,Y109i
7470  DRAW X115i,Y110i
7480  DRAW X111i,Y111i
7490  DRAW X112i,Y112i
7500  Ybottomhms=Yoi-Ji/2-Qumicrostrip+Penn/2  !ONLY POINT UNIQUE TO HORN.
7510  DRAW X113i,Ybottomhms
7520  DRAW X114i,Ybottomhms
7530  DRAW X115i,Y115i
7540  DRAW X116i,Y116i
7550  DRAW X109i,Y109i
7560  !
7570  NEXT Fill
7580  GOTO 9300 ! SKIP TO THE RULER SECTION.
7590  !
7600  ! THIS SECTION DRAWS THE MAGIC-TEE.
7610  !
7620  IF Sides="M" THEN 8230     !THIS SKIPS TO MICRO-STRIP SIDE.
7630  !
7640  ! THIS SECTION DRAWS THE FIN-LINE SIDE OF THE FIN-LINE TEE.
7650  MOVE X1i,Y1i+Im
7660  DRAW X2i,Y2i+Im
7670  DRAW X3i,Y3i+Im
7680  IF Im=-1+Flip THEN GOTO 7720
7690  DRAW X4i,Y4i+Im
7700  DRAW X5i,Y5i+Im
7710  DRAW X6i,Y6i+Im
7720  DRAW X7i,Y7i+Im
7730  DRAW X8i,Y8i+Im
7740  DRAW X10i,Y10i+Im
7750  DRAW X11i,Y11i+Im
7760  DRAW X12i,Y12i+Im
7770  DRAW X13i,Y13i+Im
7780  DRAW X14i,Y14i+Im
7790  DRAW X16i,Y16i+Im
7800  MOVE X19i-Mi/2+Penn/2,Y19i
7810  FOR Angle=89 TO 180 STEP 2
7820  X=X19i-(Mi/2-Penn/2)*SIN(Angle)
7830  Y=(Y19i-(Mi/2-Penn/2)*COS(Angle))*Im
7840  DRAW X,Y
7850  NEXT Angle
7860  DRAW X21i,Y21i+Im
7870  IF Y22i>Yoi THEN GOTO 7900 !KEEPS PEN BETWEEN COUPLED SLOTS.
7880  Y22i=Yoi
7890  Y23i=Yoi
7900  DRAW X22i,Y22i+Im
7910  DRAW X23i,Y23i+Im
7920  DRAW X24i,Y24i+Im
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7930    DRAW X25t,Y25t+Im
7940    DRAW X28t,Y28t+Im
7950    DRAW X29t,Y29t+Im
7960    DRAW X30t,Y30t+Im
7970    DRAW X31t,Y31t+Im
7980    DRAW X34t,Y34t+Im
7990    DRAW X35t,Y35t+Im
8000    DRAW X36t,Y36t+Im
8010    DRAW X37t,Y37t+Im
8020    DRAW X38t,Y38t+Im
8030    DRAW X39t,Y39t+Im
8040    FOR Angle=8 TO 91 STEP 1
8050    X=X19t-(Mt/2+Ht+Penn/2)*SIN(Angle)
8060    Y=(Y19t+(Mt/2+Ht+Penn/2)*COS(Angle))+Im
8070    DRAW X,Y
8080    NEXT Angle
8090    MOVE X40t,Y40t
8100    DRAW X41t,Y41t+Im
8110    DRAW X2at,Y2at+Im
8120    IF Im=-1+Flip THEN GOTO 8170
8130    MOVE X4at,Y4at+Im
8140    DRAW X42t,Y42t+Im
8150    DRAW X43t,Y43t+Im
8160    DRAW X6at,Y6at+Im
8170    MOVE X1,Y1
8180    NEXT Image
8190    GOTO 8560        !FIN-LINE SIDE IS COMPLETE. THIS COMMAND SKIPS THE
8200                     ! MICROSTRIP SIDE, AND CONTINUES ON WITH THE PROGRAM.
8210    !
8220    ! THE FOLLOWING SECTION DRAWS THE MICROSTRIP SIDE OF THE MAGIC-TEE.
8230    MOVE X101t,Y101t
8240    DRAW X102t,Y102t
8250    DRAW X103t,Y103t
8260    DRAW X104t,Y104t
8270    DRAW X105t,Y105t
8280    DRAW X106t,Y106t
8290    DRAW X107t,Y107t
8300    DRAW X108t,Y108t
8310    DRAW X101t,Y101t
8320    MOVE X109t,Y109t+Flip
8330    DRAW X110t,Y110t+Flip
8340    DRAW X111t,Y111t+Flip
8350    DRAW X112t,Y112t+Flip
8360    DRAW X113t,Y113t+Flip
8370    DRAW X114t,Y114t+Flip
8380    DRAW X115t,Y115t+Flip
8390    DRAW X116t,Y116t+Flip
8400    DRAW X109t,Y109t+Flip
8410    MOVE X117t,Y117t
8420    DRAW X118t,Y118t
8430    DRAW X119t,Y119t
8440    DRAW X120t,Y120t
8450    DRAW X121t,Y121t
8460    DRAW X122t,Y122t
8470    DRAW X117t,Y117t
8480    MOVE X123t,Y123t
8490    DRAW X124t,Y124t
8500    DRAW X125t,Y125t
8510    DRAW X126t,Y126t
8520    DRAW X127t,Y127t
8530    DRAW X128t,Y128t
8540    DRAW X123t,Y123t
8550    ! THIS SECTION IS COMMON TO BOTH THE FIN-LINE AND THE MICROSTRIP SIDES.
8560    ! THE NEXT FEW LINES PLACE THE NUMBER OF THE MAGIC TEE NEAR THE DEVICE.
8570    CSIZE .06+In,.5
8580    MOVE Xvt+Penn,Yct+.60*Tt+Penn
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8590    LABEL Titles    !DRAWS THE NAME OF THE TEE ON THE DEVICE.
8600    NEXT Fill
8610    GOTO 9385 !MAGIC-TEE IS DONE, SKIP TO THE RULER SECTION.
8620    !
8630    ! THIS SECTION DRAWS THE FIN-LINE SIDE OF THE MONOPULSE SYSTEM.
8640    !
8650    IF Sides="M" THEN 9170       !SKIP TO MICROSTRIP SIDE.
8660    MOVE X1t,Y1t+Im
8670    DRAW X2t,Y2t+Im
8680    DRAW X3t,Y3t+Im
8690    IF Im=-1+Flip THEN 8730     !Flip CONTROLS THE LOCATION OF PORT-3.
8700    DRAW X4t,Y4t+Im
8710    DRAW X5t,Y5t+Im
8720    DRAW X6t,Y6t+Im
8730    DRAW X7t,Y7t+Im       !THIS IS THE UPPER LEFT EDGE OF THE FIXTURE.
8740    DRAW X34mh,Y34mh+Im !THIS IS THE FIRST POINT IN THE MONOPULSE SYSTEM
8750                         ! PORTION.
8760    DRAW X31mh,Y31mh+Im
8770    Imm=Im           !USED TO DRAW HORN CORRECTLY.
8780    GOSUB 10250   !DRAW HALF OF THE HORN.
8790    DRAW X31t,Y31t+Im
8800    DRAW X34t,Y34t+Im
8810    DRAW X35t,Y35t+Im
8820    DRAW X36t,Y36t+Im
8830    DRAW X37t,Y37t+Im
8840    DRAW X38t,Y38t+Im
8850    DRAW X39t,Y39t+Im
8860    FOR Angle=0 TO 91 STEP 1    !NEXT FIVE LINES DRAW THE LEFT QUARTER ARC.
8870    X=X19t-(Ht/2+Ht+Penn/2)*SIN(Angle)
8880    Y=(Y19t+(Ht/2+Ht+Penn/2)*COS(Angle))+Im
8890    DRAW X,Y
8900    NEXT Angle
8910    MOVE X25mh,Yot        !MOVE BACK TO CENTER OF THE MONOPULSE SYSTEM PORTION.
8920    DRAW X20mh,Y20mh+Im
8930    Imm=-1+Im !FLIPS Im IN THE HORN SUBROUTINE TO DRAW BOTTOM HALF OF HORN.
8940    GOSUB 10250          !DRAW BOTTOM HALF OF THE HORN
8950    DRAW X20t,Y28t+Im    !THIS IS THE MAGIC-TEE PORTION AGAIN.
8960    DRAW X25t,Y25t+Im
8970    DRAW X24t,Y24t+Im
8980    DRAW X23t,Y23t+Im
8990    DRAW X22t,Y22t+Im
9000    DRAW X21t,Y21t+Im
9010    DRAW X20t,Y20t+Im
9020    FOR Angle=0 TO 91 STEP 1
9030    X=X19t-(Ht/2-Penn/2)*SIN(Angle)
9040    Y=(Y19t+(Ht/2-Penn/2)*COS(Angle))+Im
9050    DRAW X,Y
9060    NEXT Angle
9070    MOVE X40t,Y40t
9080    DRAW X41t,Y41t+Im
9090    DRAW X2at,Y2at+Im
9100    IF Im=-1+Flip THEN 9150
9110    MOVE X4at,Y4at+Im
9120    DRAW X42t,Y42t+Im
9130    DRAW X43t,Y43t+Im
9140    DRAW X6at,Y6at+Im
9150    NEXT Image
9160    GOTO 9360 !SKIP THE MICROSTRIP SIDE.
9170    ! THIS SECTION DRAWS THE MICROSTRIP SIDE OF THE MONOPULSE SYSTEM.
9180    MOVE X101t,Y101t
9190    DRAW X102t,Y102t
9200    DRAW X103t,Y103t
9210    DRAW X104t,Y104t
9220    DRAW X105t,Y105t
9230    DRAW X106t,Y106t
9240    DRAW X107t,Y107t
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9250    DRAW X100t,Y100t
9260    DRAW XI01t,Y101t
9270    MOVE X109t,Y109t
9280    DRAW X110t,Y110t
9290    DRAW X111t,Y111t
9300    DRAW X112t,Y112t
9310    DRAW X113t,Y113t
9320    DRAW X114t,Y114t
9330    DRAW XI15t,Y115t
9340    DRAW XI16t,Y116t
9350    DRAW X109t,Y109t
9360    NEXT Fill
9370    !
9380    ! THIS SECTION DRAWS THE RULER AND COMMENTS.
9390    !
9400    IF Rulers="N" THEN 10200
9410    ! THE NEXT PORTION OF THE PROGRAM DRAWS THE NUMBERS AND COMMENTS BELOW
9420    ! THE RULER.
9430    LINE TYPE I           !THIS RETURNS THE PLOTTER TO A NORMAL HIGH SPEED LINE.
9440    Penn=Pen
9450    FOR I=1 TO 3*Scale STEP I
9460    Penn=.8*Pen+Penn
9470    CSIZE .135*In,.35    !(HEIGHT OF LETTERS),(WIDTH/HEIGHT RATIO)
9480    MOVE XI50t+.05*In+Penn,Yajt-Penn
9490    LABEL "I"
9500    MOVE XI50t+1.05*In+Penn,Yajt-Penn
9510    LABEL "2"
9520    MOVE XI50t+2.05*In+Penn,Yajt-Penn
9530    LABEL "3"
9540    IF I>Scale THEN GOTO 9800
9550    MOVE XI50t+Penn,Yakt+Penn
9560    CSIZE .08*In,.48    !(HEIGHT OF LETTERS),(WIDTH/HEIGHT RATIO)
9570    LABEL "WHEN REDUCED TO 1:1 SCALE, THIS"
9580    MOVE XI50t+Penn,Yalt+Penn
9590    LABEL "RULER WILL BE EXACTLY IN INCHES."
9600    !
9610    ! THE FOLLOWING LINES DRAW THE COMMENTS BETWEEN THE BULL'S EYES.
9620    MOVE Xbullleft+.2*In+Penn,Ybulltop+Penn
9630    LABEL "MAGIC TEE ";Titles;" by LCDR ROWLEY"
9640    !
9650    ! THE FOLLOWING LINES DRAW THE BULL'S EYES.
9660    MOVE Xbullleft+Penn,Ybulltop+.15*In+Penn
9670    DRAW Xbullleft+Penn,Ybulltop-.05*In+Penn
9680    MOVE Xbullleft-.I*In+Penn,Ybulltop+Penn+.85*In
9690    DRAW Xbullleft+.I*In+Penn,Ybulltop+Penn+.05*In
9700    FOR Angle=0 TO 360 STEP 5
9710    DRAW .I*In*COS(Angle)+Xbullleft+Penn,.I*In*SIN(Angle)+Ybulltop+.05*In+Penn
9720    NEXT Angle
9730    MOVE Xbullright+Penn,Ybulltop+.15*In+Penn
9740    DRAW Xbullright+Penn,Ybulltop-.05*In+Penn
9750    MOVE Xbullright-.I*In+Penn,Ybulltop+Penn+.05*In
9760    DRAW Xbullright+.I*In+Penn,Ybulltop+Penn+.05*In
9770    FOR Ang=0 TO 360 STEP 5
9780    DRAW .I*In*COS(Ang)+Xbullright+Penn,.I*In*SIN(Ang)+Ybulltop+.85*In+Penn
9790    NEXT Ang
9800    NEXT I
9810    ! THE NEXT SEGMENT OF THE PROGRAM DRAWS THE RULER.
9820    J=I
9830    Penn=Pen
9840    FOR I=0 TO 6 STEP 1
9850    J=J+I
9860    XI50t=Xat+I*(.5*In)
9870    MOVE XI50t+Penn,Yaet       !TOP OF THE RULER.
9880    ! THE RULER IS DRAWN IN SIX HALF INCH SEGMENTS. THE ODD SEGMENTS START
9890    ! WITH A LONG INCH MARK, WHILE THE EVEN SEGMENTS START WITH A SHORTER
9900    ! HALF INCH MARK. THE NEXT SEVEN LINES CONSTRUCT THE FIRST MARK IN EACH
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9910  I HALF IHCH SEGMENT.
9920  IF I=1 THEH GOTO 9980      ISKIPS TO THE HALF IHCH MARK LEHGTH.
9930 .IF I=3 THEH GOTO 9980      ISKIPS TO THE HALF IHCH MARK LEHGTH.
9940  IF I=5 THEH GOTO 9980      ISKIPS TO THE HALF INCH MARK LEHGTH.
9950  DRAW X150t+Penn,Yajt       ITHIS IS THE IHCH MARK (LOHGEST LIHE).
9960  IF I=6 THEH GOTO 10150     ITHIS TERMINATES THE RULER AFTER THREE INCHES.
9970  GOTO 10000                 !THIS SKIPS THE HALF IHCH MARK LENGTH.
9980  DRAW X150t+Penn,Yait       ITHIS IS THE HALF INCH MARK LEHGTH.
9990  I THE HEXT 14 LIHES DRAWS EACH HALF IHCH SEGMEHT OF THE RULER.
10000 MOVE X150t+Penn+Sixteenth,Yaet
10010 DRAW X150t+Penn+Sixteenth,Yaft
10020 MOVE X150t+Penn+2*Sixteenth,Yaet
10030 DRAW X150t+Penn+2*Sixteenth,Yagt
10040 MOVE X150t+Penn+3*Sixteenth,Yaet
10050 DRAW X150t+Penn+3*Sixteenth,Yaft
10060 MOVE X150t+Penn+4*Sixteenth,Yaet
10070 DRAW X150t+Penn+4*Sixteenth,Yaht
10080 MOVE X150t+Penn+5*Sixteenth,Yaet
10090 DRAW X150t+Penn+5*Sixteenth,Yaft
10100 MOVE X150t+Penn+6*Sixteenth,Yaet
10110 DRAW X150t+Penn+6*Sixteenth,Yagt
10120 MOVE X150t+Penn+7*Sixteenth,Yaet
10130 DRAW.X150t+Penn+7*Sixteenth,Yaft
10140 NEXT I
10150 IF J>8*Scale THEH GOTO 10200 ITHIS TERMINATES THE PROGRAM WHEN LIHES ARE
10160                              I ARE THICK ENOUGH.
10170 Penn=Penn+Pen               I THIS SHIFTS THE RULER SLIGHTLY AHD REDRAWS
10180                             I IT. THIS MAKES THE LIHES THICK EHOUGH.
10190 GOTO 9940                   I THIS COHTIHUES THE LOOP.
10200 PEH 0                       I THE PEH IS PUT BACK IH THE HOLDER.
10210 MOVE 5000,500               I THE PEH DRIVE ARM IS HOVED ASIDE.
10220 STOP
10230 I
10240 I STEP TEH: HORN SUBROUTIHE.
10250 I THE FOLLOWIHG SUBROUTIHE DRAWS HALF OF THE HORN AT A TIME.
10260 IF Drawings="HORN" THEH 10320ISKIP OFFSET FOR SINGLE HORN.
10270 Xoff=Xact-Xvt+Lambdad       IMOVES HORN OPEHING OHE HAVELEHGTH LEFT OF THE
10280                             I LAST SLOT BEHD (MOHOHORH .ONLY).
10290 Yoff=(Vt+Jt/2)*1m           I HOVES HORN OPENIHG UP/DOWN TO MATCH SLOTS IH
10300                             I THE HOHOPULSE SYSTEH.
10310 GOTO 10350                  ISKIP TO DRAWIHG ROUTIHE.
10320 Imm=Im    IHEEDED FOR SIHGLE HORN OHLY.
10330 Xoff=0    IIHSURES OFFSETS ARE ZERO FOR SINGLE HORH.
10340 Yoff=0
10350 DRAW X1h+Xoff,Y1h*Imm+Yoff
10360 DRAW X2h+Xoff,Y2h*Imm+Yoff
10370 DRAW X3h+Xoff,Y3h*Imm+Yoff
10380 DRAW X4h+Xoff,Y4h*Imm+Yoff
10398 IF Drawings="MOHO" THEH 10420 IPOIHTS 5&6 ARE FOR THE SINGLE HORN ONLY.
10400 DRAW X5h+Xoff,Y5h*1m+Yoff
10410 DRAW X6h+Xoff,Y6h*1m+Yoff
10420 I THE HEXT PORTIOH DRAWS THE DOTTED LINE OF THE LEHS ARC.
10430 IF F(11)>I THEH 7430                IDOTTED LINE.
10448 LINE TYPE 4,.5*Scale                IBeta(MAX)=Angleh.'
10455 FOR Beta=0 TO Angleh STEP I          IBeta(MAX)=Angleh.
18468 I THE FOLLOWIHG LIHES ARE THE EQUATIOH THAT DEFINES A PARALLEL
10470 I PHASE FROHT LEHS.
10480 Aarc=I-COS(Beta)
10490 Barc=I-COS(Angleh)
10500 Carc=I/SQR(Er)-COS(Beta)
10510 Darc=COS(Angleh)-1/SQR(Er)
10520 Earc=COS(Beta)*(I-SQR(Er))
10530 Brh=Bh*(Aarc+Barc*Carc/Darc)/Earc   ITHICKHESS OF LEHS.
10540 Radh=Bh+Brh+Pen                      IRADIUS FROM Xoh.Yoh TO LEH0 EDGE.
10550                                      I Bh 10 RADIU0 OF ARC WITHOUT LEH0.
10560 Xarc=Xoh+Radh*COS(Beta)             IX POSITIOH OF ARC.
```

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10570 Yarc=Yoh+Radh*SIN(Beta)              !Y POSITION OF ARC.
10580 IF Beta=0 THEN MOVE Xarc+Xoff,Yarc+Yoff  !POSITIONS PEN FOR ARC.
10590 DRAW Xarc+Xoff,Yarc*Imm+Yoff
10600 NEXT Beta
10610 IF Drawing$="H" THEN 10650 ! FOLLOWING LINES ARE FOR MONOHORN ONLY.
10620 MOVE X4h+Xoff,Y4h*Imm+Yoff  !RETURNS DRAWING TO PROPER POINT AFTER LEHS IS
10630                             ! IS DRAWN.
10640 LINE TYPE Linenumber,Segmentsize  !RETURNS LINE TYPE TO PREVIOUS SETTING.
10650 RETURN
10660 END
```

FIN-LINE MAGIC-TEE NUMBER THREE
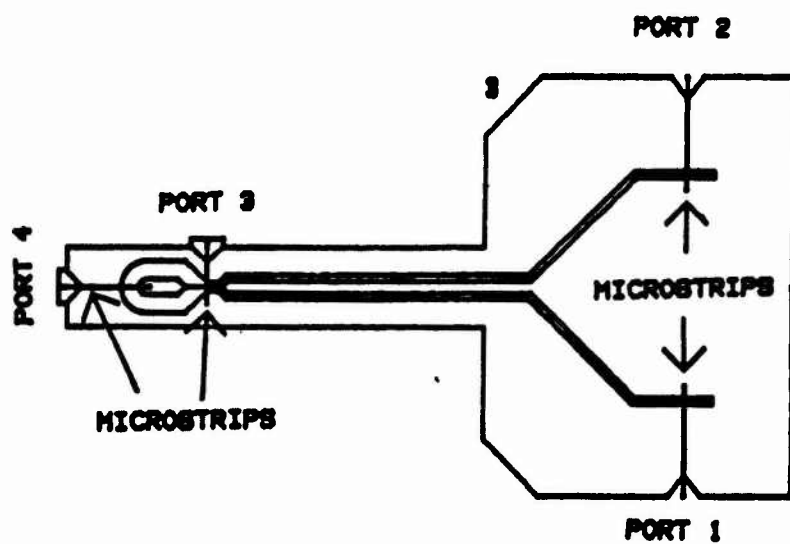
Figure 1   Fin-Line Magic-Tee.
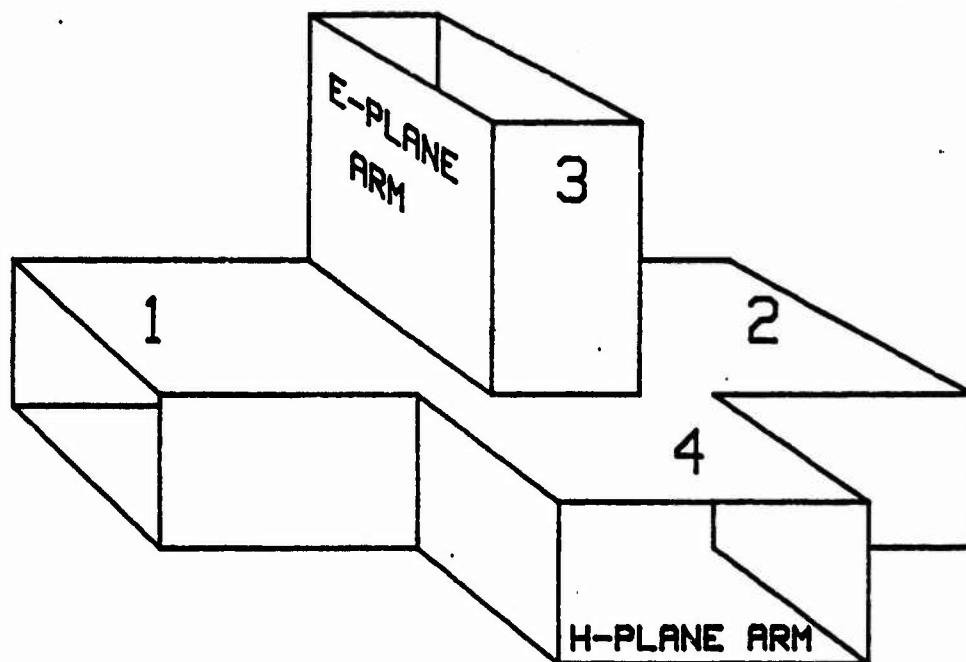
Figure 2   Waveguide Magic-Tee.

$$S = \begin{bmatrix} 0 & 0 & .7070^{J000} & .7070^{J000} \\ 0 & 0 & .7070^{-J180} & .7070^{J000} \\ .7070^{J000} & .7070^{-J180} & 0 & 0 \\ .7070^{J000} & .7070^{J000} & 0 & 0 \end{bmatrix}$$

Figure 3   Theoretical Magic-Tee Scattering Matrix.
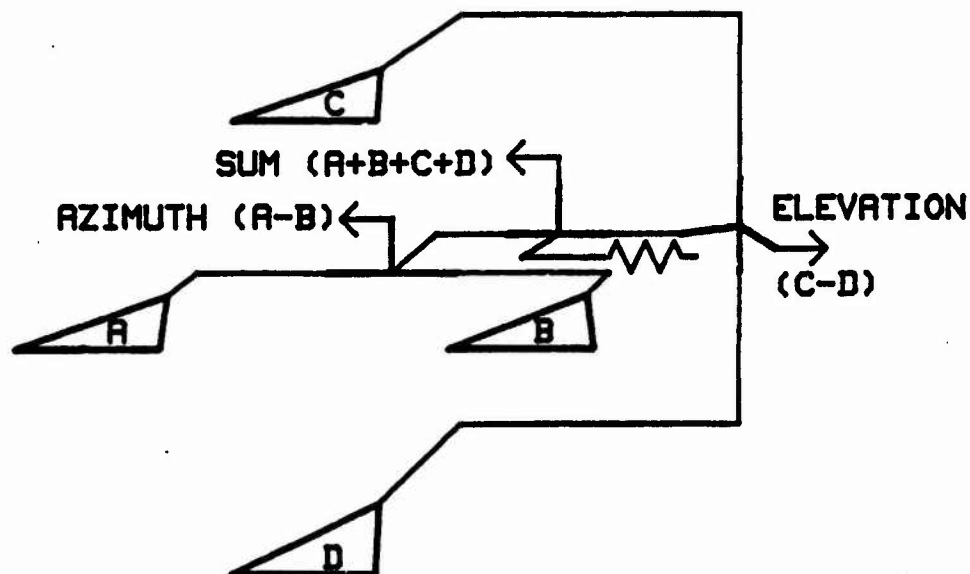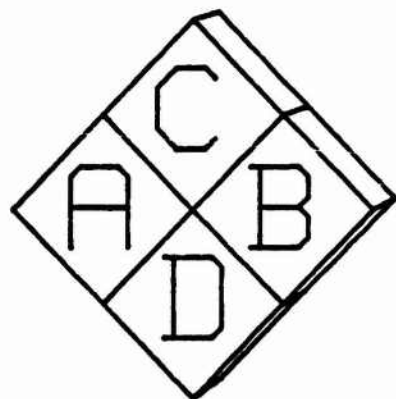
Figure 4   Diamond Monopulse Feed.

Figure 5   Square Monopulse Feed.

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Degrees Left(-) and Right(+) of Boresight .
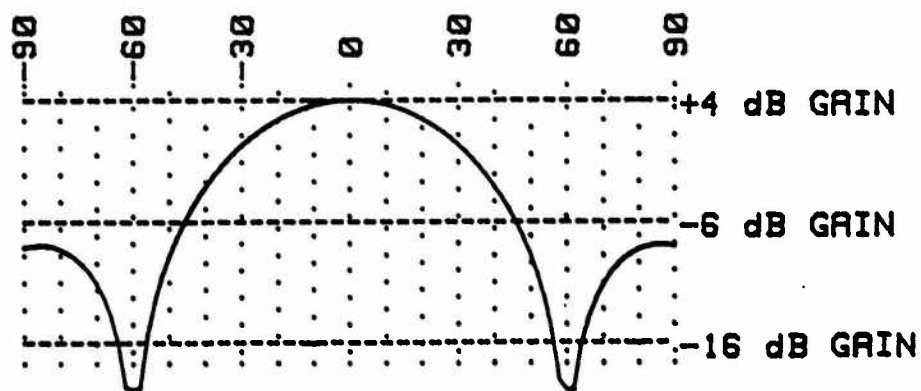
Computer Simulation of Actual
Element Pattern.
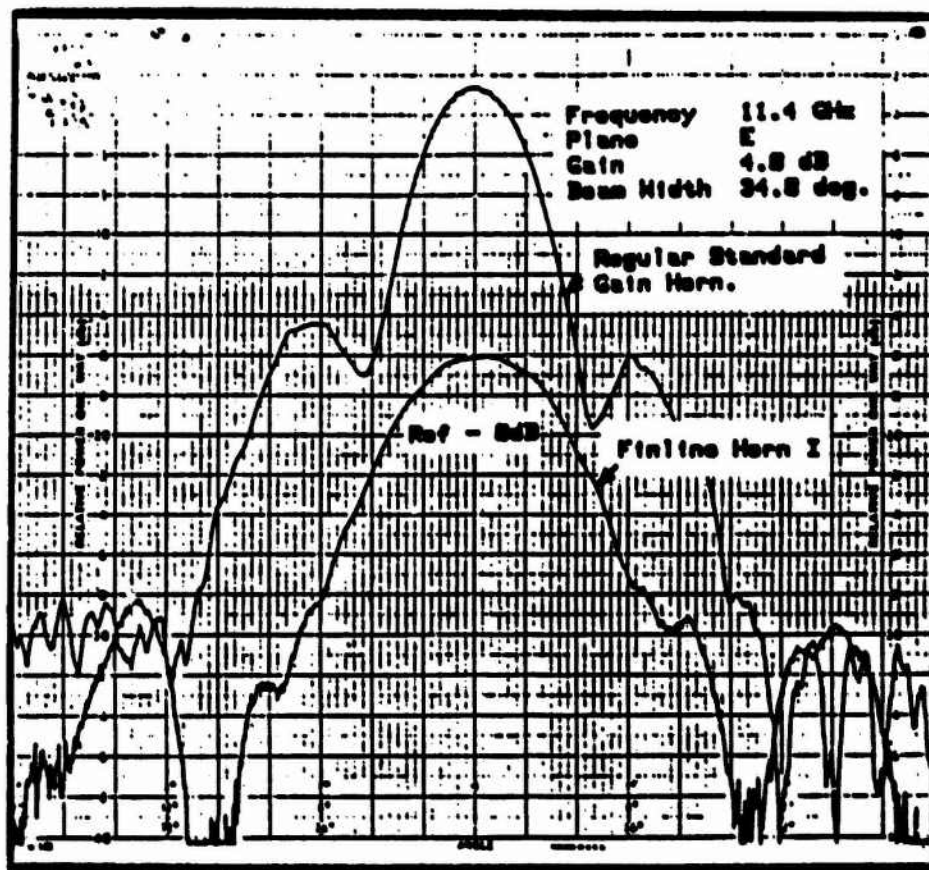
Actual Element Pattern at 11.4 GHZ.

Figure 6   Simulated and Actual Element Pattern.

$I(y)$
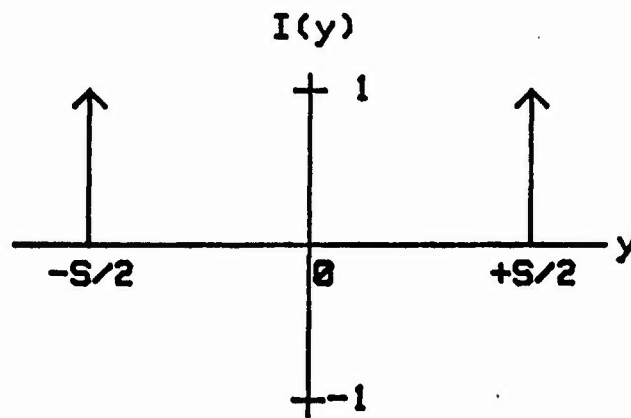
Spacing between elements of a Monopulse Feed,
showing the elements in phase with each other.
Each element is represented by a delta function.
The element pattern is accounted for later.

$|COS(X)|$
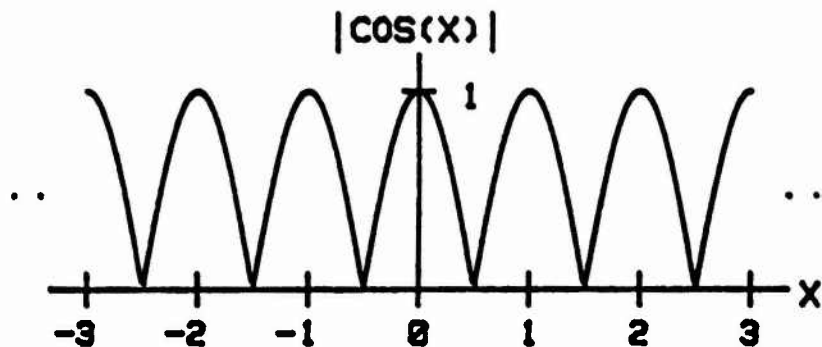
Fourier Transformed Sum Group pattern.

NOTES:

1. X=SIN(Theta)*(S/Lambda); Theta is the angle
left(-), or right(+), of the antenna's boresight.
'S' is the spacing between the antenna elements.

2. At +/- 90 degrees, SIN(Theta) reaches it's
maximum value of one. These two pionts are called
the pattern's visible limits. They ocour where
X=+/-(1)*(S/Lambda) on the transformed pattern.

Figure 7 Theoretical Sum Group Pattern.

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I(y)

-S/2   +S/2
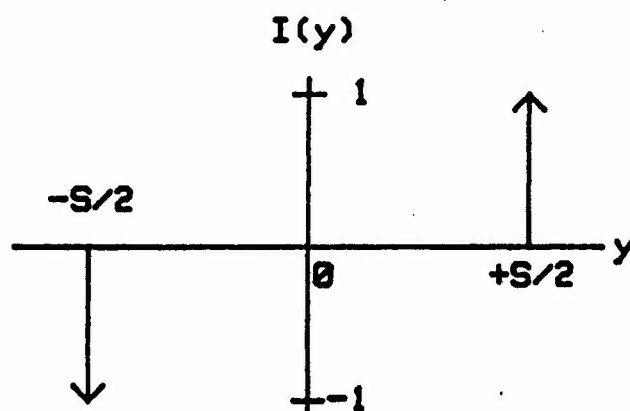
Spacing between elements of a Monopulse Feed,
showing the elements out of phase with each other.
Each element is represented by a delta function.
The element pattern is accounted for later.

|SIN(X)|
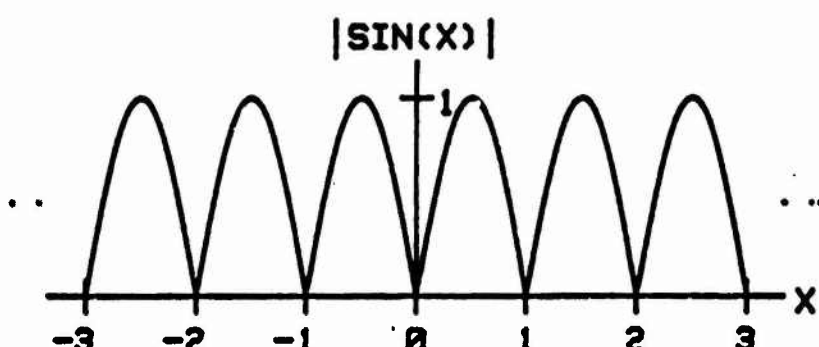
Fourier Transformed Difference Group Pattern.

NOTES:

1. X=SIN(Theta)*(S/Lambda); Theta is the angle
left(-), or right(+), of the antenna's bore sight.
'S' is the spacing between the antenna elements.

2. At +/- 90 degrees, SIN(Theta) reaches it's
maximum value of one. These two pionts are called
the pattern's visible limits. They occur where
X=+/-(1)*(S/Lambda) on the transformed pattern.

Figure 8   Theoretical Difference Group Pattern.

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Figure 9 Theoretical Sum Pattern.

Degrees Left(-) and Right(+) of Boresight
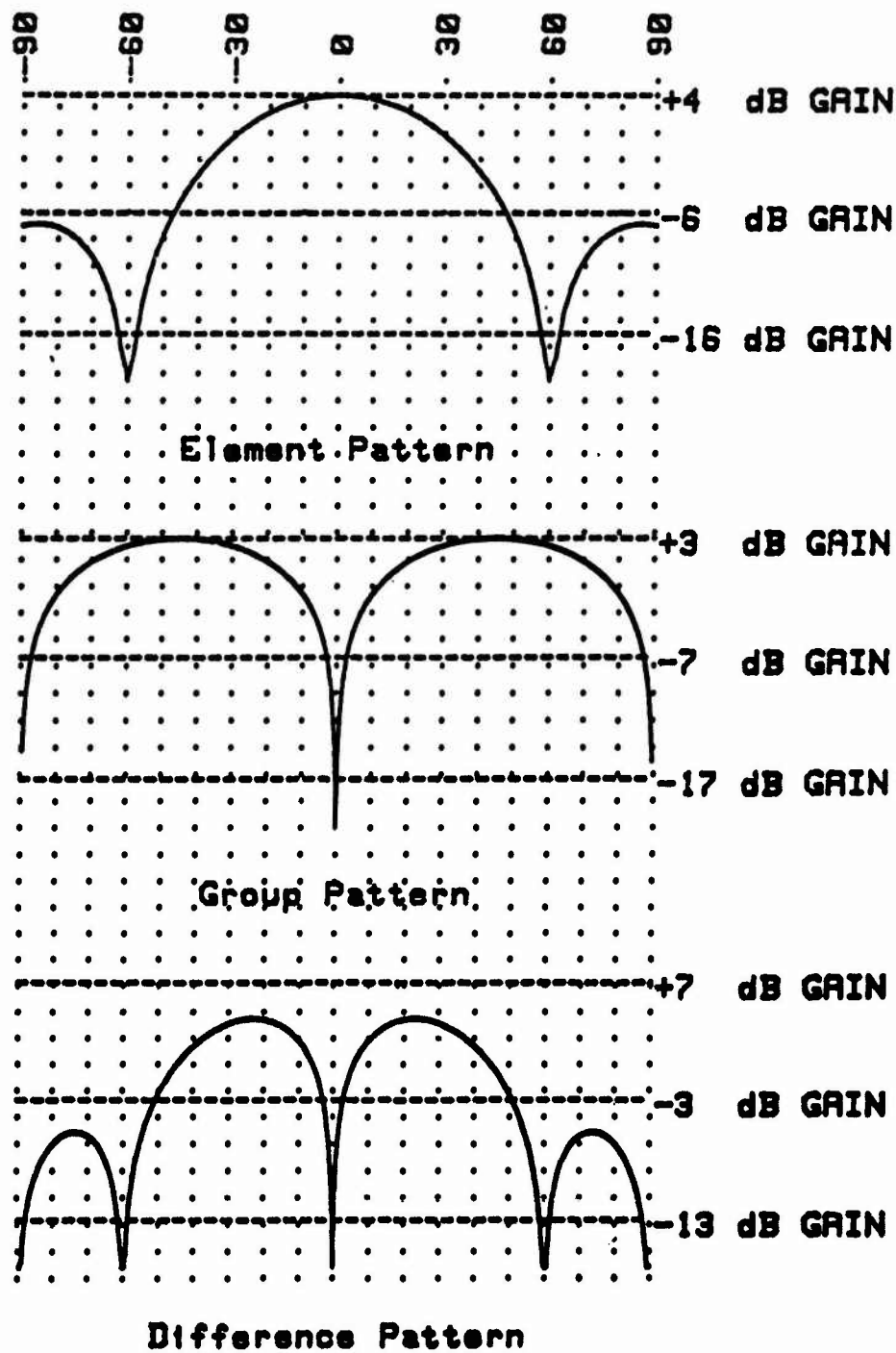
Figure 10   Theoretical Difference Pattern.
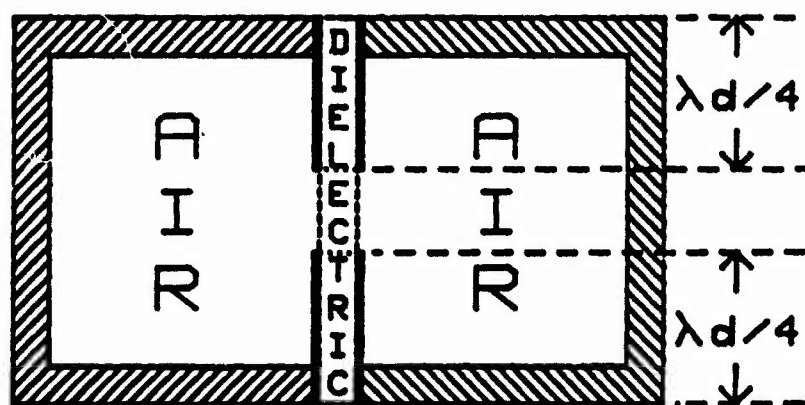
Figure 11 Bilateral Fin-Line.
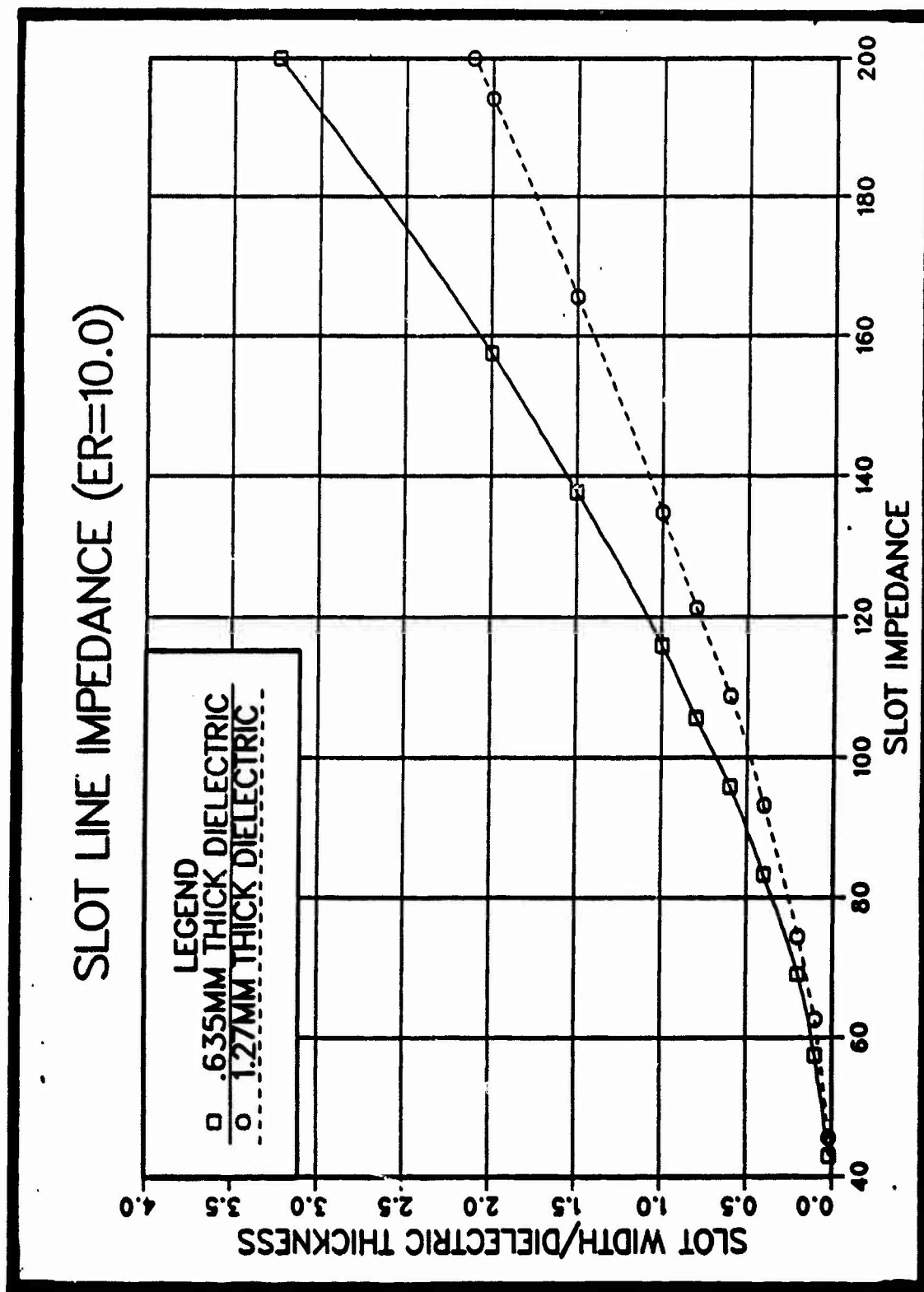
Figure 12   Slotline Impedance.
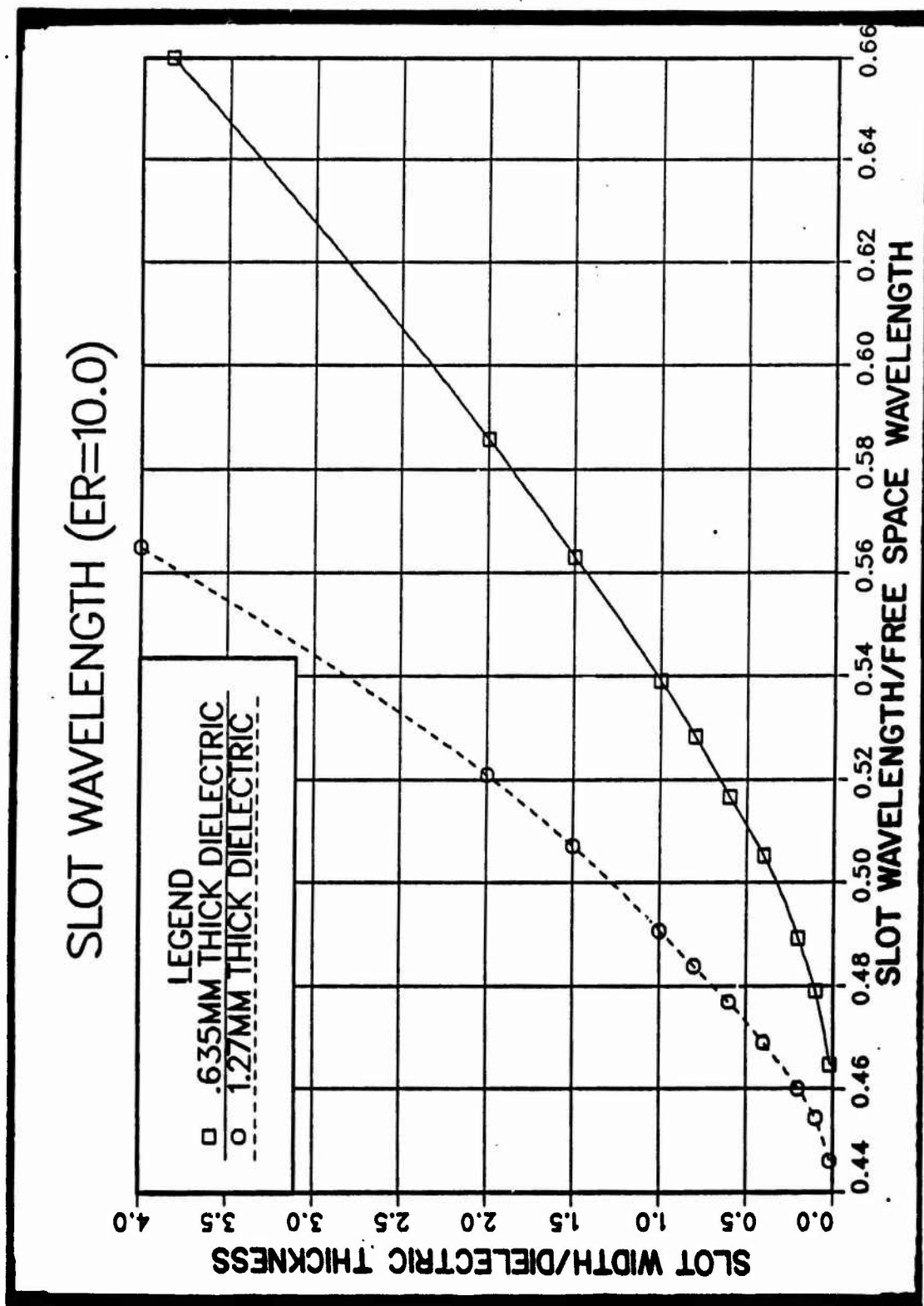
Figure 13   Slotline Wavelength.
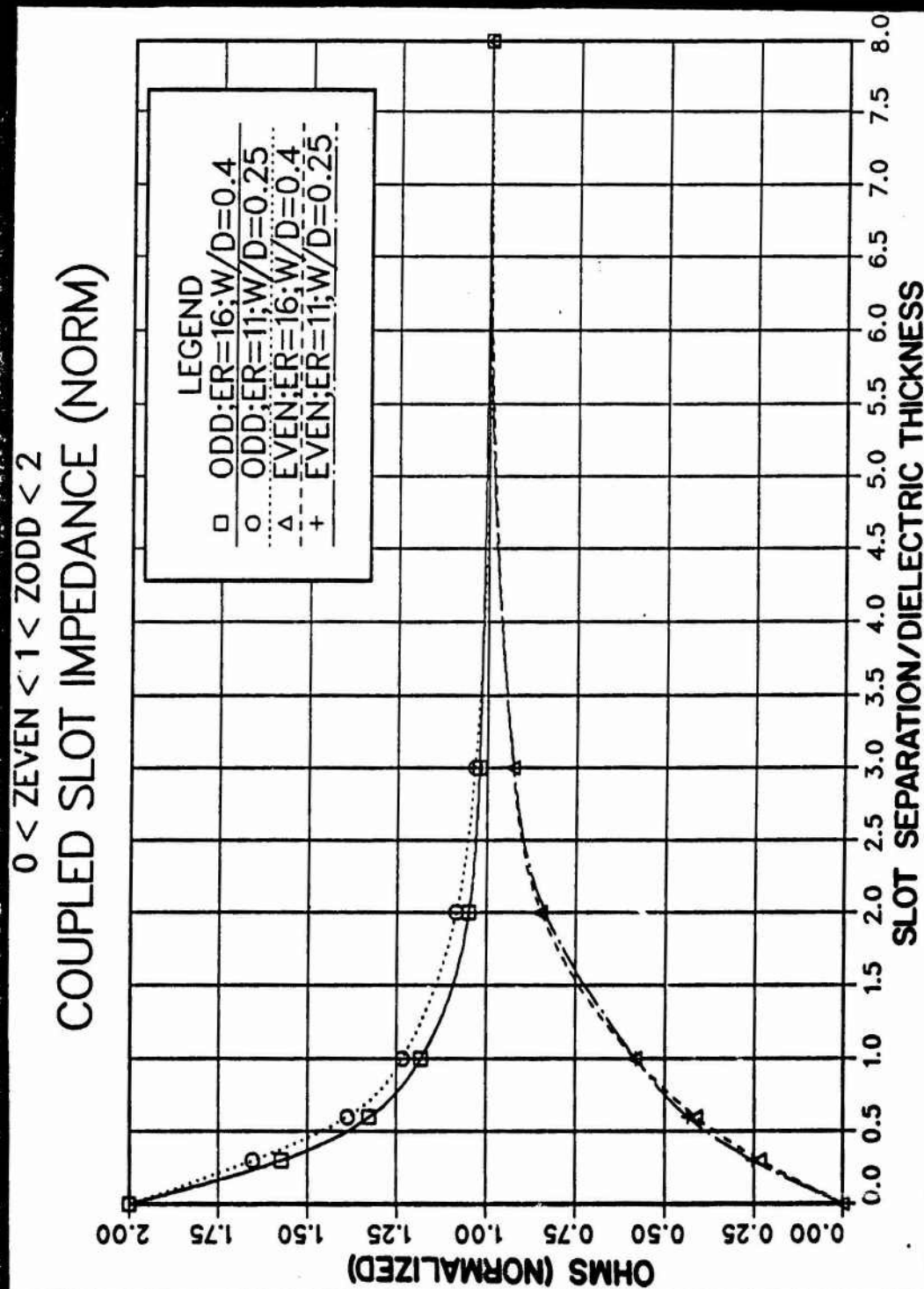
Figure 14   Normalized Coupled Slotline Impedance.
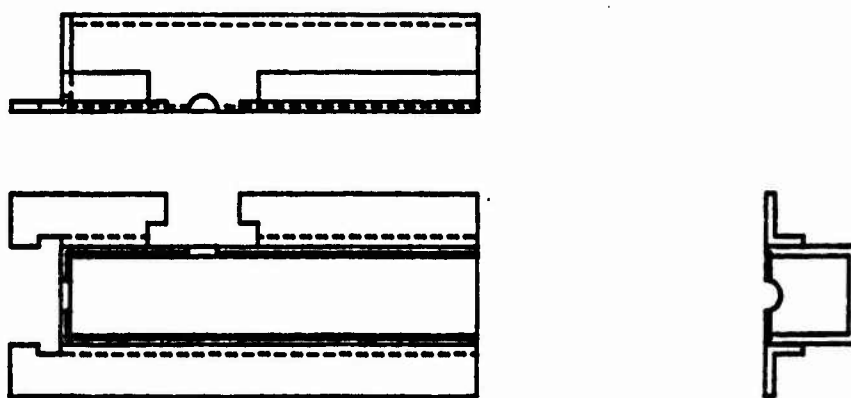
Figure 15   Original Fixture Design.

$$S = \begin{bmatrix} .202e^{-j169} & .023e^{-j023} & .462e^{+j004} & .310e^{+j006} \\ .030e^{-j030} & .269e^{-j148} & .314e^{-j176} & .192e^{-j020} \\ .487e^{-j008} & .313e^{-j179} & .132e^{-j001} & .077e^{-j069} \\ .316e^{-j002} & .277e^{+j017} & .052e^{-j042} & .154e^{+j162} \end{bmatrix}$$

This data was taken on an HP 8409B
Vector Network Analyzer at 10 GHZ.

This matrix is corrected for uneven
transmission line lengths caused by
launcher placement and fixture design.
10 degrees is added to port-1, 84
degrees is subtracted from port-3 and
22 degrees is added to port-4. The
corrections are uniform throughout the
matrix.

Figure 16   S-Matrix for Magic-Tee Number Two.

$$S = \begin{bmatrix} .452e^{-j157} & .042e^{+j046} & .473e^{+j005} & .327e^{+j004} \\ .100e^{+j042} & .483e^{-j169} & .282e^{-j173} & .410e^{-j009} \\ .442e^{-j008} & .280e^{-j182} & .435e^{+j024} & .046e^{+j012} \\ .305e^{-j007} & .376e^{+j003} & .046e^{+j011} & .170e^{+j000} \end{bmatrix}$$

This data was taken on an HP 8409B
Vector Network Analyzer at 11.4 GHZ.

This matrix is corrected for uneven
transmission line lengths caused by
launcher placement and fixture design.
106 degrees is added to port-1, 122
degrees is added to port-2 and 180
degrees is added to port-3. The
corrections are uniform throughout the
matrix.

Figure 17   S-Matrix for Magic-Tee Number Three.

Figure 18 SI: Phase and Magnitude for Magic-Tee Three.

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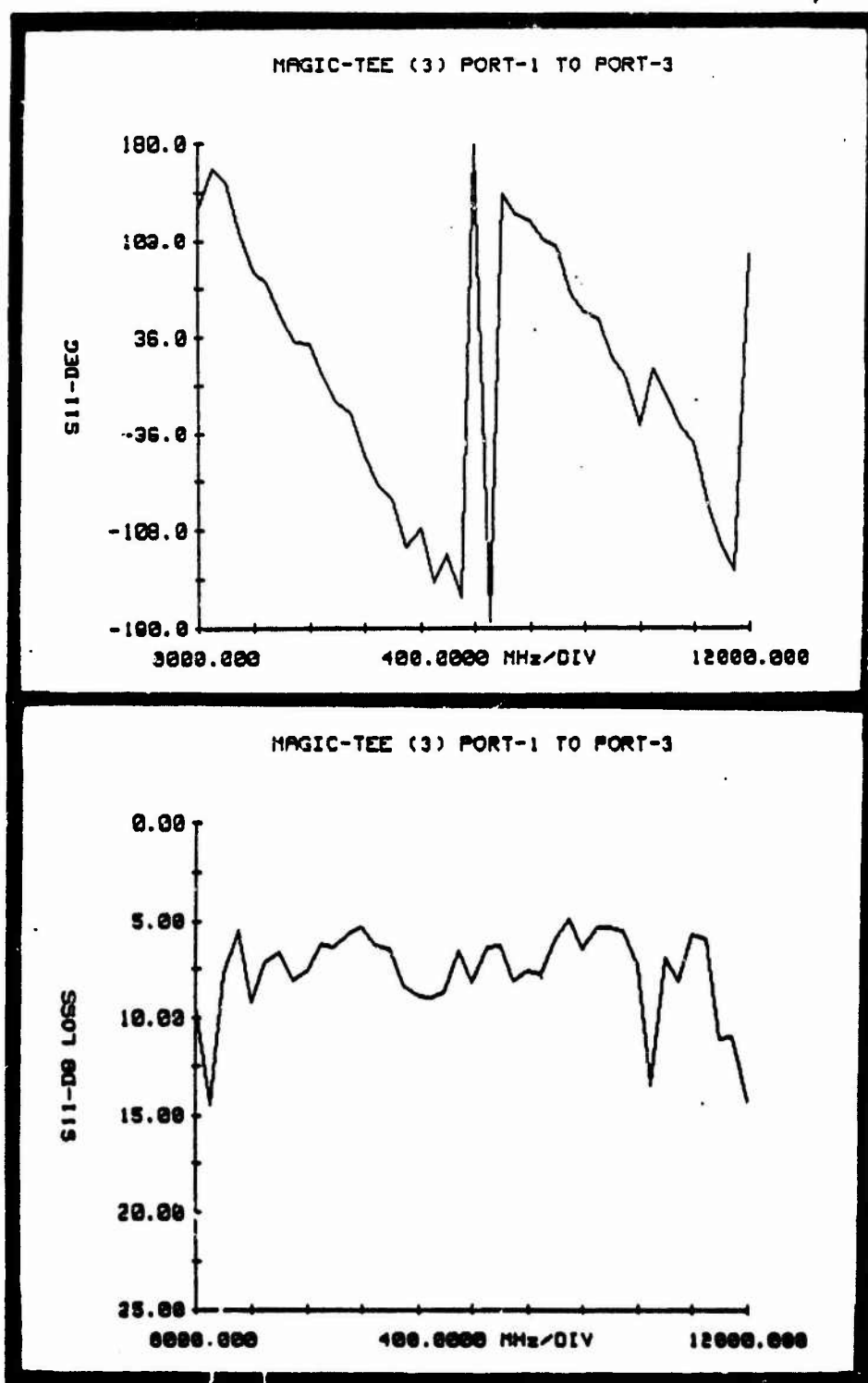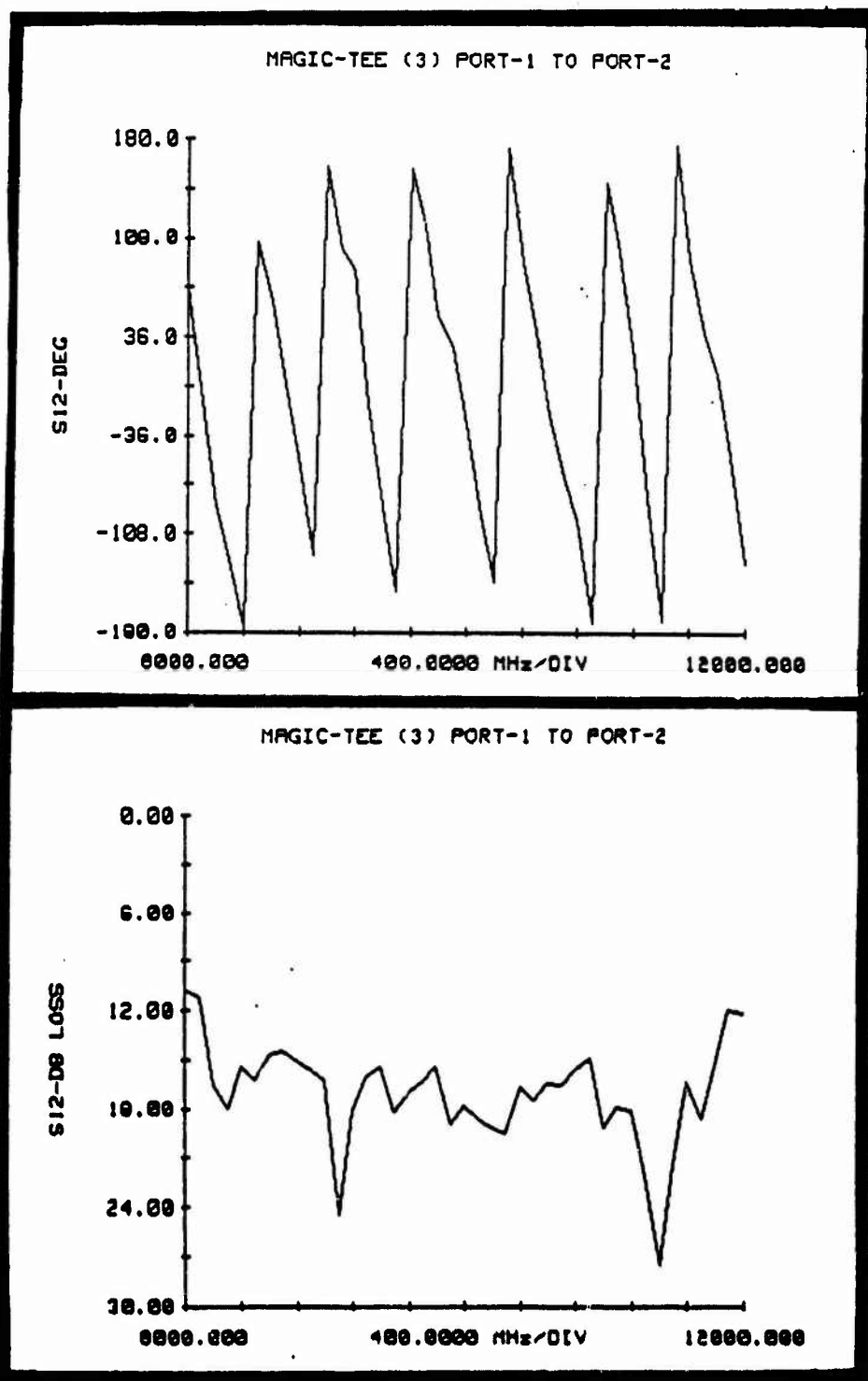
Figure 19 S12 Phase and Magnitude for Magic-Tee Three.
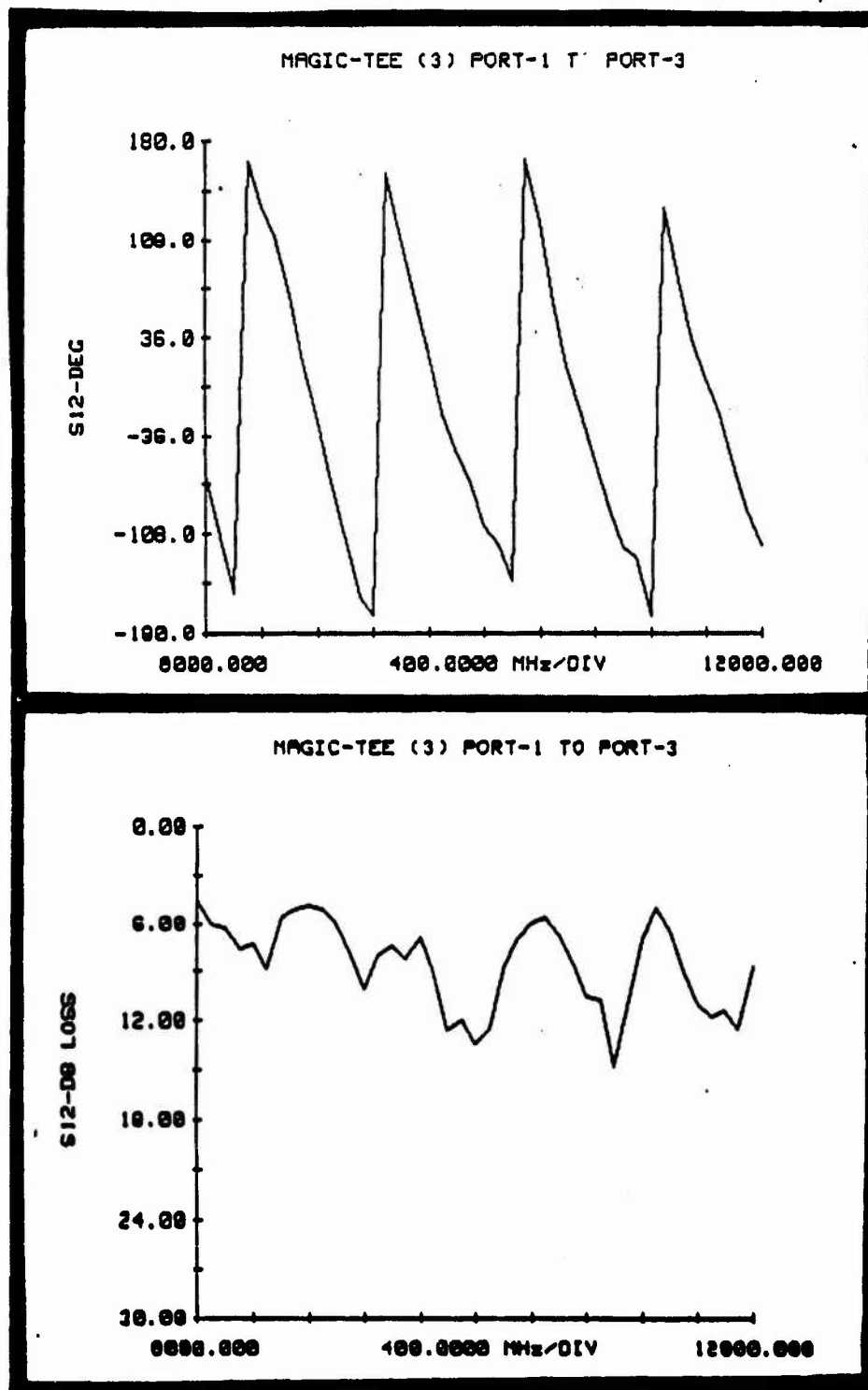
Figure 20   S13 Phase and Magnitude for Magic-Tee Three.
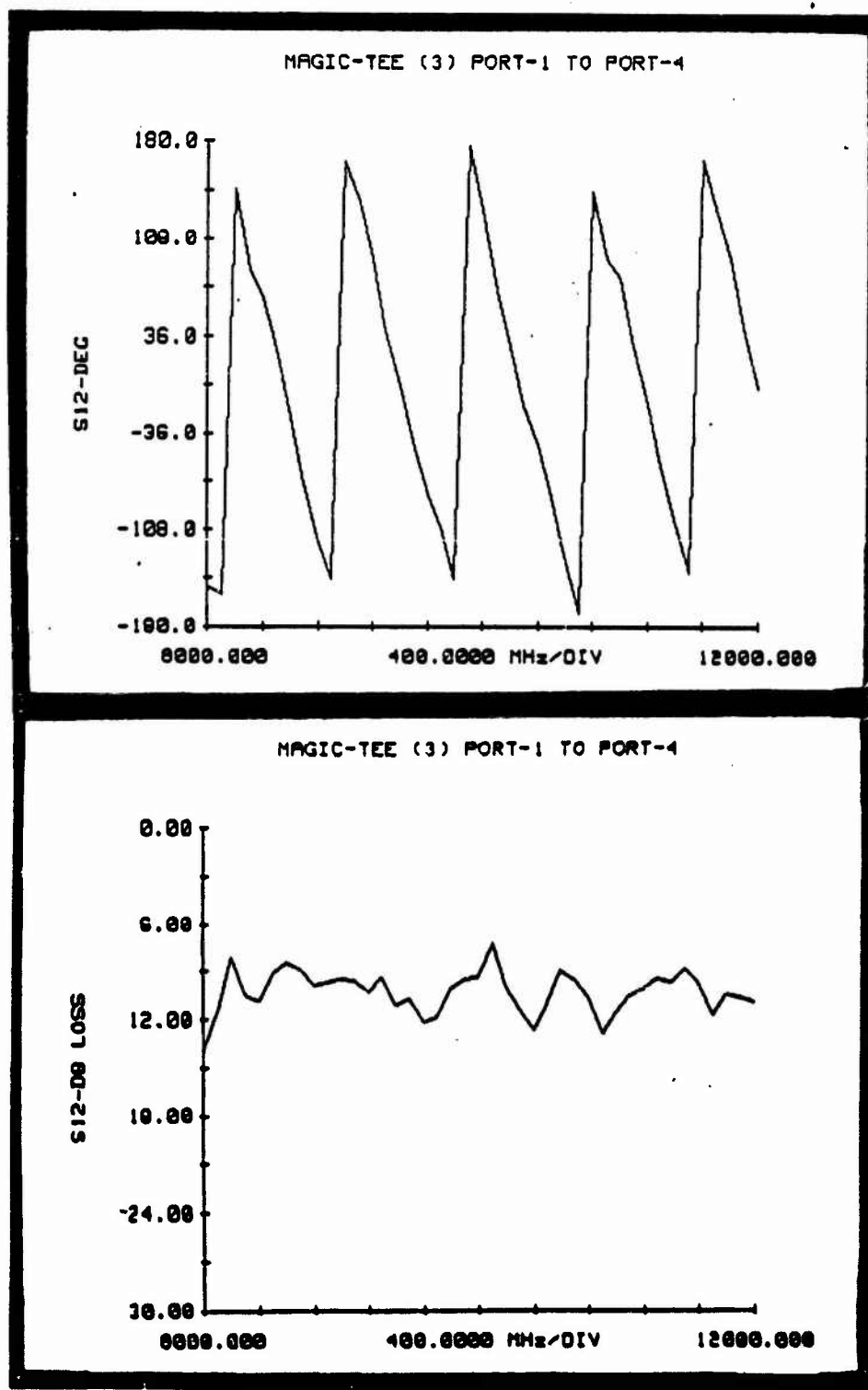
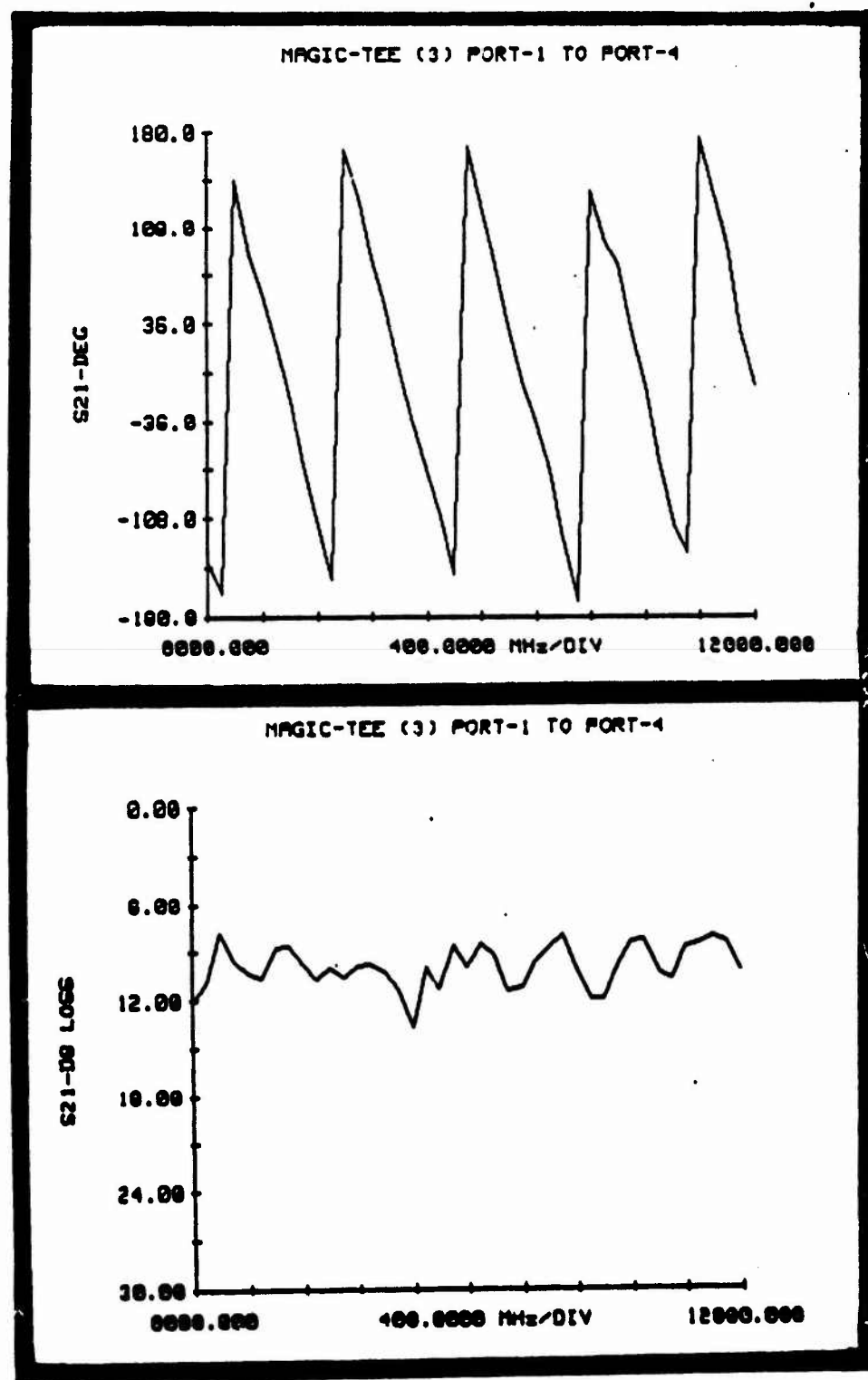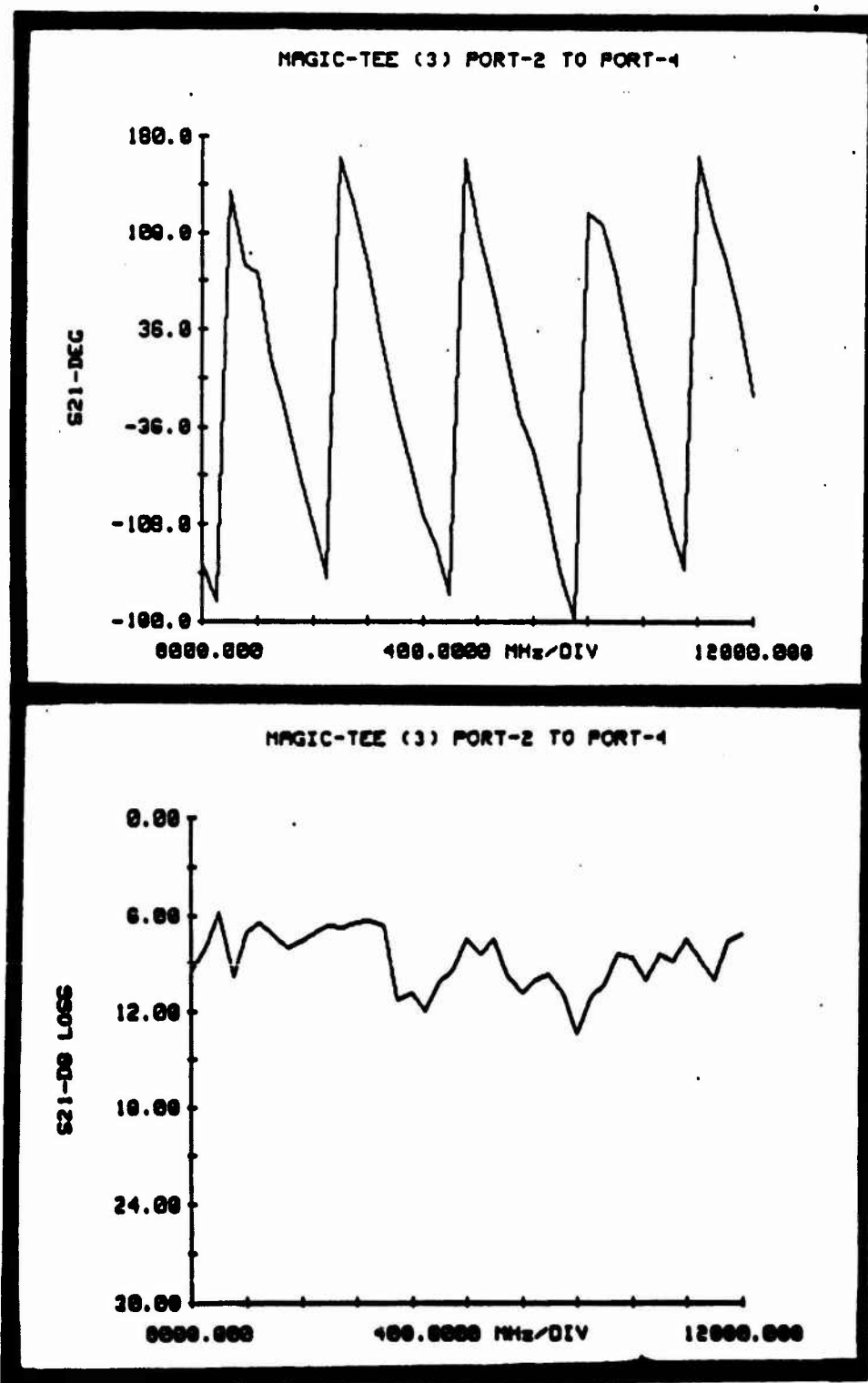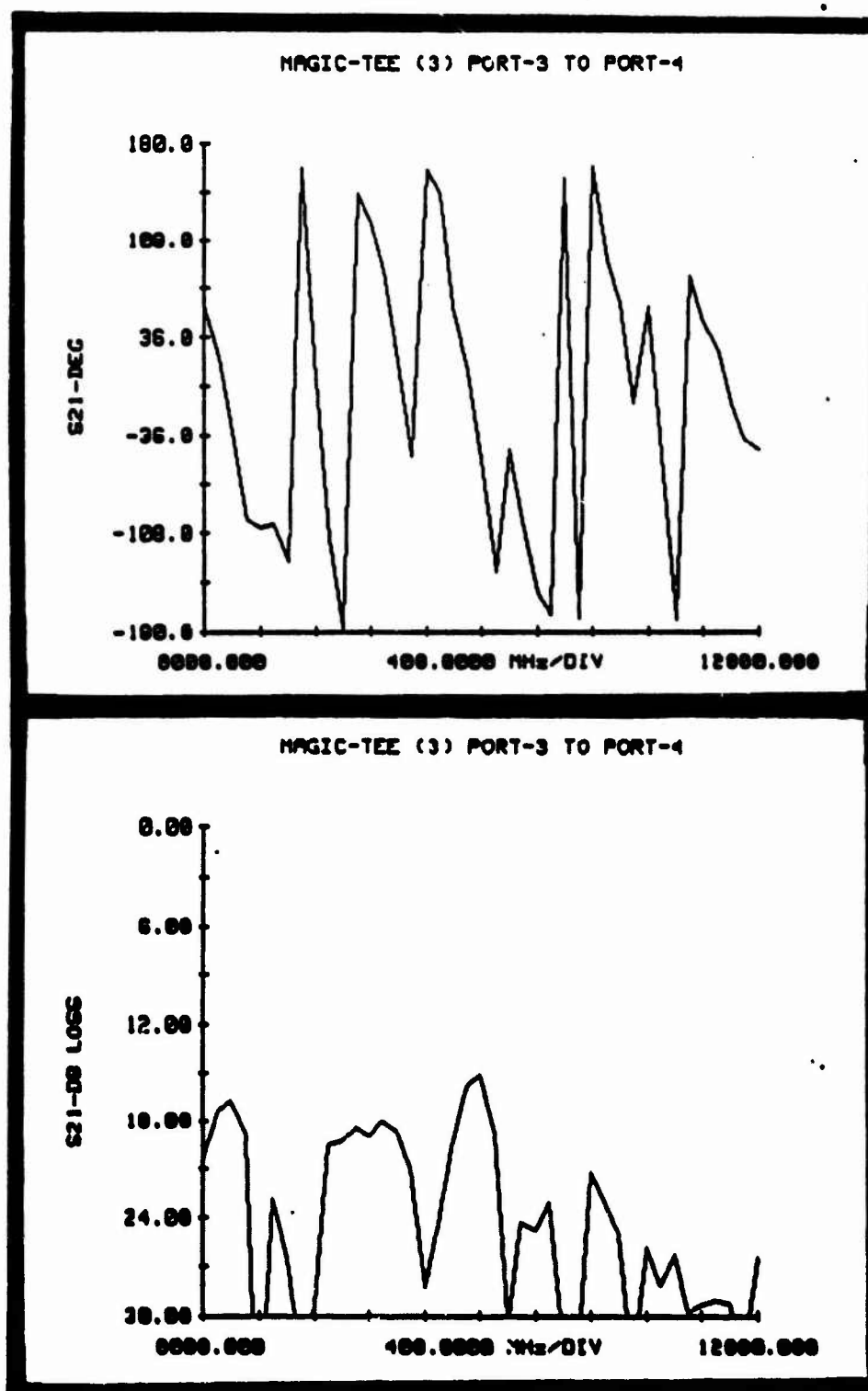Figure 21   S14 Phase and Magnitude for Magic-Tee Three.
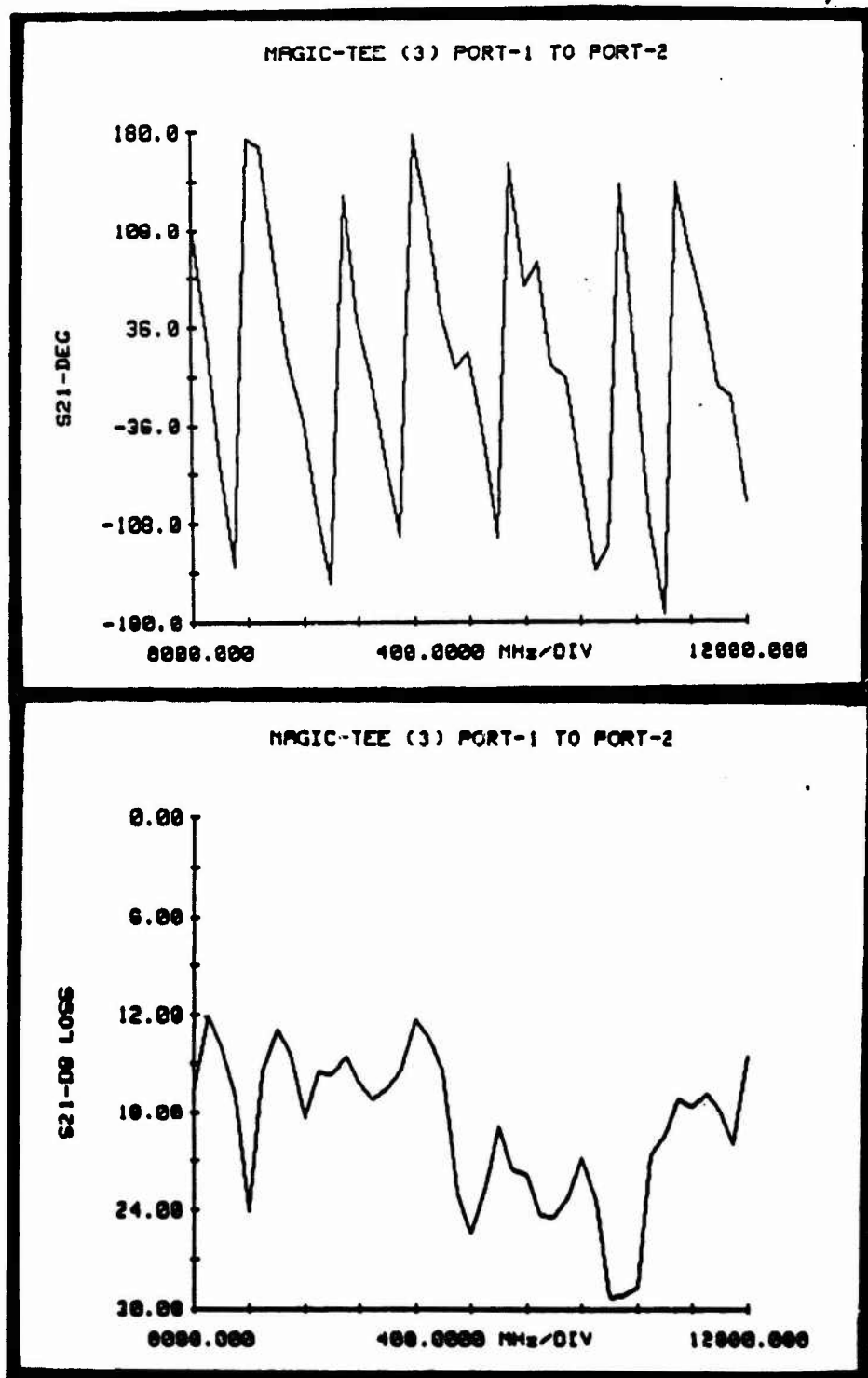
Figure 22   S21 Phase and Magnitude for Magic-Tee Three.
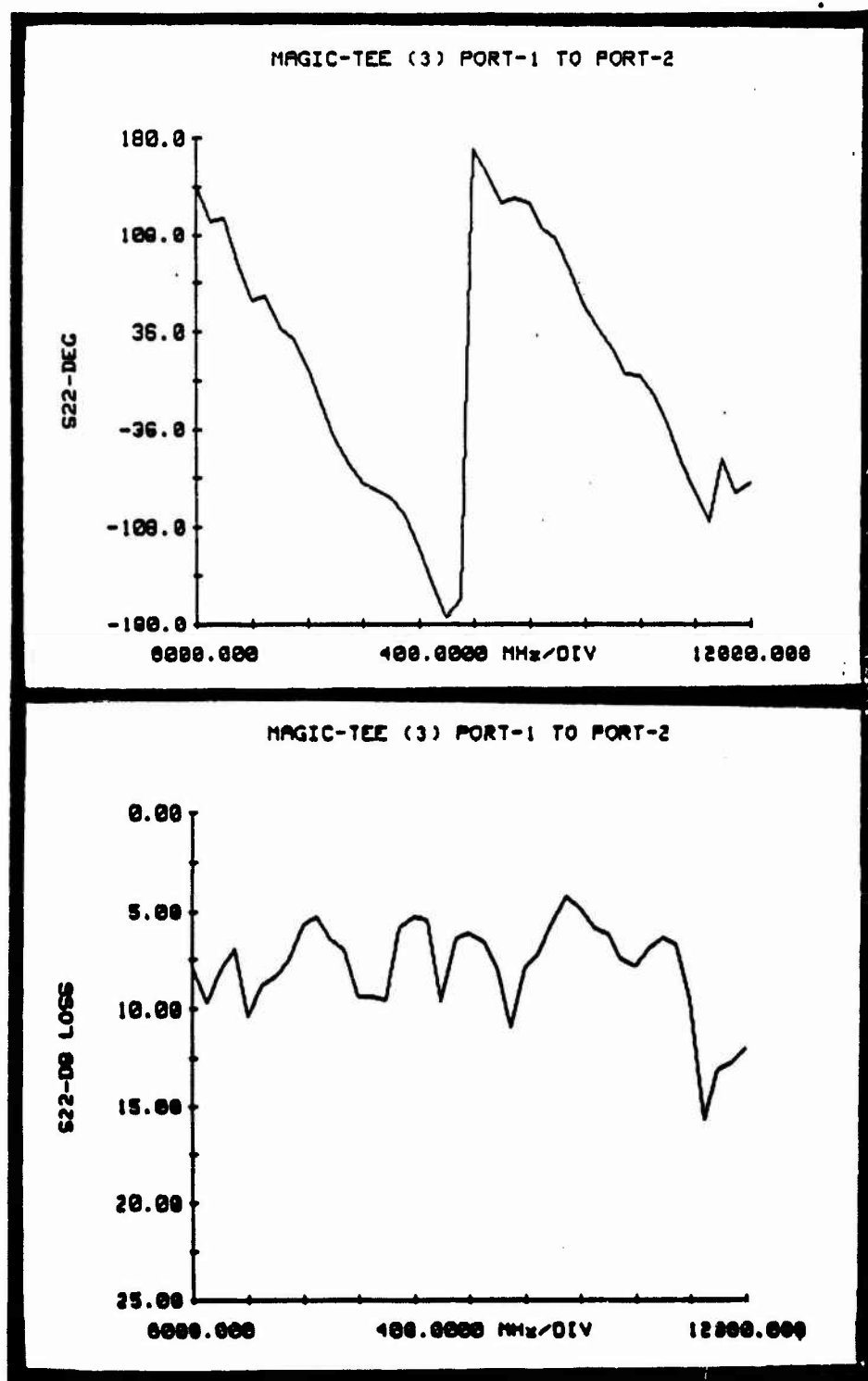
Figure 23 S22 Phase and Magnitude for Magic-Tee Three.
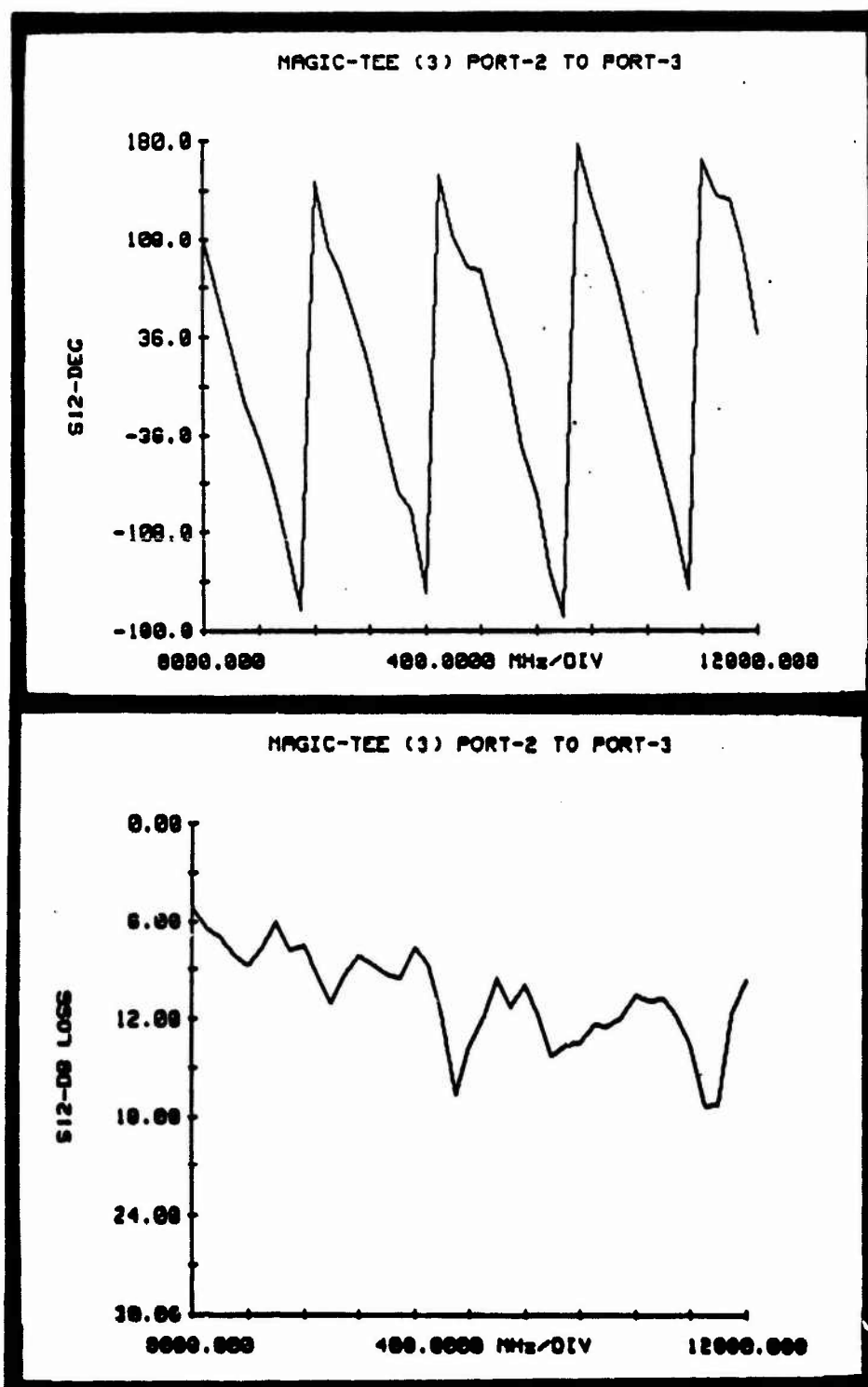
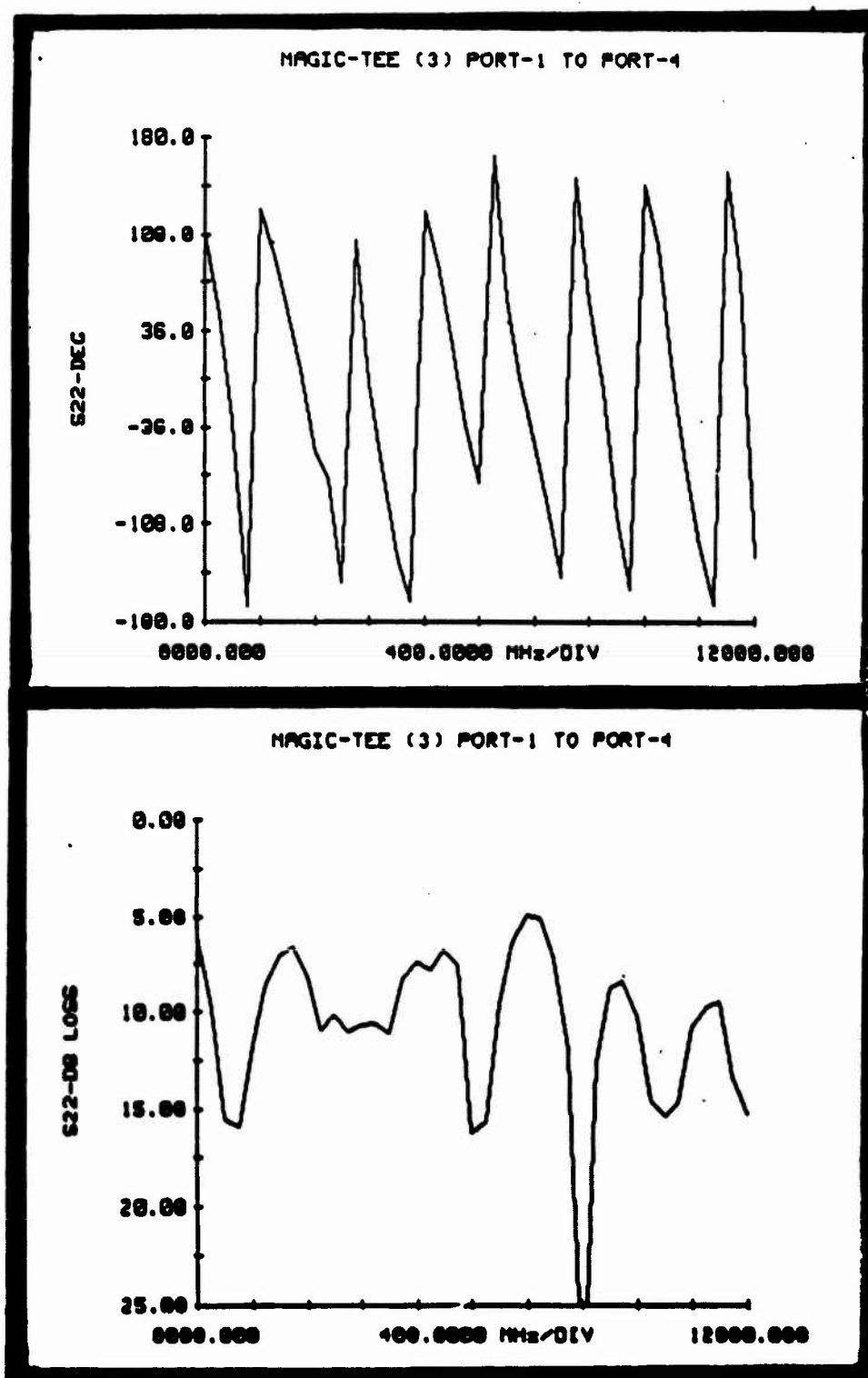Figure 24 S23 Phase and Magnitude for Magic-Tee Three.
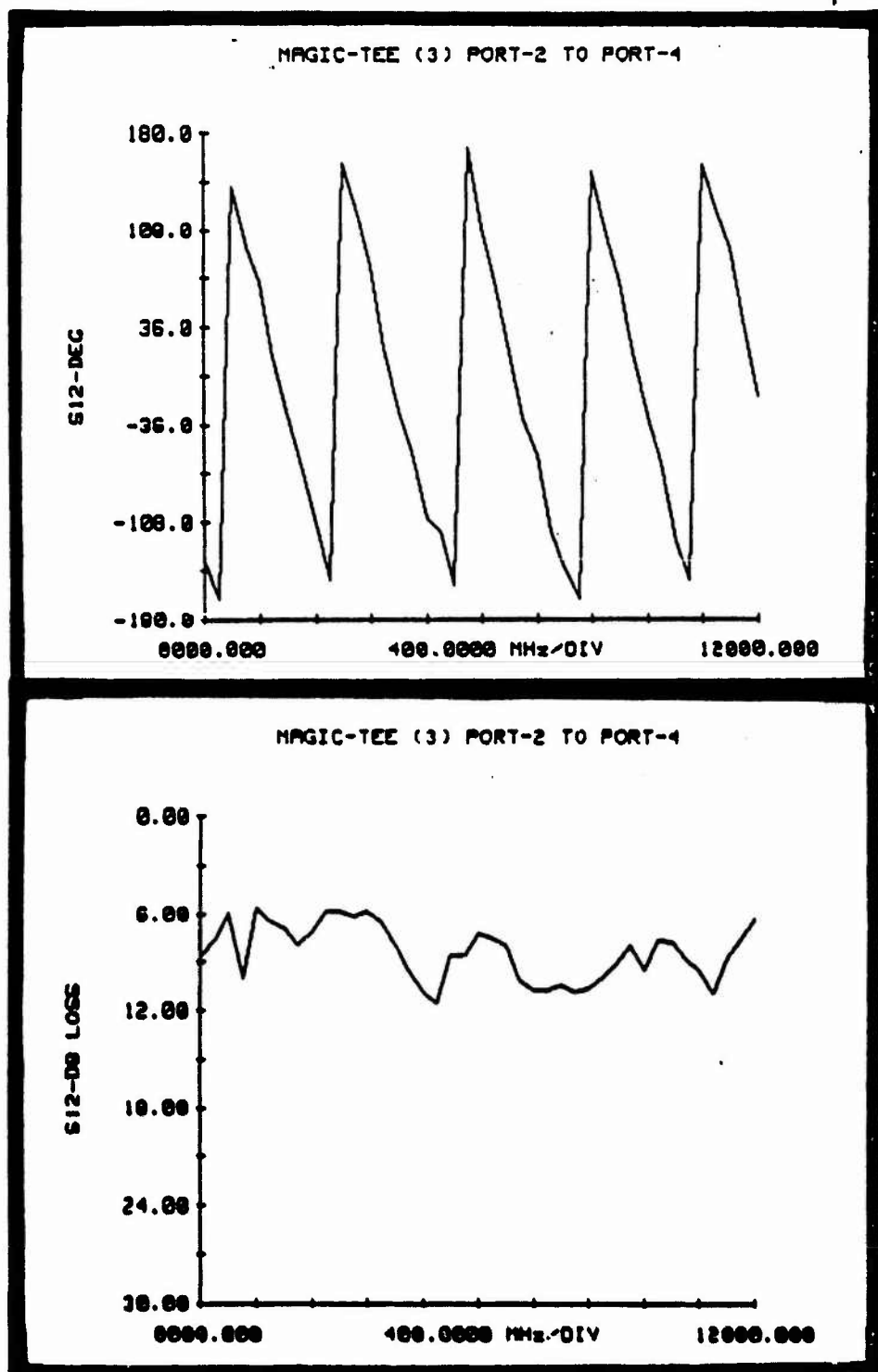
Figure 25 S24 Phase and Magnitude for Magic-Tee Three.
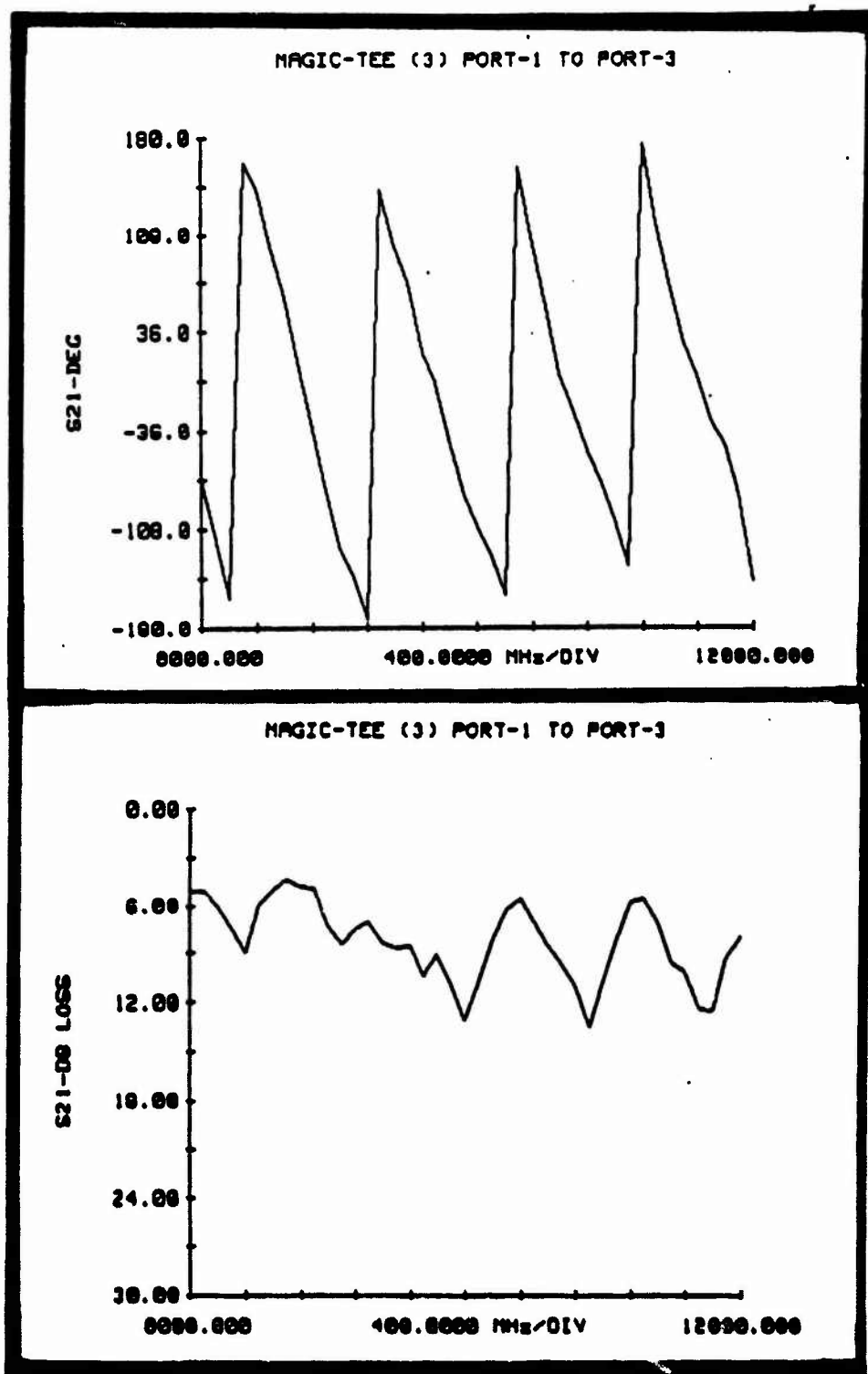
Figure 26   S31 Phase and Magnitude for Magic-Tee Three.
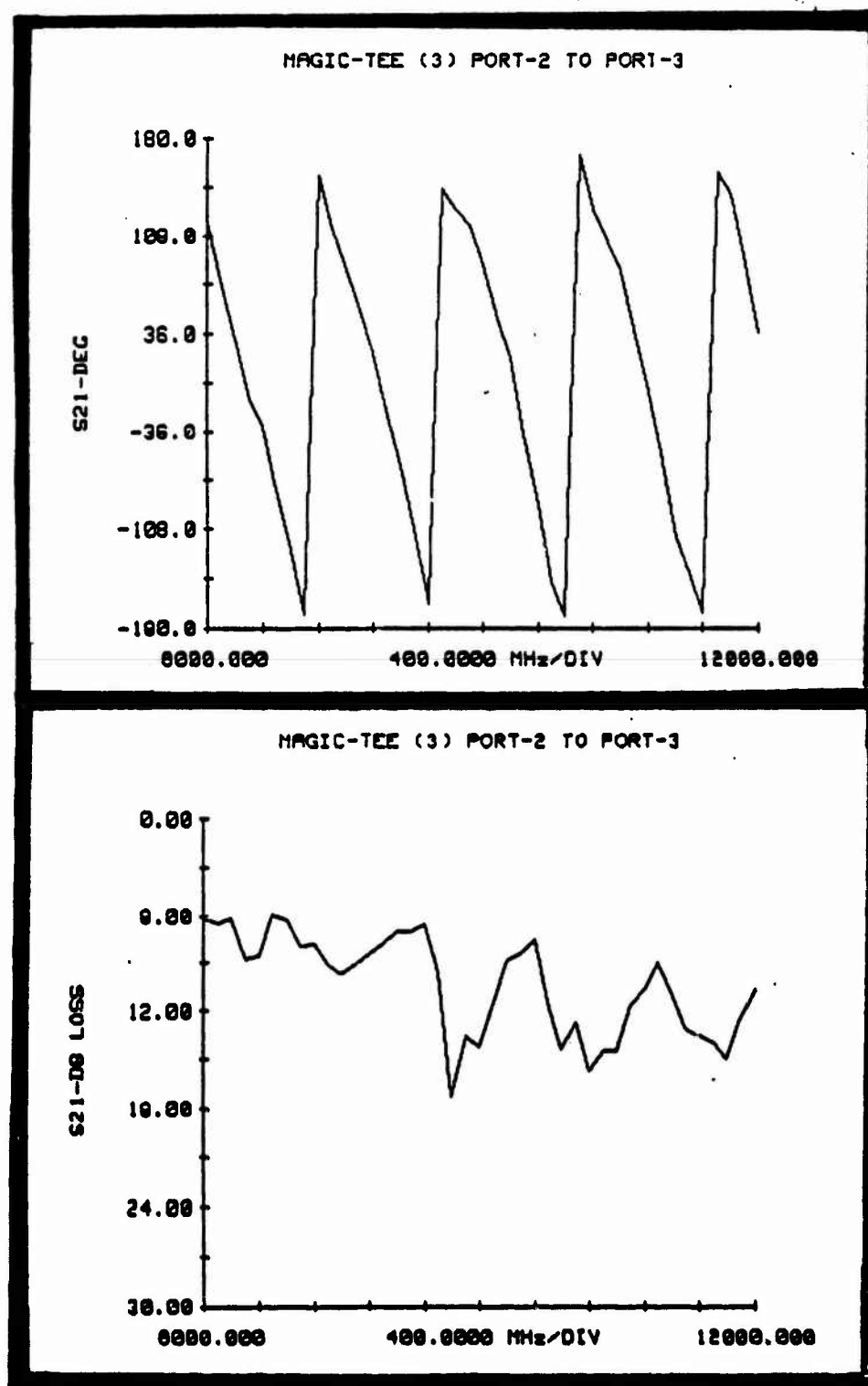
Figure 27   S32 Phase and Magnitude for Magic-Tee Three.

96
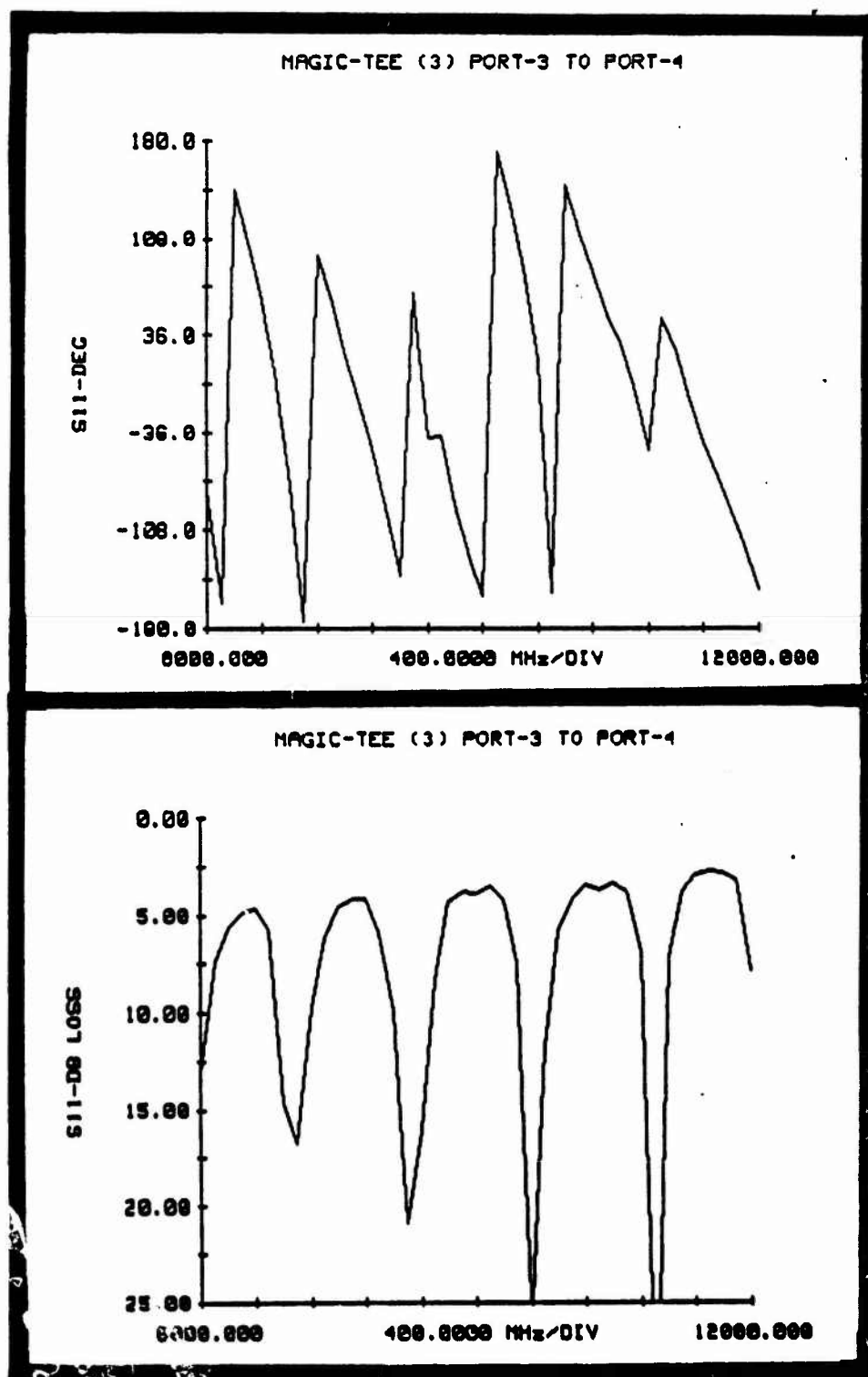
Figure 28 S33 Phase and Magnitude for Magic-Tee Three.

97
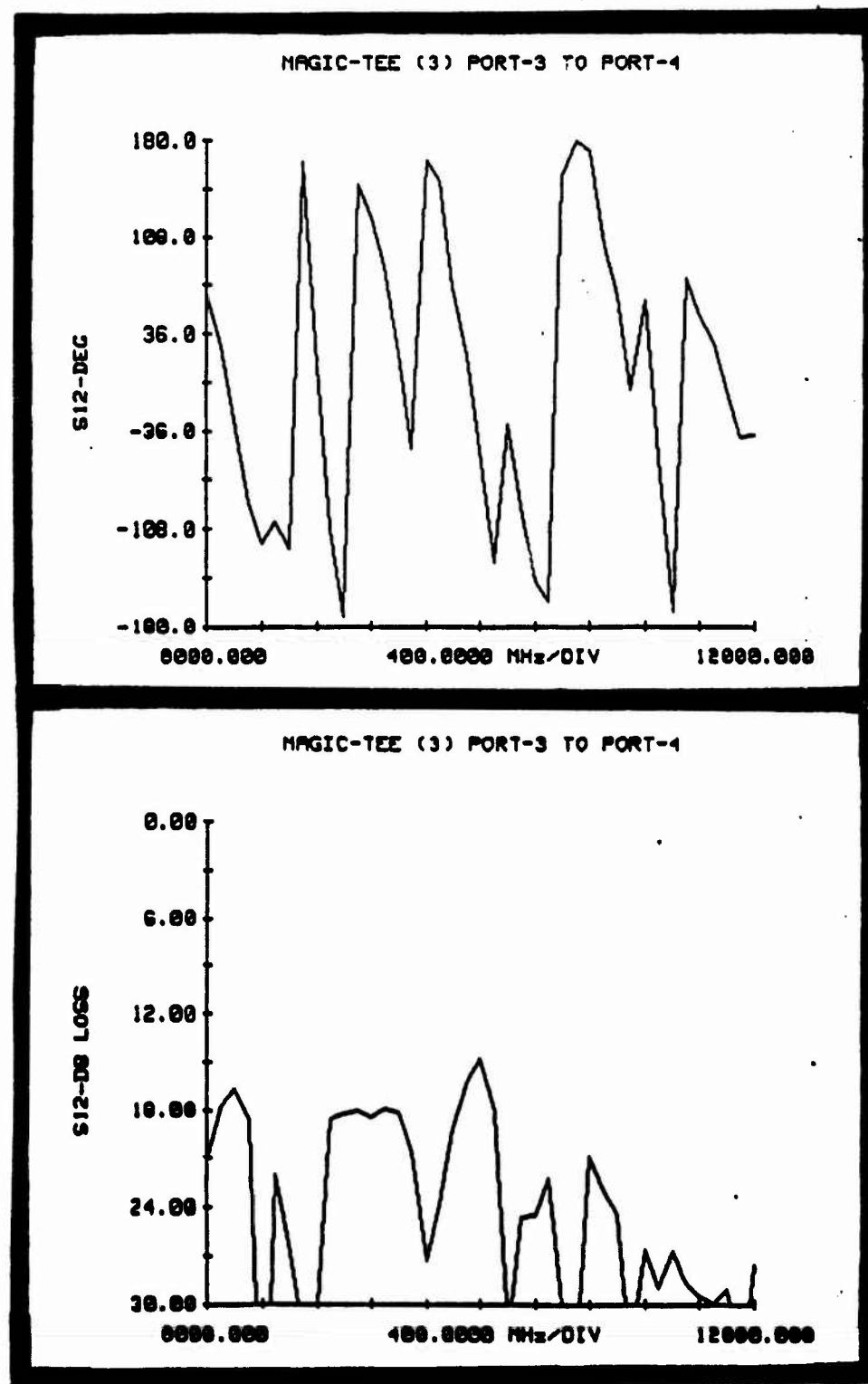
Figure 29   S34 Phase and Magnitude for Magic-Tee Three.

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Figure 38 S41 Phase and Magnitude for Magic-Tee Three.

Figure 31   S42 Phase and Magnitude for Magic-Tee Three.

Figure 32 S43 Phase and Magnitude for Magic-Tee Three.

Figure 33   S44 Phase and Magnitude for Magic-Tee Three.

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FINLINE MONOPULSE COMPARATOR

Sum Port.

Microstrips.

Difference Port.
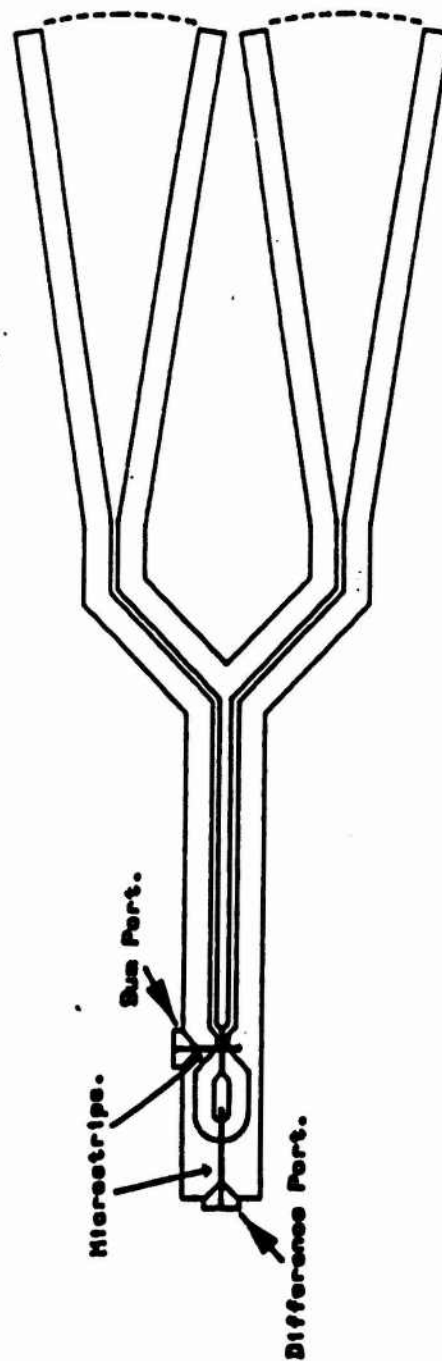
Figure 34   Fin-Line Monopulse System.
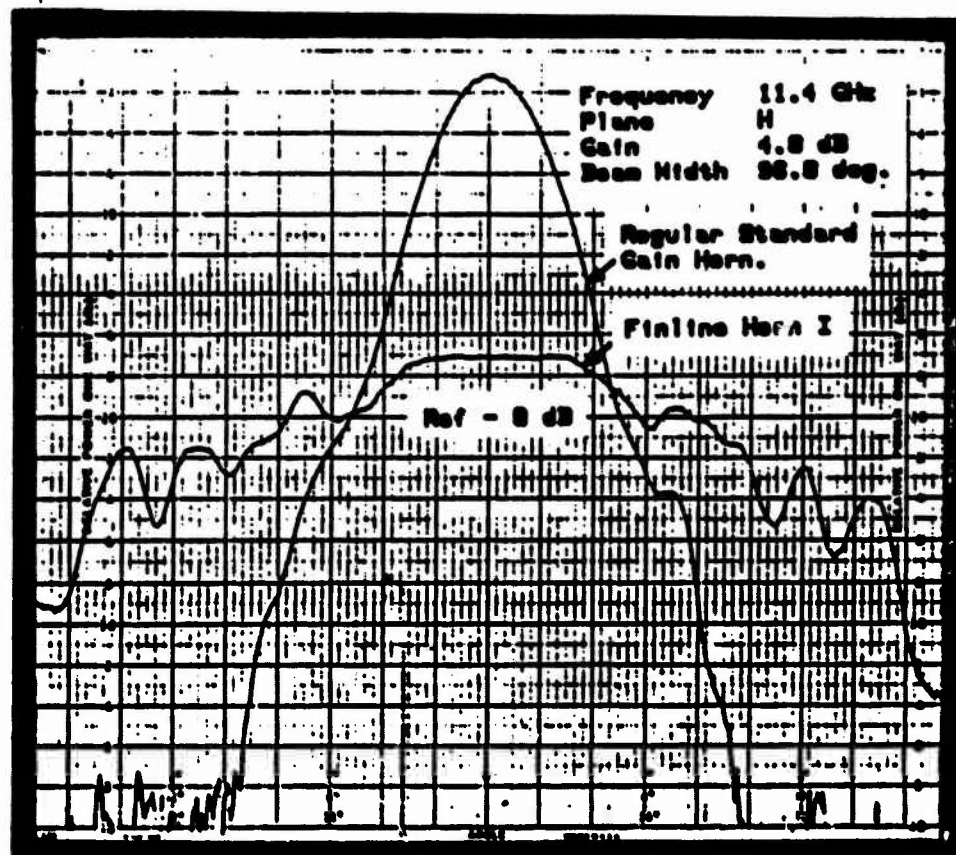
Figure 35 H-Plane Element Pattern.
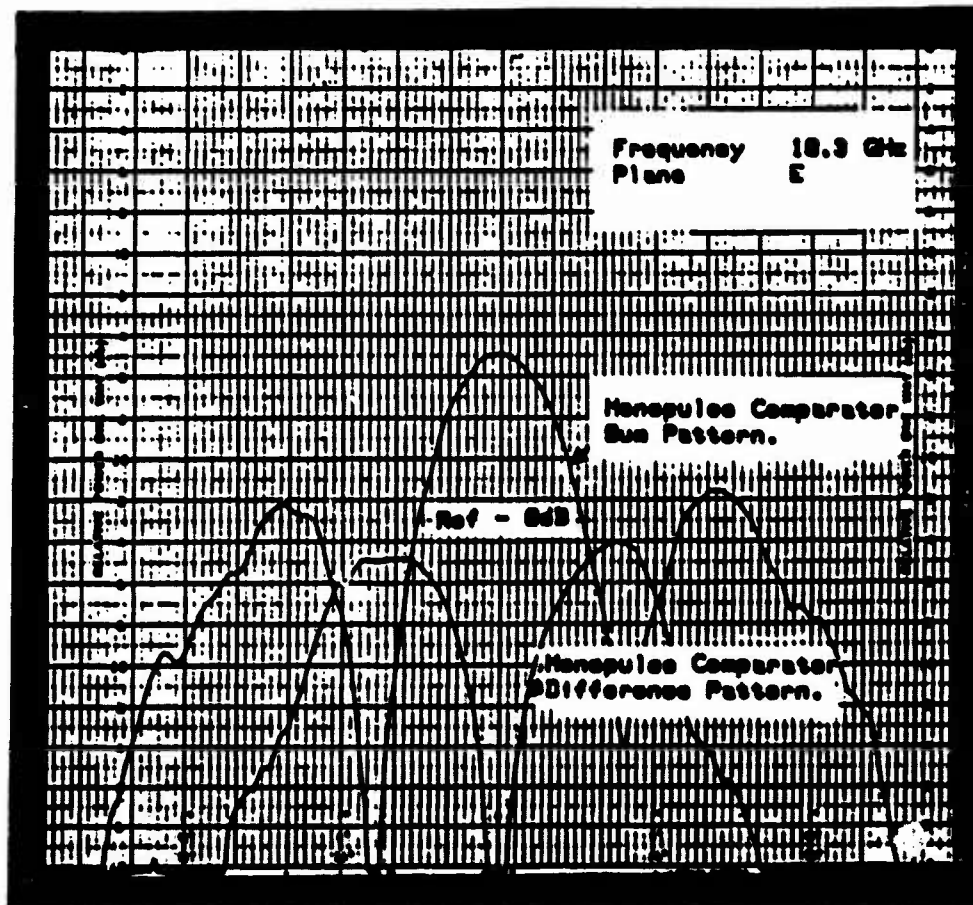
Figure 36 E-Plane System Pattern.
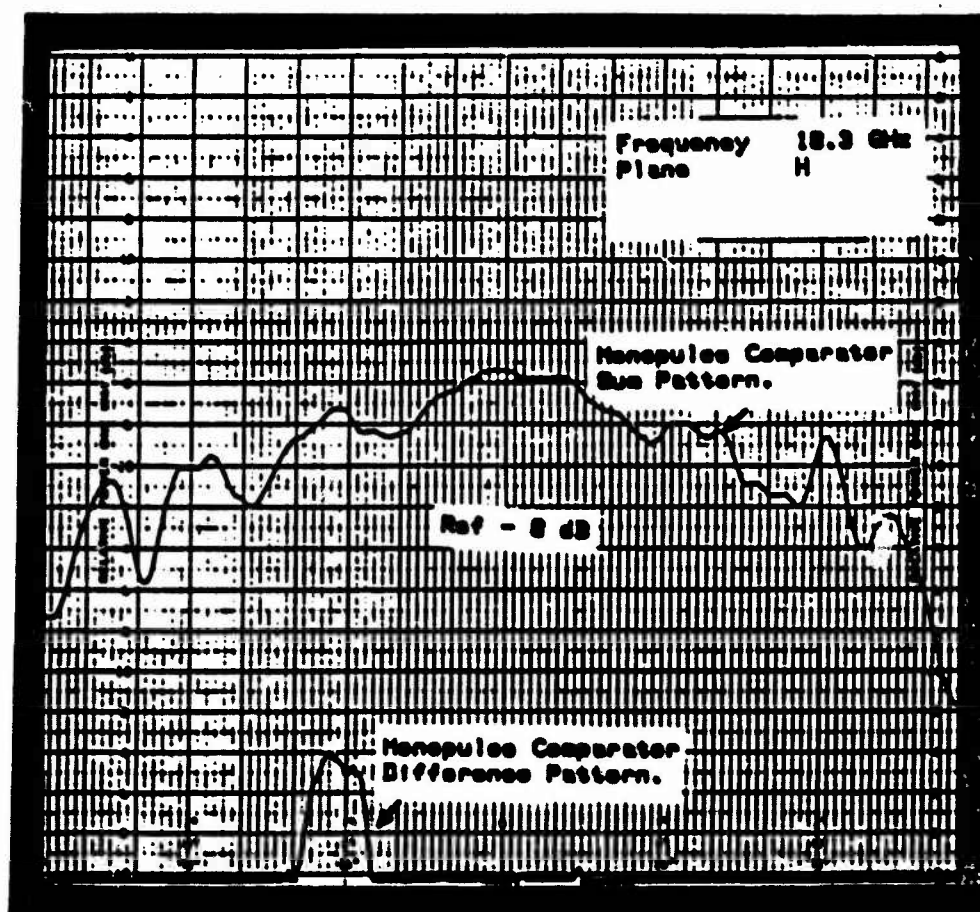
Figure 37 H-Plane System Pattern.
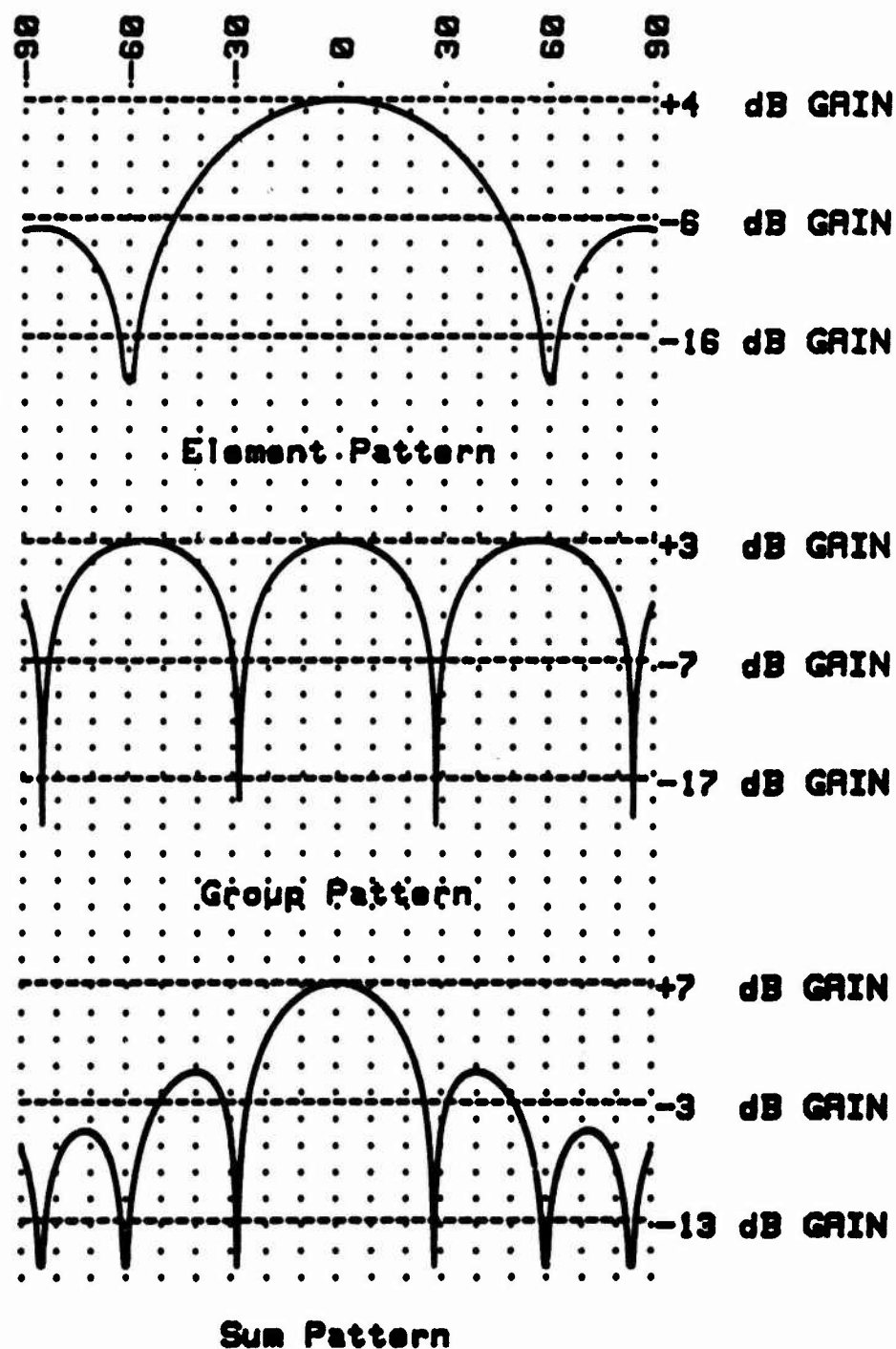
106

Degrees Left(-) and Right(+) of Boresight
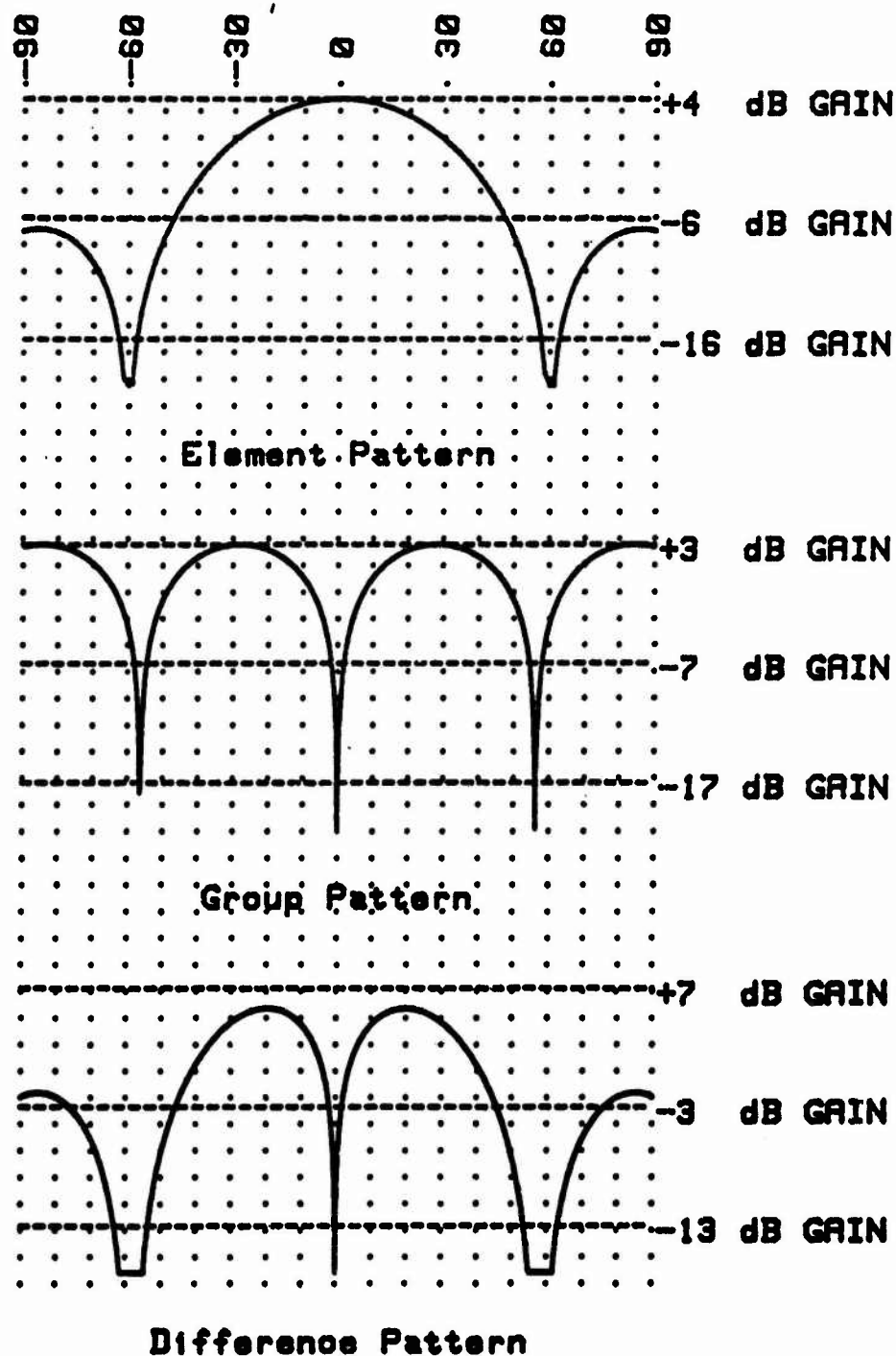
Figure 38 Simulated Sum Pattern.
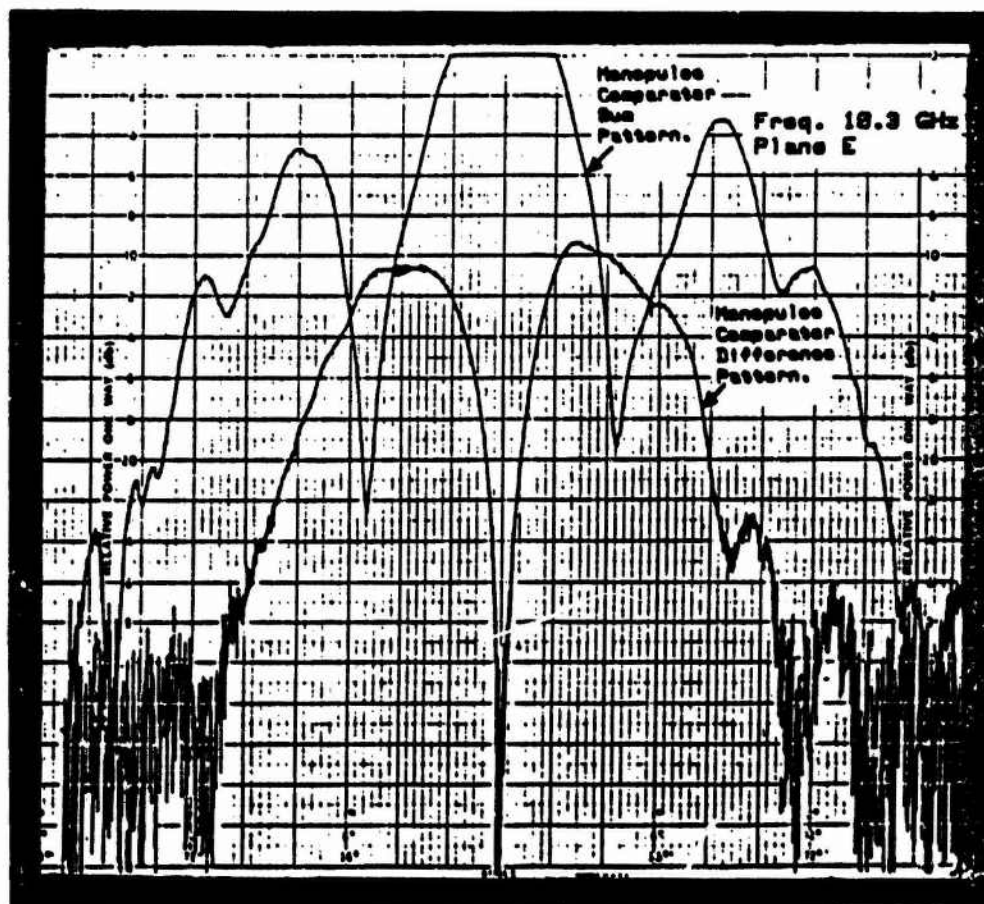
Figure 39 Simulated Difference Pattern.
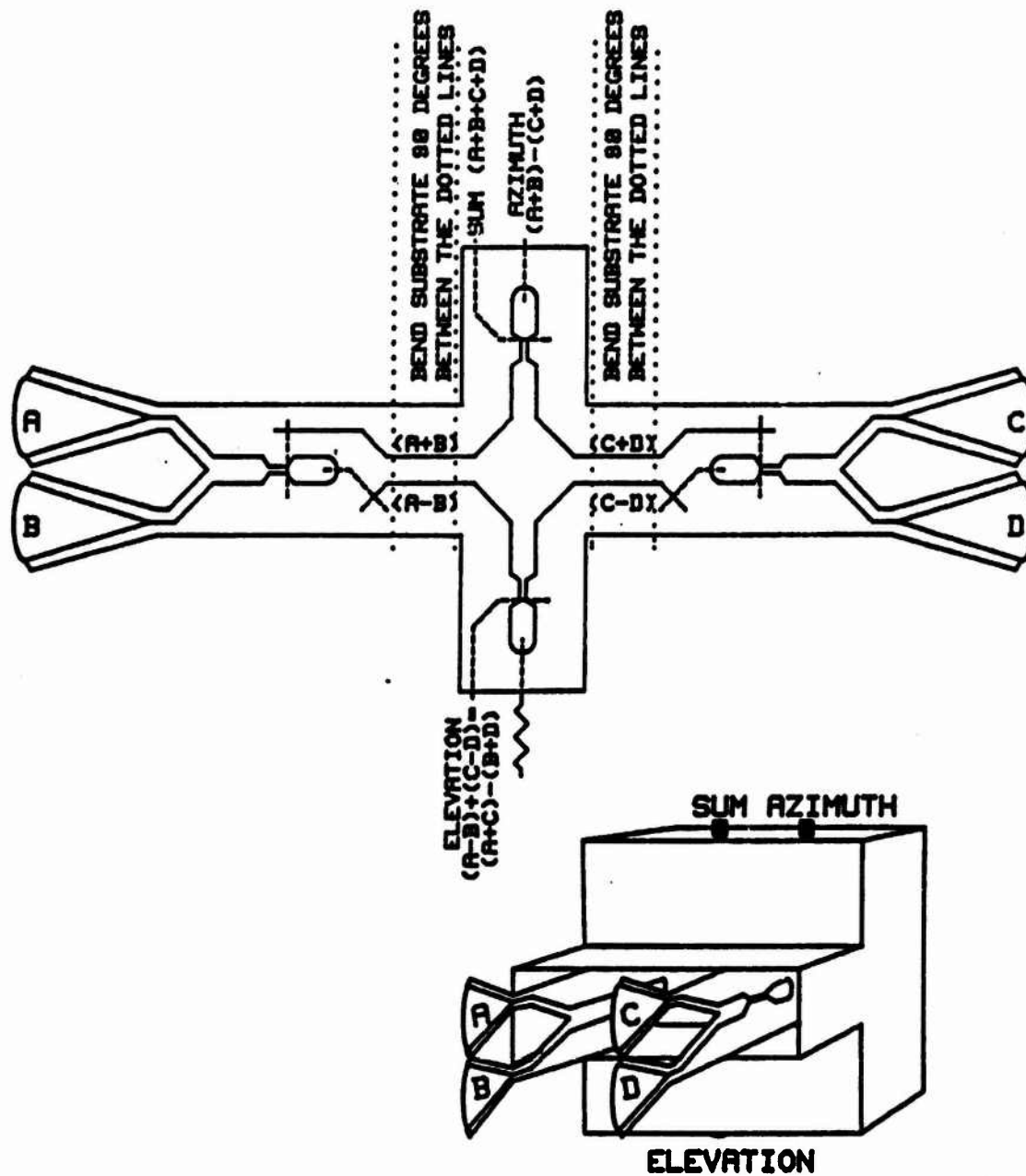
Figure 48 Depth of the Difference Null.

Figure 41   Dual Plane Fin-Line Monopulse System.

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